

Lecture 4

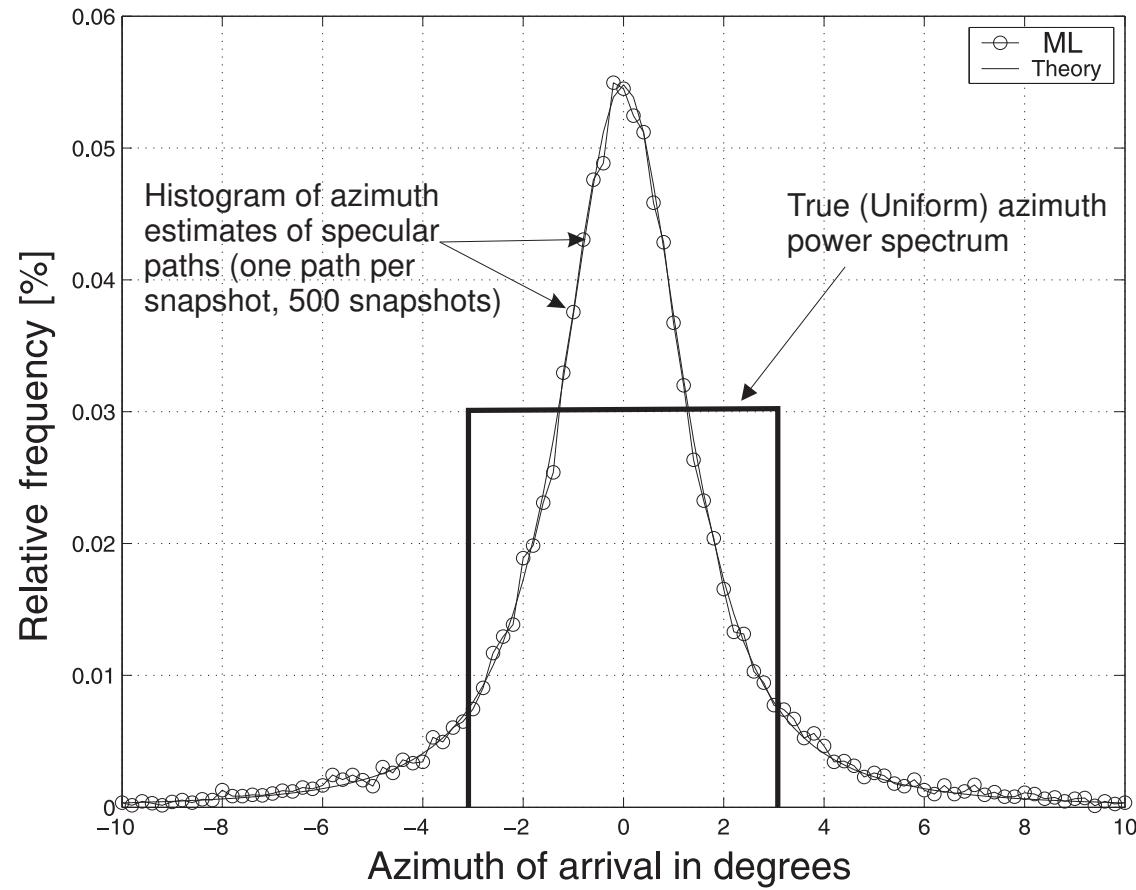
Characterization and estimation of dispersed components in the channel response

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Estimation of Path Dispersion

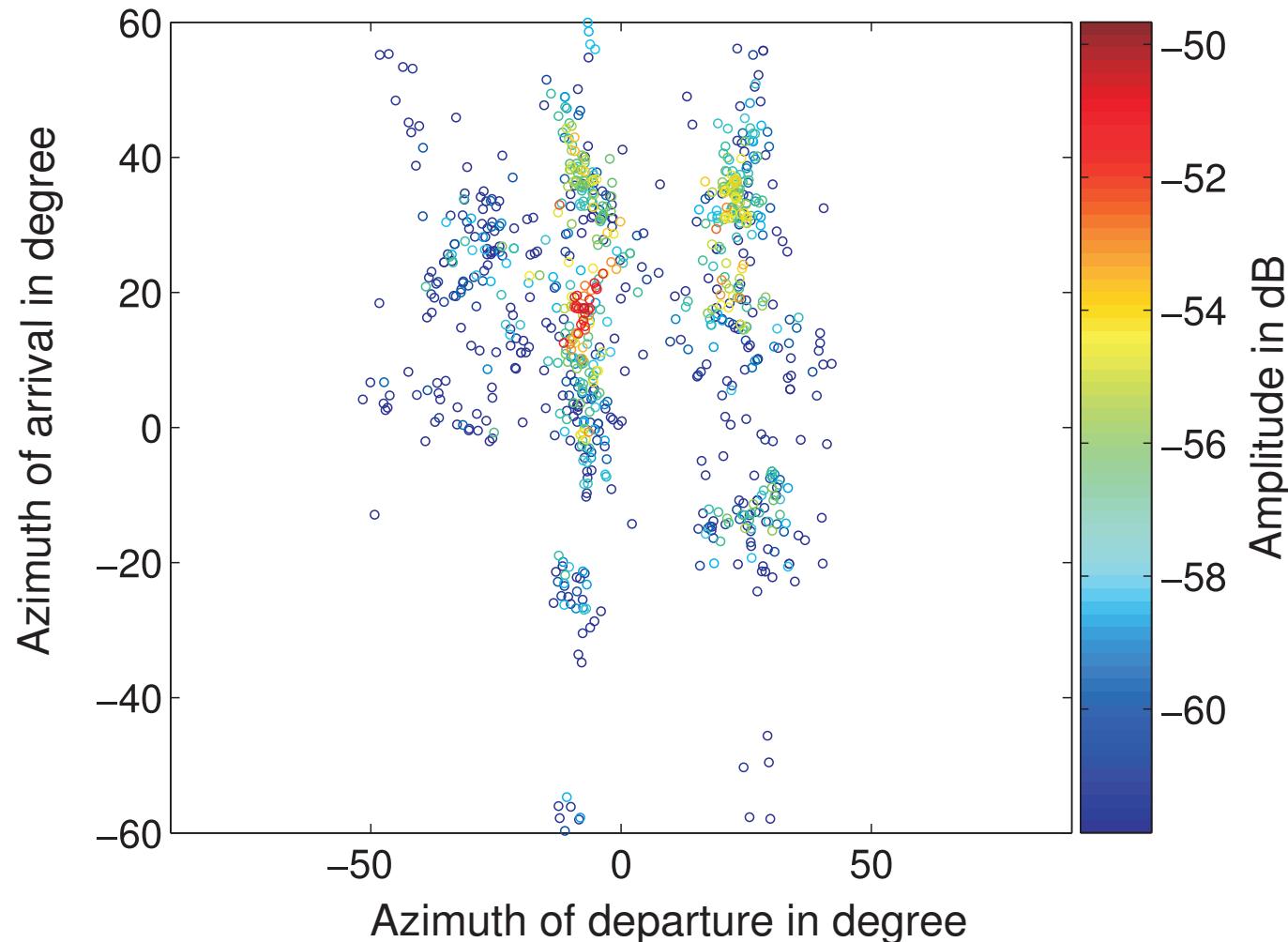
- Beamformers, such as Bartlett and Capon beamformers
- Clustering of the estimated dispersion parameters of specular components
- Estimate the parameters of (simple) parametric models characterizing path dispersion

Performance of specular-based estimators in dispersive scenarios



An Experimental Example of Specular Path Estimates in an Indoor Environment

- Wideband channel sounder RUSK ATM
- A SAGE algorithm is used to estimate multiple specular paths.



Dispersion of Individual Components

- Probability density functions (pdfs) are used to characterize the power spectral density function (psdf) of individual components:
 - ◆ Azimuth: Von-Mises pdf
 - ◆ Azimuth-elevation: Fisher-Bingham-5 pdf
 - ◆ Azimuth-elevation-delay: Von-Mises-Fisher pdf
 - ◆ Azimuth of arrival - azimuth of departure - delay:
Von-Mises-Fisher pdf
- These pdfs maximize the entropy under the constraint that
 - ◆ their first moments (center of gravity) are fixed
 - ◆ their second moments characterizing dispersion are fixed.

Dispersion in Azimuth

We seek for a density function with

- specified first moment $\mu_{\Omega} \doteq \int \Omega f(\Omega) d\Omega$
- the entropy maximized.

This function is the density function of the von-Mises distribution [Mardia, 1975]:

$$f(\Omega) = c(\kappa) \exp\{\kappa \bar{\Omega}^T \Omega\}$$

with

- $\bar{\Omega} = \|\mu_{\Omega}\|^{-1} \mu_{\Omega}$: the mode of the distribution
- κ : concentration parameter
- $c(\kappa)$: normalization factor.

Horizontal-only propagation: $\Omega = e(\phi) \doteq [\cos(\phi) \sin(\phi)]^T$.

Via the mapping $\phi \mapsto e(\phi)$, we obtain the azimuth density function:

$$f(\phi) = \frac{1}{2\pi I_0(\kappa)} \exp\{\kappa \cos(\phi - \bar{\phi})\}.$$

Dispersion in Biazimuth (AoA-AoD)

The entropy-maximizing density function $f(\Omega_1, \Omega_2)$ with specified

- first moments $\mu_{\Omega_1}, \mu_{\Omega_2}$
- second moments in $\int \Omega_1 \Omega_2^T f(\Omega_1, \Omega_2) d\Omega_1 d\Omega_2$

is the von-Mises-Fisher density function [Mardia, 1975]

$$f(\Omega_1, \Omega_2) \propto \exp\{\mathbf{a}_1^T \Omega_1 + \mathbf{a}_2^T \Omega_2 + \Omega_1^T \mathbf{A} \Omega_2\},$$

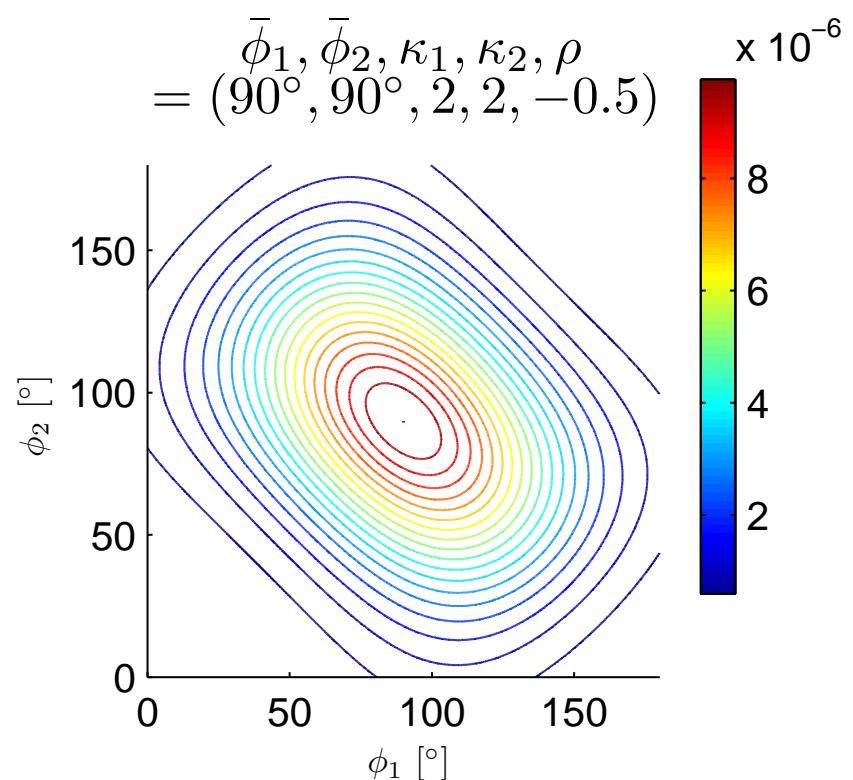
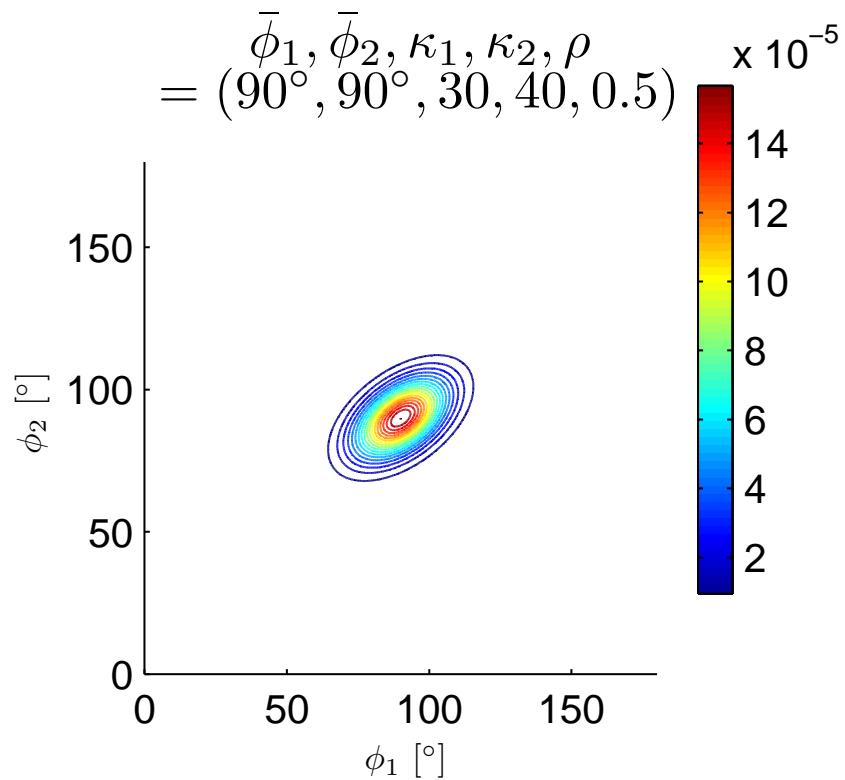
with \mathbf{a}_1 , \mathbf{a}_2 and \mathbf{A} being free parameters.

A biazimuth density function $f(\phi_1, \phi_2)$ derived via the mapping $[\phi_1, \phi_2] \mapsto [\mathbf{e}(\phi_1)^T, \mathbf{e}(\phi_2)^T]$ reads,

$$\begin{aligned} f(\phi_1, \phi_2) &= c(\kappa_1, \kappa_2, \rho) \cdot \exp \left\{ \left(\frac{\kappa_1 - \rho \sqrt{\kappa_1 \kappa_2}}{1 - \rho^2} \right) \cos(\phi_1 - \bar{\phi}_1) \right. \\ &\quad \left. + \left(\frac{\kappa_2 - \rho \sqrt{\kappa_1 \kappa_2}}{1 - \rho^2} \right) \cos(\phi_2 - \bar{\phi}_2) + \frac{\rho \sqrt{\kappa_1 \kappa_2}}{1 - \rho^2} \cos[(\phi_1 - \bar{\phi}_1) - (\phi_2 - \bar{\phi}_2)] \right\} \end{aligned}$$

Dispersion in Biazimuth (Cont.)

Contour plots of the biazimuth density function:



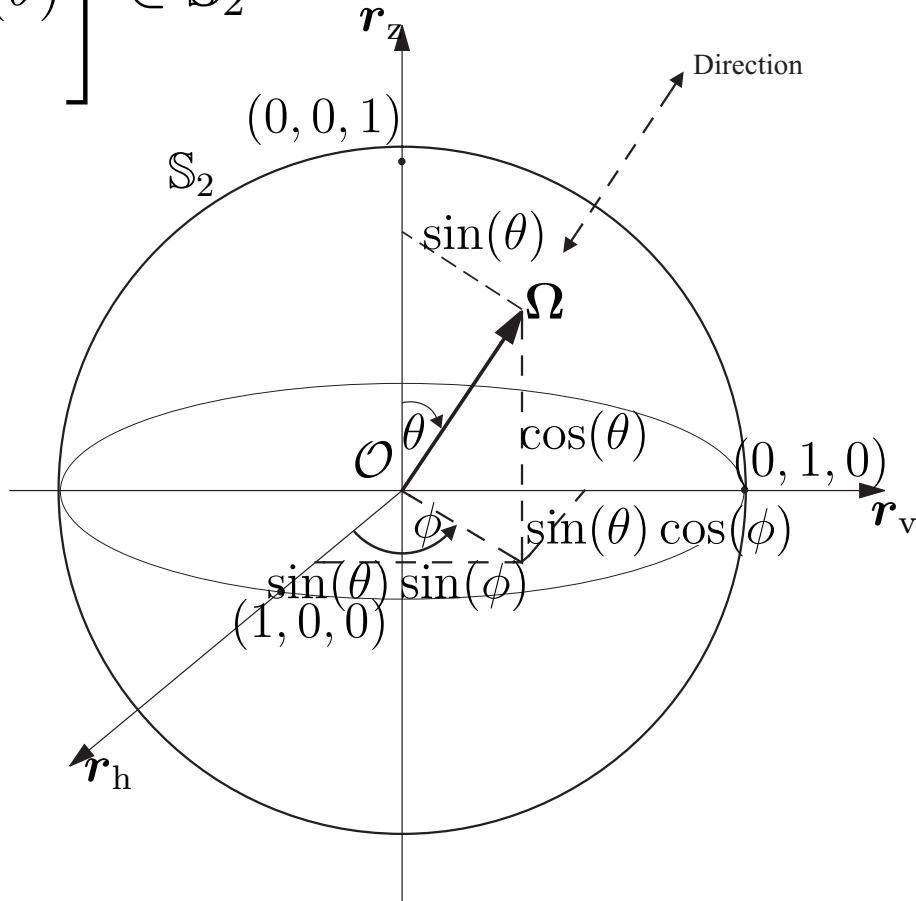
Dispersion in Direction (Azimuth-Elevation)

Definition of a direction:

$$\Omega = e(\phi, \theta) \doteq \begin{bmatrix} \cos(\phi) \sin(\theta) \\ \sin(\phi) \sin(\theta) \\ \cos(\theta) \end{bmatrix} \in \mathbb{S}_2$$

where

- \mathbb{S}_2 : unit sphere
- ϕ : azimuth
- θ : elevation



Dispersion in Direction (Azimuth-Elevation)

The entropy-maximizing density function on \mathbb{S}_2 with specified

- first moment (mean direction) $\mu_{\Omega} \doteq \int \Omega f(\Omega) d\Omega$
- second moment matrix $\int \Omega \Omega^T f(\Omega) d\Omega$

is the density function of the Fisher-Bingham 5 distribution [Kent 1982]:

$$f(\Omega) = c(\kappa, \beta) \cdot \exp \left\{ \kappa \gamma_1^T \Omega + \beta [(\gamma_2^T \Omega)^2 - (\gamma_3^T \Omega)^2] \right\},$$

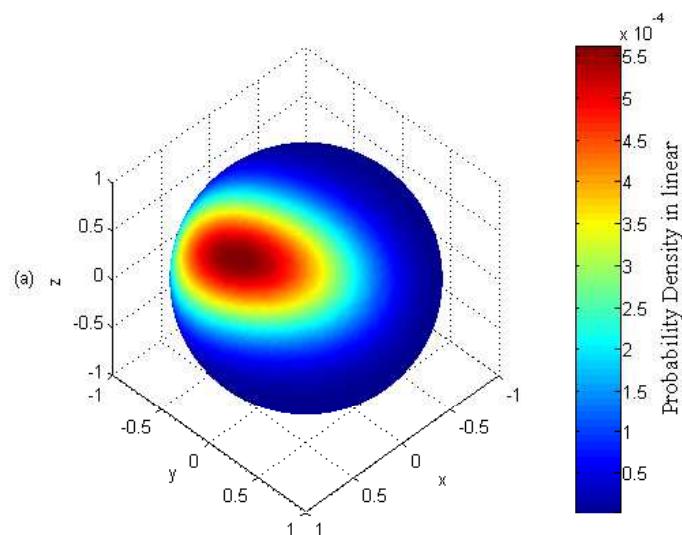
where

- κ : concentration parameter
- β : ovalness parameter
- $\gamma_1, \gamma_2, \gamma_3$: orthonormal vectors determined by angles $\bar{\phi}, \bar{\theta}, \alpha$
- $\bar{\phi}$: azimuth of the mean direction μ_{Ω}
- $\bar{\theta}$: elevation of μ_{Ω}
- α : angle describing how the distribution is tilted on \mathbb{S}_2 .

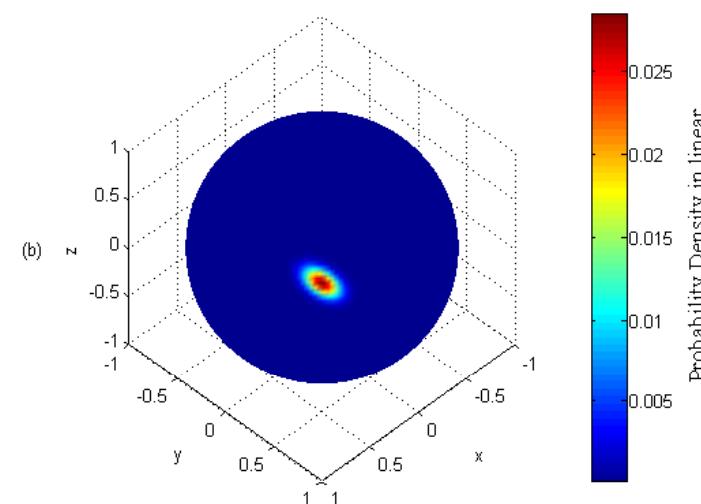
Dispersion in Direction (Azimuth-Elevation) (Cont.)

Two examples of the direction density function $f(\Omega)$:

$$(\bar{\phi}, \bar{\theta}, \alpha, \kappa, \beta) = \\ (0^\circ, 45^\circ, 160^\circ, 5, 1.5)$$



$$(\bar{\phi}, \bar{\theta}, \alpha, \kappa, \beta) = \\ (45^\circ, 70^\circ, 35^\circ, 200, 100)$$



The azimuth-elevation density function $f(\phi, \theta)$ induced by $f(\Omega)$ reads

$$f(\phi, \theta) = c(\kappa, \beta) \cdot \exp \left\{ \kappa \boldsymbol{\gamma}_1^T \boldsymbol{\Omega} + \beta [(\boldsymbol{\gamma}_2^T \boldsymbol{\Omega})^2 - (\boldsymbol{\gamma}_3^T \boldsymbol{\Omega})^2] \right\} \cdot \sin(\theta).$$

Dispersion in Biazimuth and Delay

Define $\psi \doteq [\Omega_1^T, \Omega_2^T, \tau]^T$.

The entropy-maximizing density function $f(\psi)$ with specified

- first moments μ_ψ
- second moments $\int \psi \psi^T f(\psi) d\psi$

is [Mardia, 1975]

$$f(\psi) \propto \exp\{\mathbf{b}^T \psi + \psi^T \mathbf{B} \psi\},$$

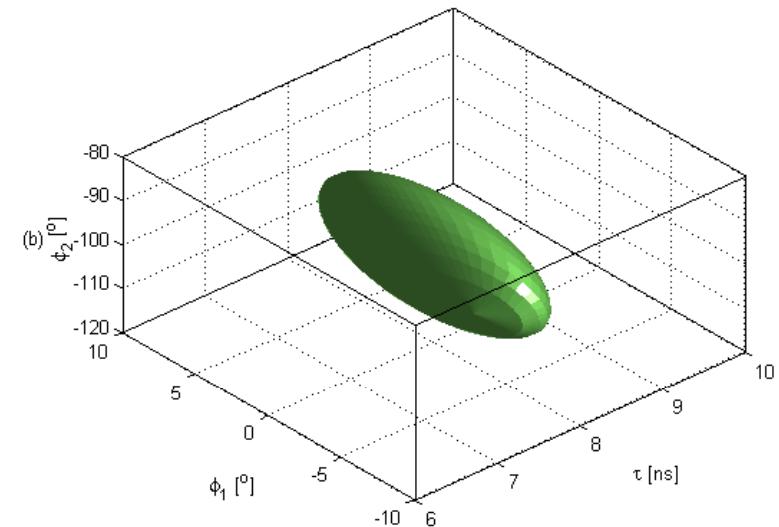
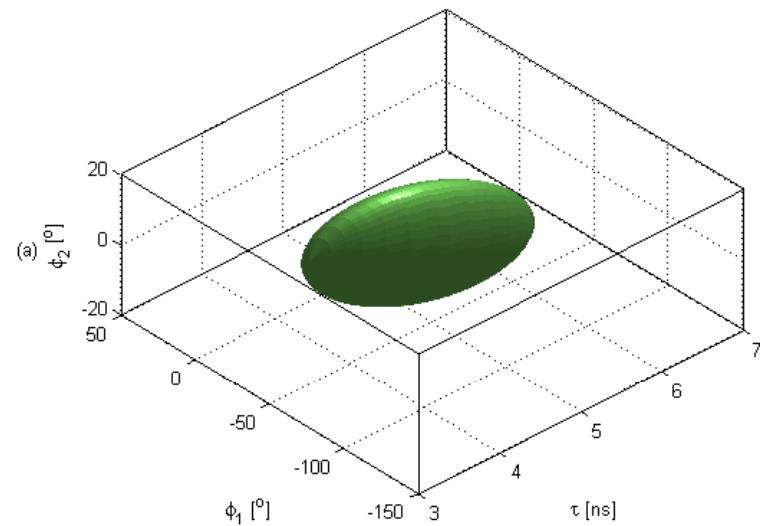
where $\mathbf{b} \in \mathbb{R}^{5 \times 1}$ and $\mathbf{B} \in \mathbb{R}^{5 \times 5}$ are free parameters.

Via the mapping $[\phi_1, \phi_2, \tau] \mapsto [\mathbf{e}(\phi_1)^T, \mathbf{e}(\phi_2)^T, \tau]$, the biazimuth-delay density function $f(\phi_1, \phi_2, \tau)$ can be induced by $f(\Omega_1, \Omega_2, \tau)$,

$$\begin{aligned} f(\phi_1, \phi_2, \tau) = C \exp \{ & c_1 \cos(\phi_1 - \bar{\phi}_1) + c_2 \cos(\phi_2 - \bar{\phi}_2) \\ & + (\tau - \bar{\tau})[c_3 \sin(\phi_1 - \bar{\phi}_1) + c_4 \sin(\phi_2 - \bar{\phi}_2)] \\ & + c_5(\tau - \bar{\tau})^2 + c_6 \cos[(\phi_1 - \bar{\phi}_1) - (\phi_2 - \bar{\phi}_2)] \}. \end{aligned}$$

Dispersion in Biaximuth and Delay (Cont.)

3dB-spread surfaces $\{(\phi_1, \phi_2, \tau) : f(\phi_1, \phi_2, \tau) = \frac{1}{2}f(\bar{\phi}_1, \bar{\phi}_2, \bar{\tau})\}$:



	$\bar{\tau} [\text{ns}]$	$\bar{\phi}_1 [^\circ]$	$\bar{\phi}_2 [^\circ]$	$\sigma_\tau [\text{ns}]$	κ_1	$\sigma_{\phi_1} [^\circ]$	κ_2	$\sigma_{\phi_2} [^\circ]$	ρ_{12}	ρ_1	ρ_2
(a)	5	-40	0	1	5	25.6	10	18.1	-0.4	-0.3	-0.3
(b)	8	0	-100	0.5	50	8.1	30	10.5	-0.5	0.6	-0.2

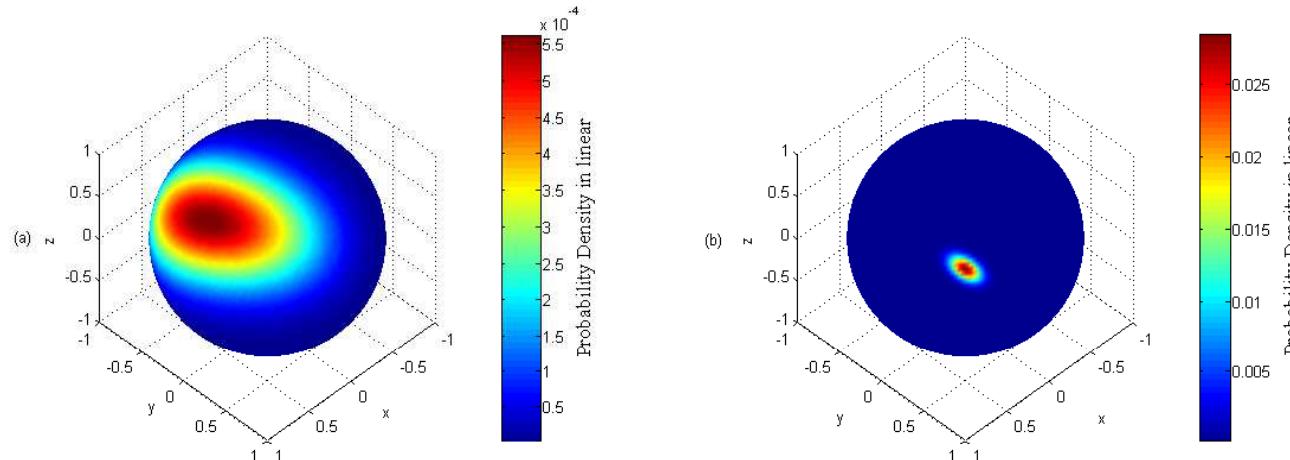
Direction-Delay Dispersion

The entropy-maximizing pdf of direction Ω and delay τ with specified first moment and second moment is of the form

$$f_{\text{ME}}(\Omega, \tau) \propto \exp \left\{ \begin{bmatrix} \Omega - \bar{\Omega} \\ \tau - \bar{\tau} \end{bmatrix}^T \begin{bmatrix} A & c \\ c^T & -b \end{bmatrix} \begin{bmatrix} \Omega - \bar{\Omega} \\ \tau - \bar{\tau} \end{bmatrix} \right\}.$$

Two conditions are used to derive the expressions of the parameters:

- The conditional pdf of direction is Fisher-Bingham-5 (FB5)
Two examples of FB5 pdf:



- The conditional pdf of delay is Gaussian.

Direction-Delay Dispersion (Cont.)

The direction-delay pdf is calculated to be

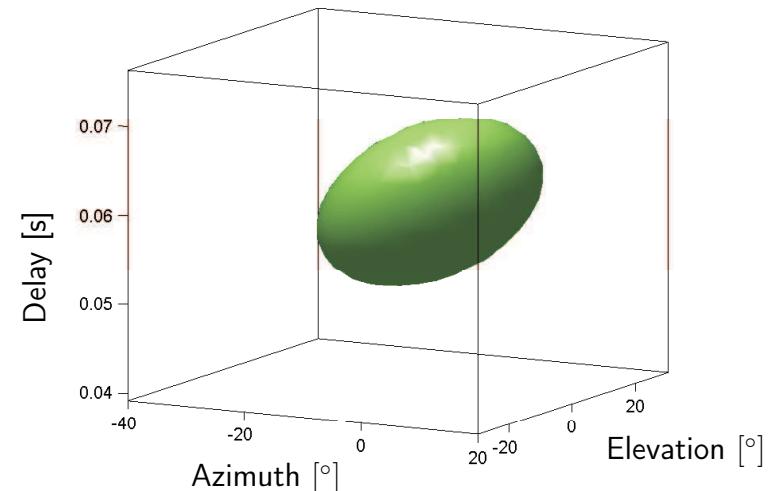
$$f_{\text{ME}}(\Omega, \tau) = C \cdot \exp\left(\kappa \bar{\Omega}^T \Omega + \Omega^T A(\tau, \zeta, \alpha, \beta) \Omega - b(\tau - \bar{\tau})^2 - 2\eta g^T (\Omega - \bar{\Omega})(\tau - \bar{\tau})\right),$$

where

- C : Constant for normalization
- $\bar{\Omega}$: Mean direction (equivalently $\bar{\phi}, \bar{\theta}$)
- $\bar{\tau}$: Mean delay
- κ : Concentration parameter in direction
- b : Concentration parameter in delay
- ζ : Ovalness parameter in direction
- η : Dependence between the spread in direction and in delay
- α, β : Two angles describing how the pdf is tilted on S_2
- g : Vector jointly determined by $\bar{\Omega}$ and β

3-dB spread surface of $f_{\text{ME}}(\Omega, \tau)$ with parameter setting

$\bar{\theta}$ [°]	$\bar{\phi}$ [°]	$\bar{\tau}$ [ms]	κ	ζ
0	0	58	100	0.01
α [°]	β [°]	η	b	
60	270	1×10^3	4×10^4	



Experimental Result I: Estimate of Direction (Azimuth-Elevation) Dispersion

Signal Model for SIMO Channel Sounding (Narrowband)

Received signal in a narrowband SIMO scenario

$$\mathbf{Y}(t) = \int_{-\pi}^{+\pi} \int_{-\pi/2}^{+\pi/2} \mathbf{c}(\phi, \theta) s(t) h(t; \phi, \theta) d\phi d\theta + \mathbf{W}(t),$$

where

- $\mathbf{Y}(t) \in \mathbb{C}^{M \times 1}$: output signals of the Rx array
- $\mathbf{c}(\phi, \theta) \in \mathbb{C}^{M \times 1}$: Rx antenna array response
- $s(t)$: complex envelope of the transmitted signal
- $h(t; \phi, \theta) \in \mathbb{C}^{1 \times 1}$: (time-variant) AoA-EoA spread function of the propagation channel.
- $\mathbf{W}(t) \in \mathbb{C}^{M \times 1}$: complex white Gaussian noise.

Signal Model for SIMO Channel Sounding (Narrowband)

In a scenario with D path components:

$$h(t; \phi, \theta) = \sum_{d=1}^D h_d(t; \phi, \theta).$$

The azimuth-elevation power spectrum of the radio channel under the WSS uncorrelated scattering assumption reads

$$P(\phi, \theta) \doteq \mathbb{E}[|h(t; \phi, \theta)|^2] = \sum_{d=1}^D P_d(\phi, \theta),$$

where $P_d(\phi, \theta)$ is the power spectrum of the d th path component, i.e.

$$P_d(\phi, \theta) \doteq \mathbb{E}[|h_d(t; \phi, \theta)|^2] = P_d \cdot f_d(\phi, \theta)$$

with P_d being the average power, and $f_d(\phi, \theta)$ the spectral density of the path component.

Signal Model for SIMO Channel Sounding (Narrowband)

We assume

$$f_d(\phi, \theta) = f(\phi, \theta; \boldsymbol{\theta}_d)$$

with

- $f(\phi, \theta; \boldsymbol{\theta}_d)$: the Fisher-Bingham-5 density function
- $\boldsymbol{\theta}_d \doteq [\bar{\phi}_d, \bar{\theta}_d, \kappa_d, \beta_d, \alpha_d]$: parameter vector
of the d th path component.

Parameter vector for the D -path-component scenario:

$$\boldsymbol{\Theta} \doteq [P_d, \boldsymbol{\theta}_d; d = 1, \dots, D].$$

The SAGE algorithm can be used as an approximation of the maximum-likelihood estimate $\hat{\boldsymbol{\Theta}}$ (See the paper for the details of the algorithm.)

Experimental Investigations

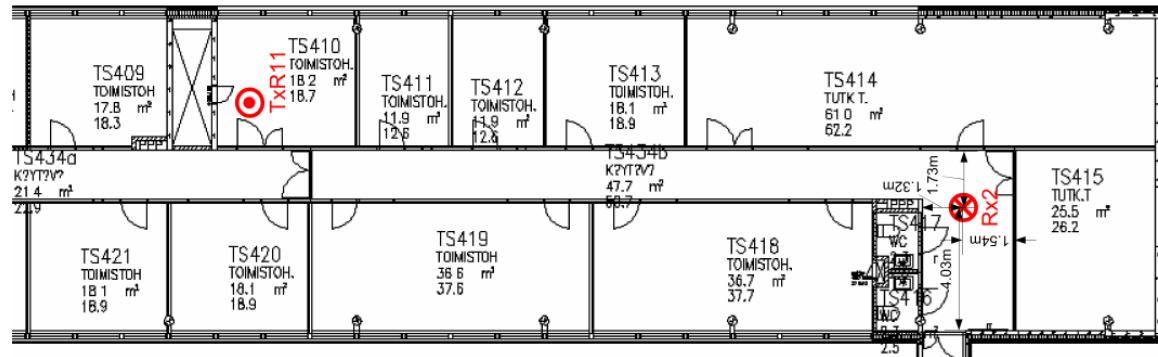
(a) Surroundings of the Tx.



(b) Surroundings of the Rx.

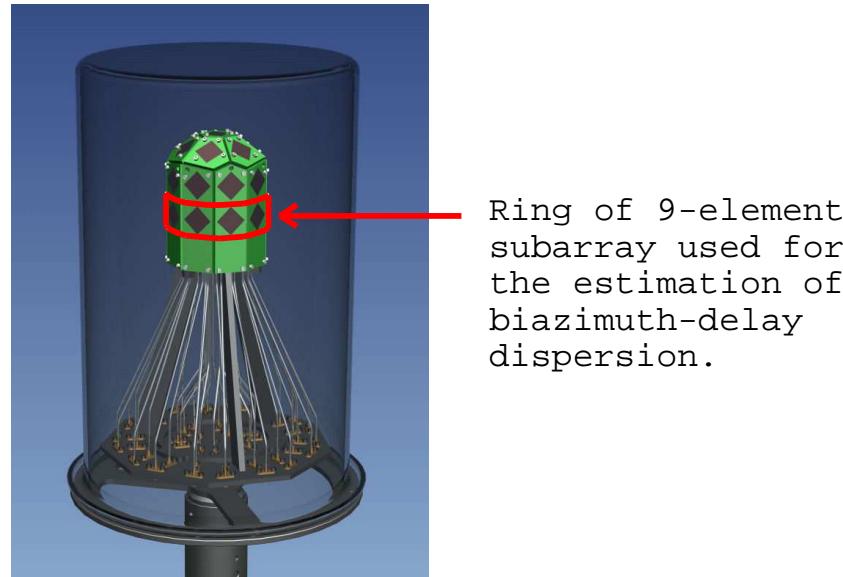


(c) Map of the premises.

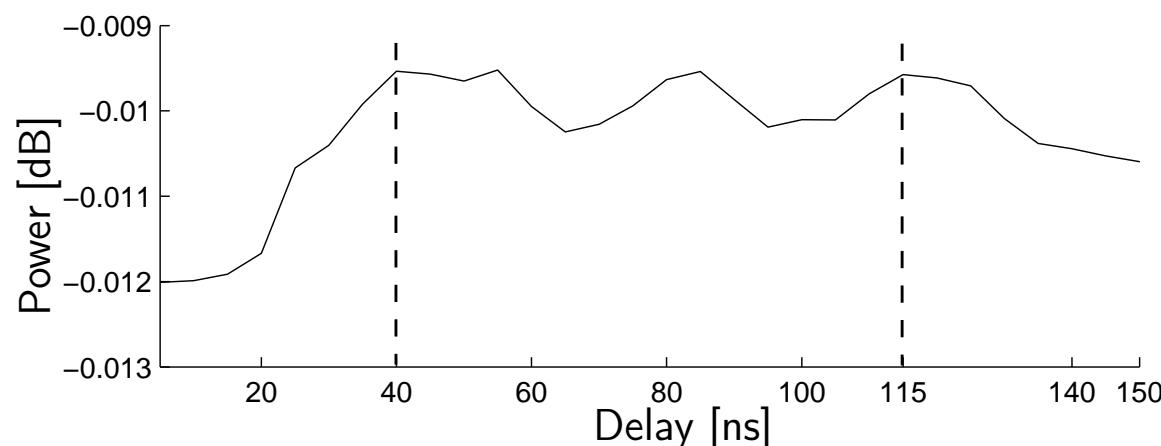


Experimental Investigations (Cont.)

- 1×9 SIMO system with Rx array consisting of dual-polarized antennas in the lower ring of the array



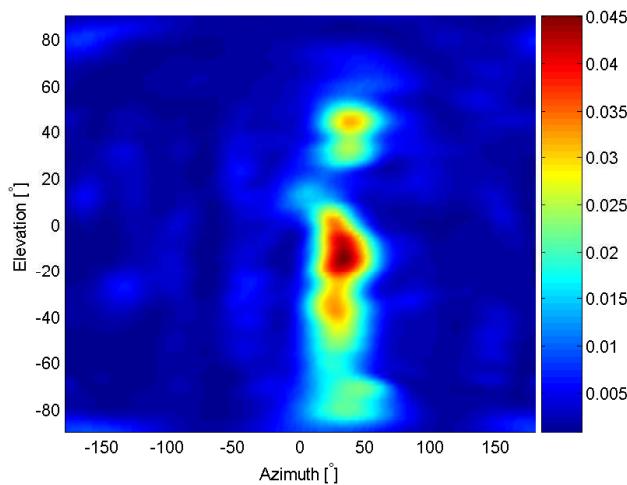
- Estimated delay power spectrum



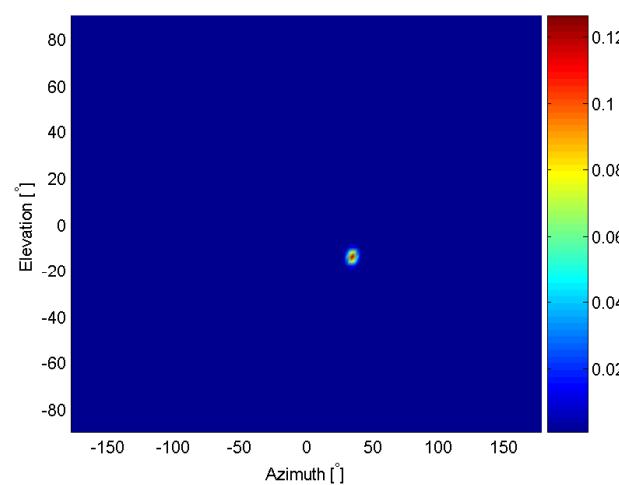
Estimate of Azimuth-Elevation Power Spectrum

$\tau = 40 \text{ ns}$

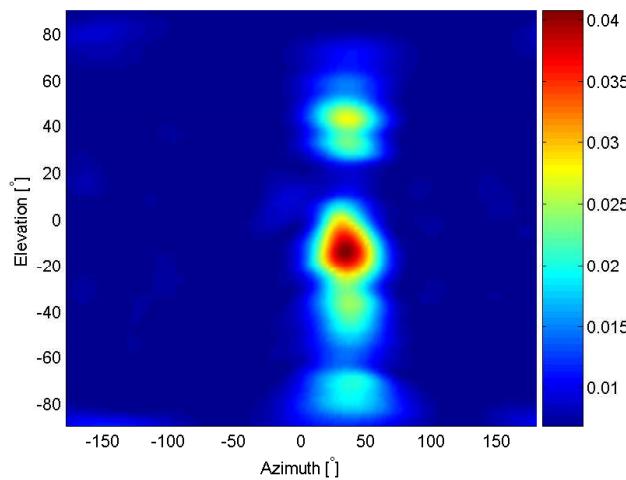
Bartlett($\hat{\Sigma}$)



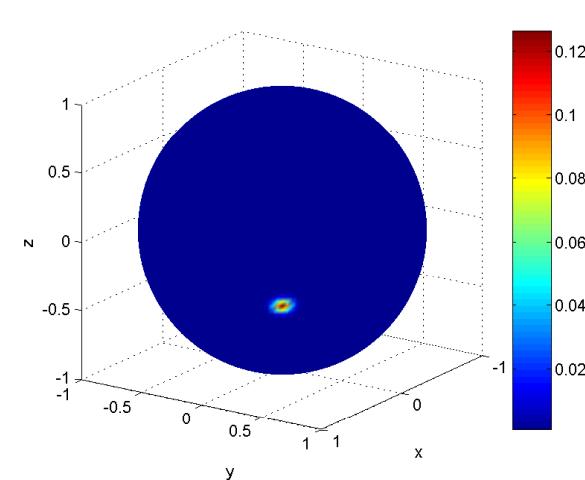
$\hat{P}(\phi, \theta)$



Bartlett($\hat{\Sigma}(\hat{\Theta})$)



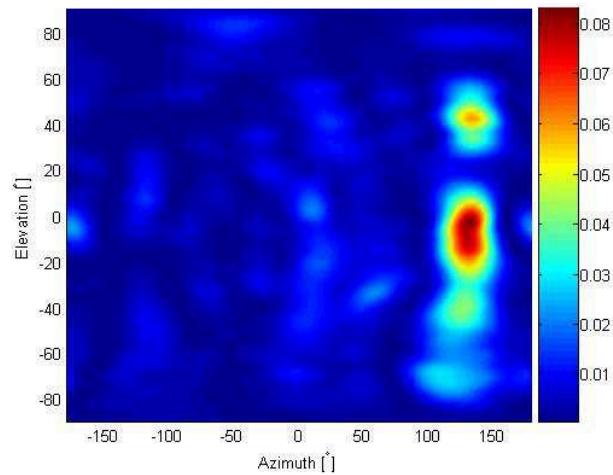
$\hat{P}(\Omega)$



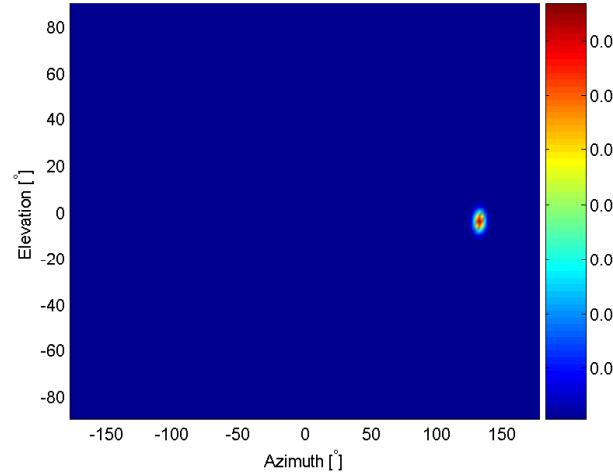
Estimate of Azimuth-Elevation Power Spectrum

$$\tau = 115 \text{ ns}$$

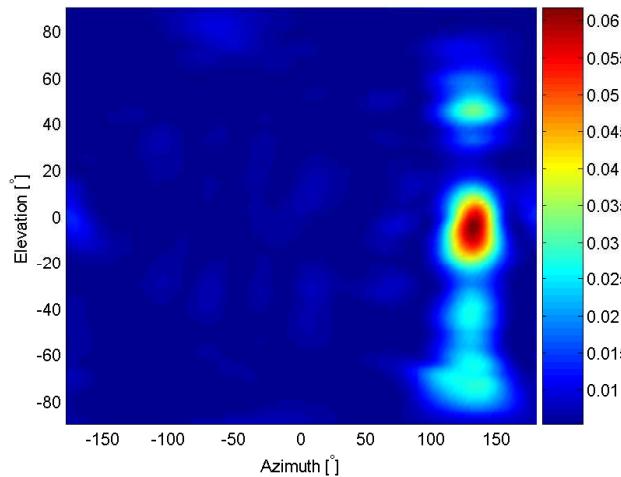
$$\text{Bartlett}(\hat{\Sigma})$$



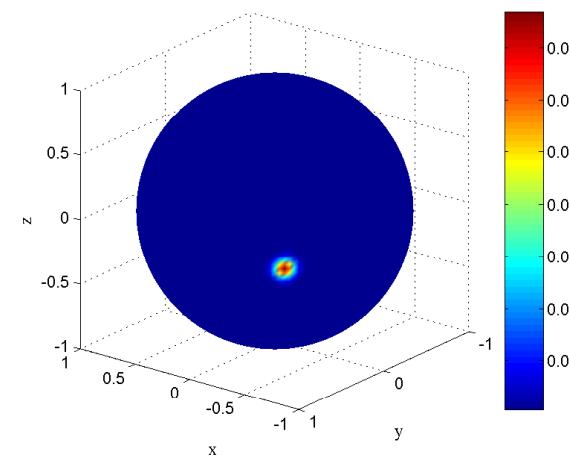
$$\hat{P}(\phi, \theta)$$



$$\text{Bartlett}(\Sigma(\hat{\Theta}))$$

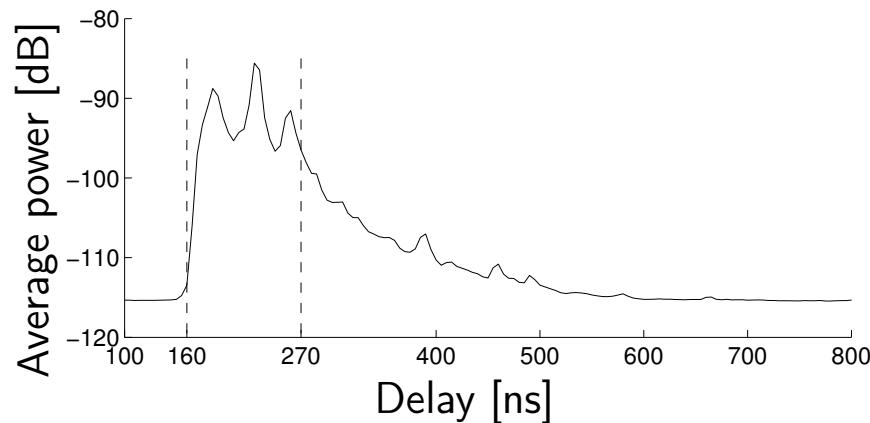


$$\hat{P}(\Omega)$$



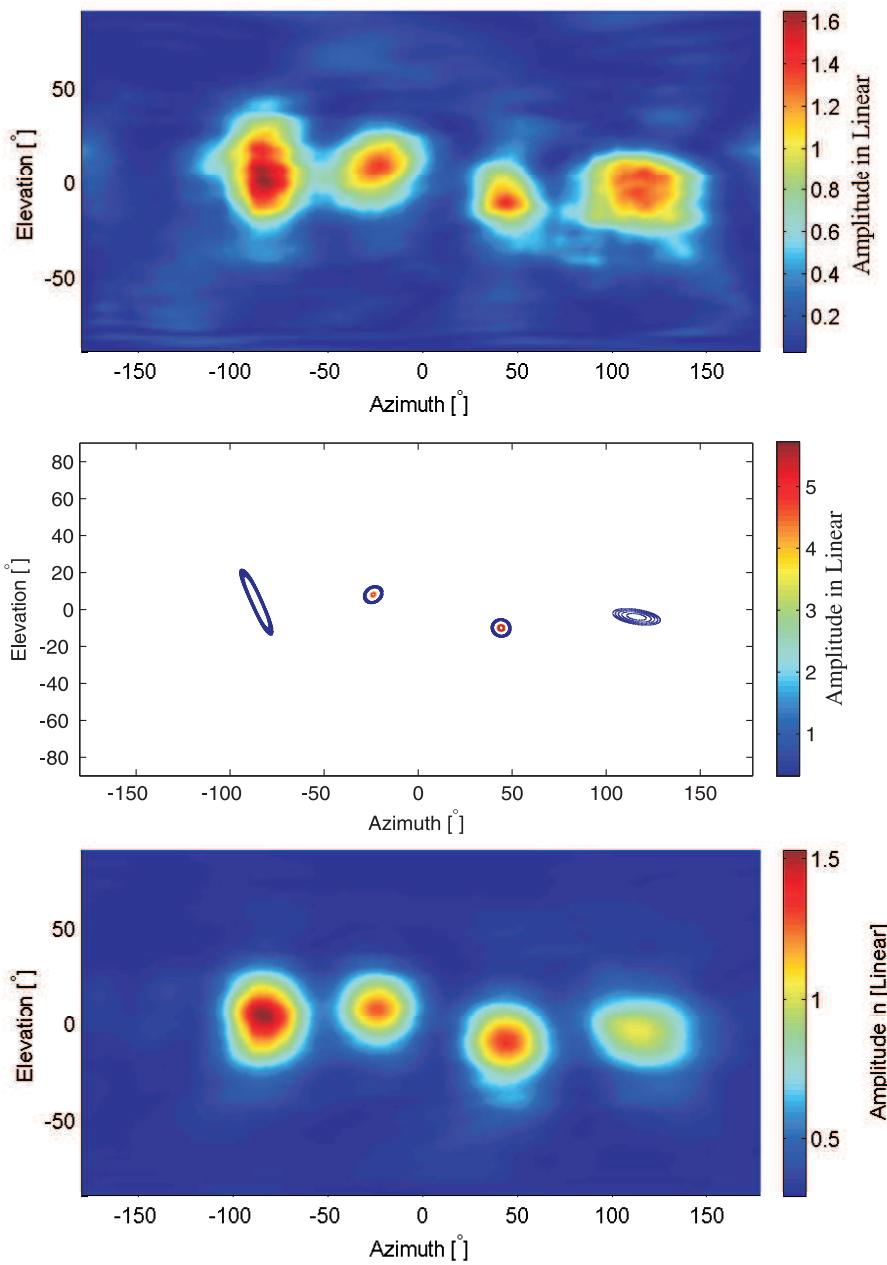
Experimental Investigations (Cont.)

- 50×1 MISO system
- Estimated delay power spectrum



Estimate of Azimuth-Elevation Power Spectrum

- Bartlett spectrum of the signal received at delay 160 ns
- Power spectrum estimate using the proposed characterization method
- Bartlett spectrum of the reconstructed signal



Experimental Result II: Estimate of Direction-Delay Dispersion

Signal Model for SIMO Channel Sounding (Wideband)

Received signal in a $1 \times M$ SIMO scenario

$$\mathbf{Y}(t) = \int_{-\infty}^{+\infty} \int_{\mathbb{S}_2} \mathbf{c}(\Omega) u(t - \tau) H(\Omega, \tau) d\Omega d\tau + \mathbf{W}(t),$$

where

- $\mathbf{Y}(t) \in \mathbb{C}^{M \times 1}$: output signals of the Rx array
- $\mathbf{c}(\Omega) \in \mathbb{C}^{M \times 1}$: Rx antenna array response
- $u(t)$: complex envelope of the transmitted signal
- $H(\Omega, \tau)$: direction-delay spread function of the propagation channel
- $\mathbf{W}(t) \in \mathbb{C}^{M \times 1}$: complex white Gaussian noise.

Signal Model for SIMO Channel Sounding (Wideband)

In a scenario with D components

$$H(\Omega, \tau) = \sum_{d=1}^D H_d(\Omega, \tau).$$

The component spread function $H_d(\Omega, \tau)$: $d \in \{1, \dots, D\}$ satisfy

$$\mathbb{E}[H_d(\Omega, \tau)^* H_{d'}(\Omega', \tau')] = P_d(\Omega, \tau) \delta_{dd'} \delta(\Omega - \Omega') \delta(\tau - \tau').$$

Thus,

$$\mathbb{E}[H(\Omega, \tau)^* H(\Omega', \tau')] = P(\Omega, \tau) \delta(\Omega - \Omega') \delta(\tau - \tau')$$

and

$$P(\Omega, \tau) = \sum_{d=1}^D P_d(\Omega, \tau).$$

Signal Model for SIMO Channel Sounding (Wideband)

The component power spectrum $P_d(\Omega, \tau)$ is written as

$$P_d(\Omega, \tau) = P_d \cdot f_d(\Omega, \tau).$$

We assume

$$f_d(\Omega, \tau) = f(\Omega, \tau; \boldsymbol{\theta}_d),$$

- $f(\Omega, \tau; \boldsymbol{\theta}_d)$: the derived entropy-maximizing direction-delay pdf
- $\boldsymbol{\theta}_d \doteq [\bar{\phi}_d, \bar{\theta}_d, \bar{\tau}_d, \kappa_d, \zeta_d, \alpha_d, \beta_d, \eta_d, b_d]$: component-specific parameters.

Parameter vector for the D -component scenario:

$$\boldsymbol{\Theta} \doteq [P_d, \boldsymbol{\theta}_d; d = 1, \dots, D].$$

The SAGE algorithm can be used as an approximation of the maximum-likelihood estimate of $\boldsymbol{\Theta}$.

Experimental Investigations

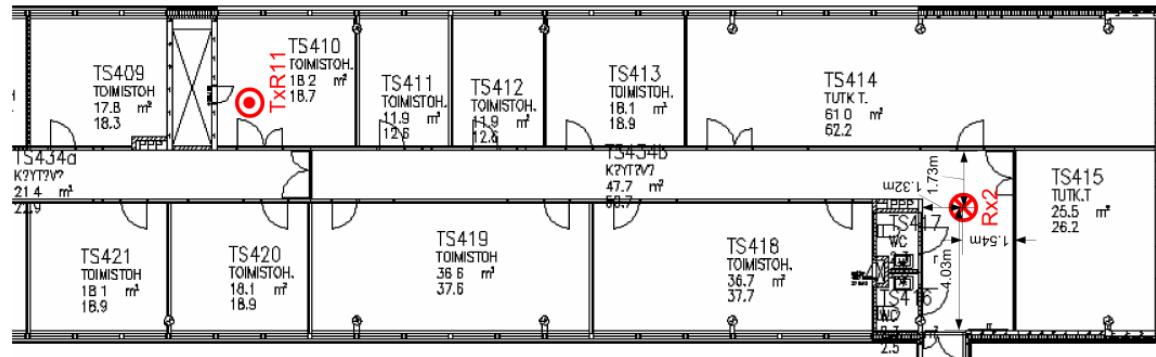
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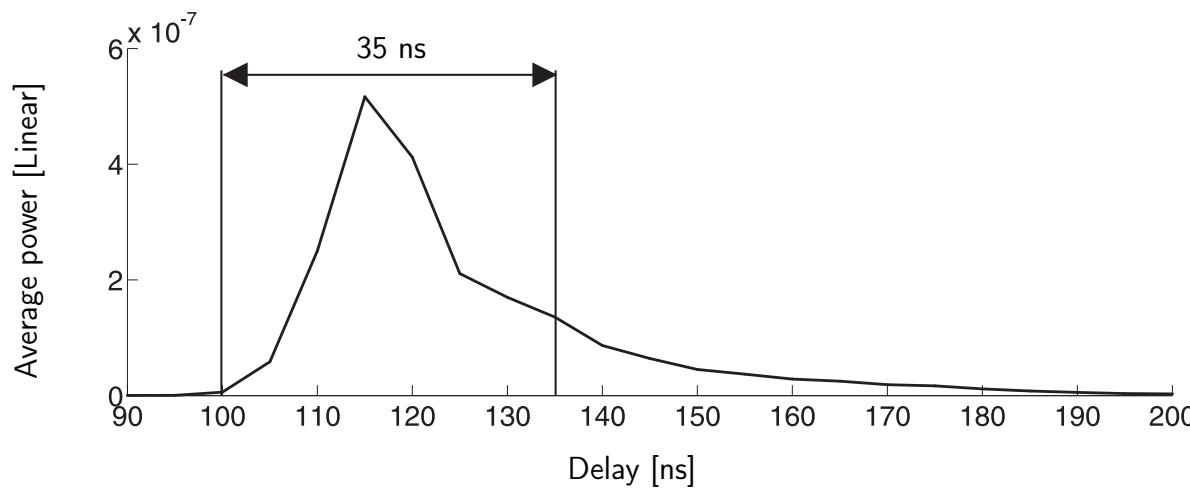


(c) Map of the premises.



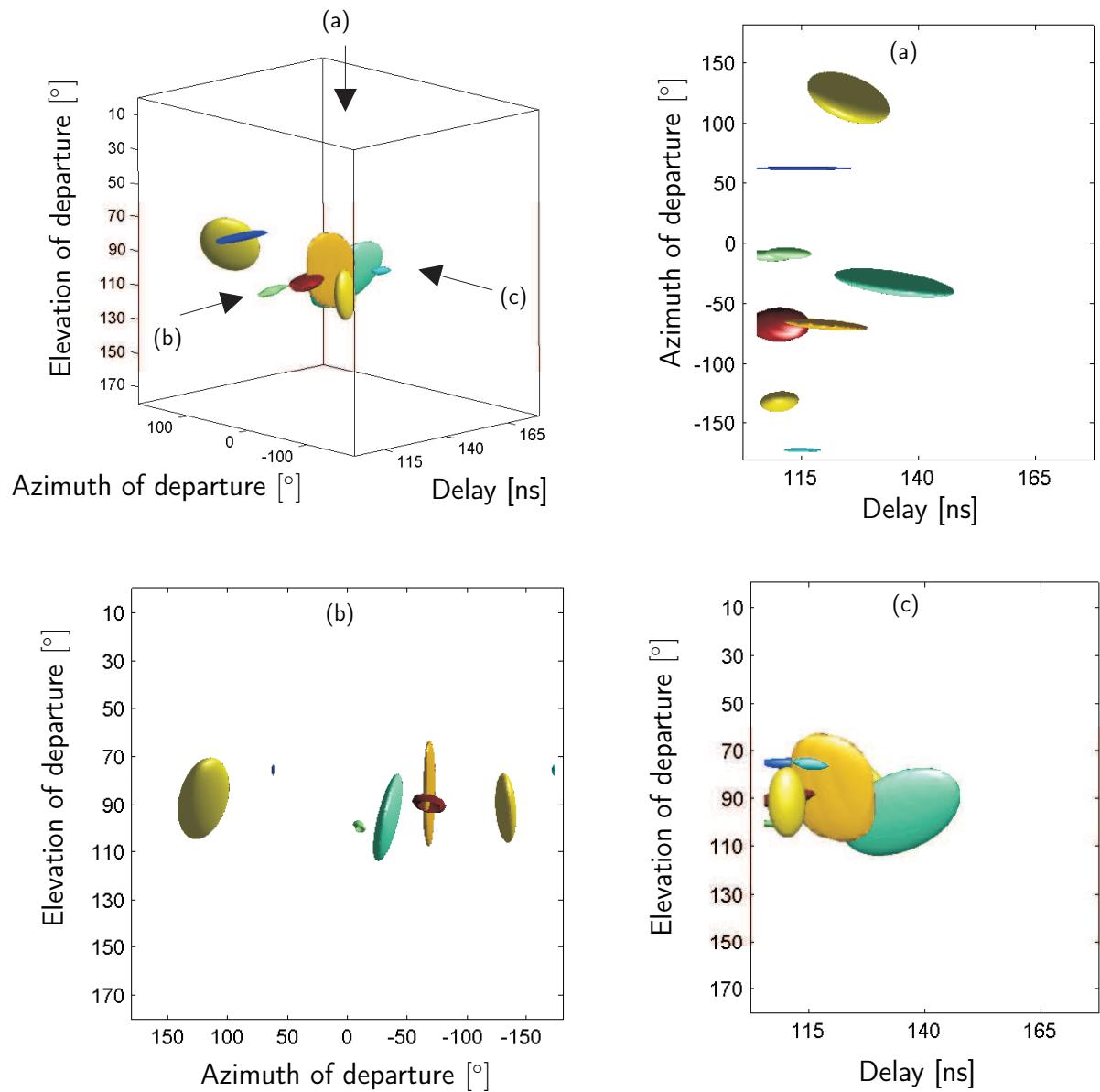
Experimental Investigations

- 50×1 Multiple-Input-Single-Output (MISO) system.
- Delay range considered: [100 ns, 135 ns].
Average power delay profile:



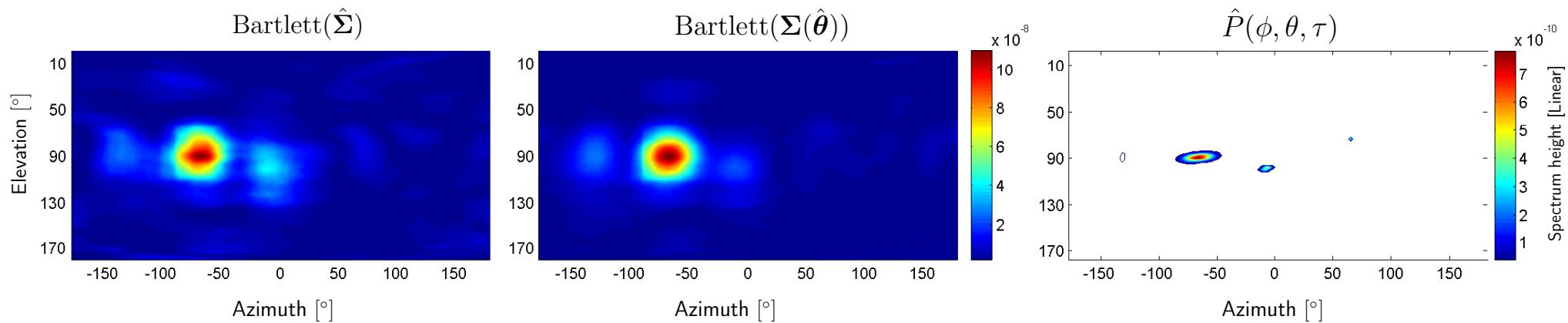
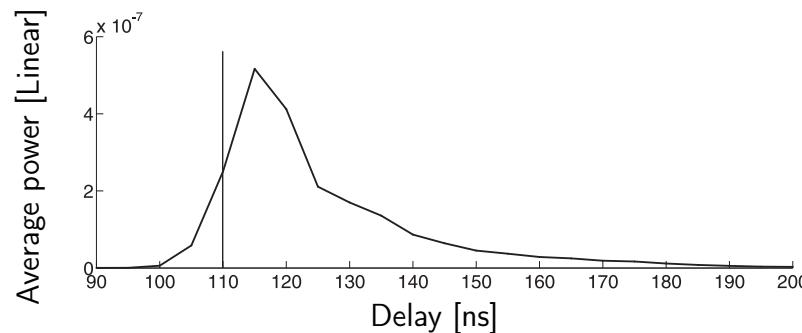
- Number of components: 10, equal to the observed dominant local maxima of the direction-delay Bartlett spectrum.

3 dB-Spread Surfaces of Estimated Component Azimuth–Elevation–Delay Power Spectra



Estimates of Azimuth–Elevation–Delay Spectrum

Delay $\tau = 110$ ns:

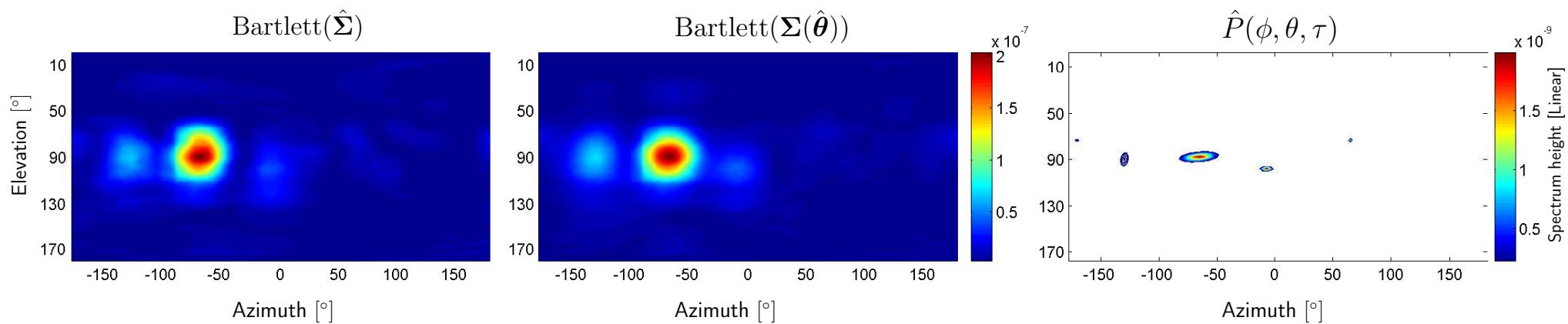
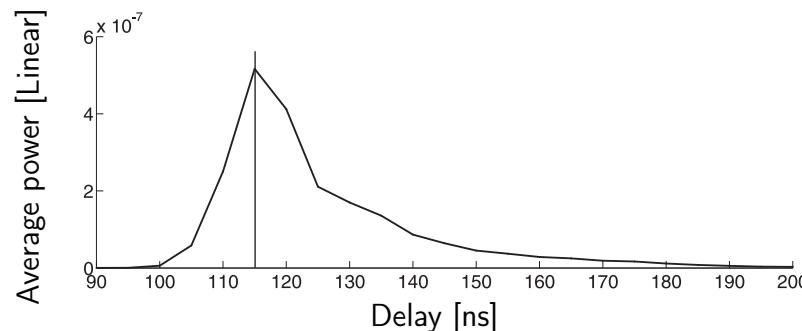


- Bartlett(Σ) : Power spectrum estimate calculated using the Bartlett beamformer
- $\hat{\Sigma}$: Sample covariance matrix
- $\Sigma(\hat{\theta})$: Covariance matrix computed based on $\hat{P}(\Omega, \tau)$
- $\hat{P}(\phi, \theta, \tau)$: Azimuth–elevation–delay power spectrum estimate

$$\hat{P}(\phi, \theta, \tau) = \sin(\theta) \cdot \hat{P}(\Omega, \tau) |_{\Omega=e(\phi, \theta)}$$

Estimates of Azimuth–Elevation–Delay Spectrum

Delay $\tau = 115$ ns:

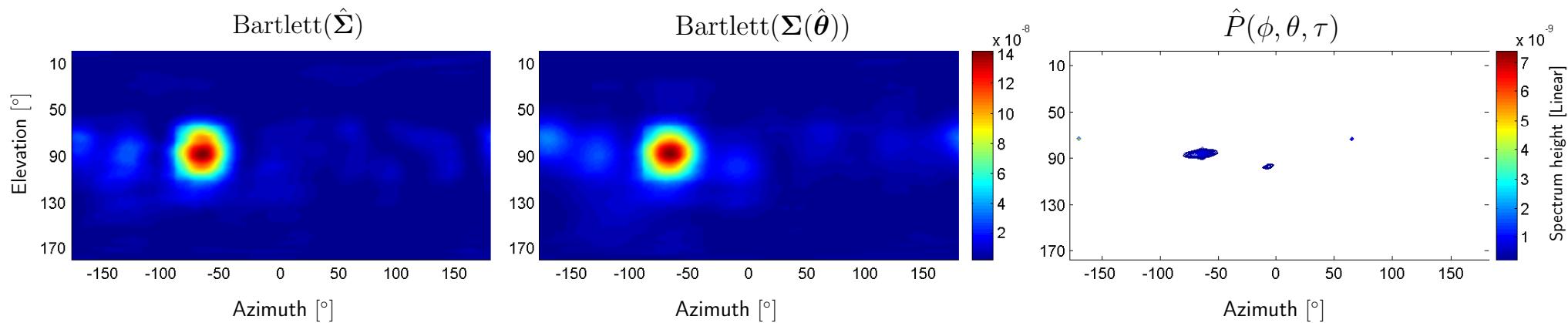
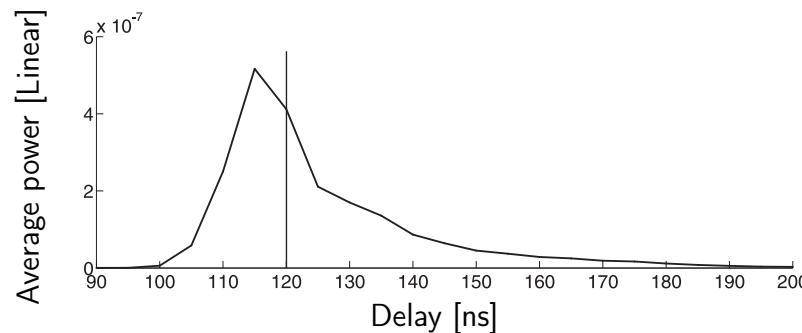


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- $\hat{P}(\phi, \theta, \tau)$: Azimuth–elevation–delay power spectrum estimate

$$\hat{P}(\phi, \theta, \tau) = \sin(\theta) \cdot \hat{P}(\Omega, \tau)|_{\Omega=e(\phi, \theta)}$$

Estimates of Azimuth–Elevation–Delay Spectrum

Delay $\tau = 120$ ns:

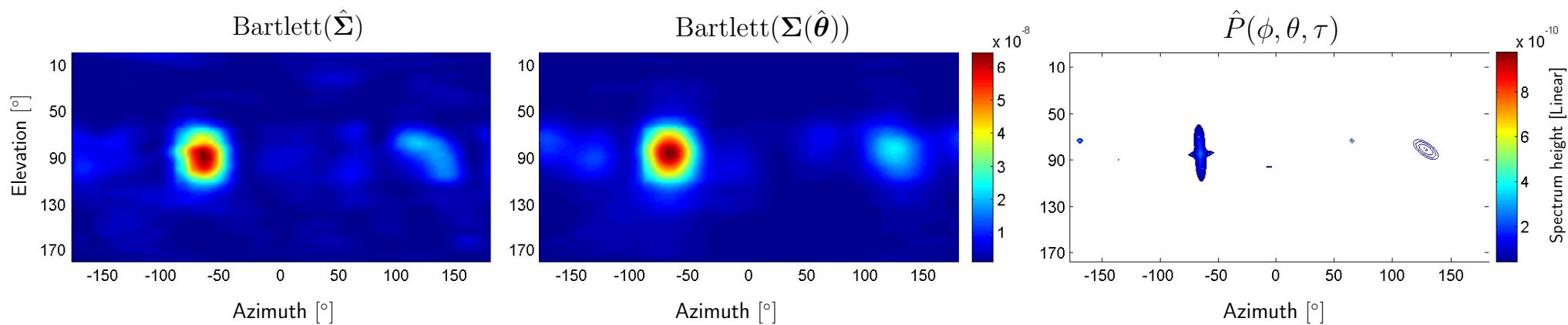
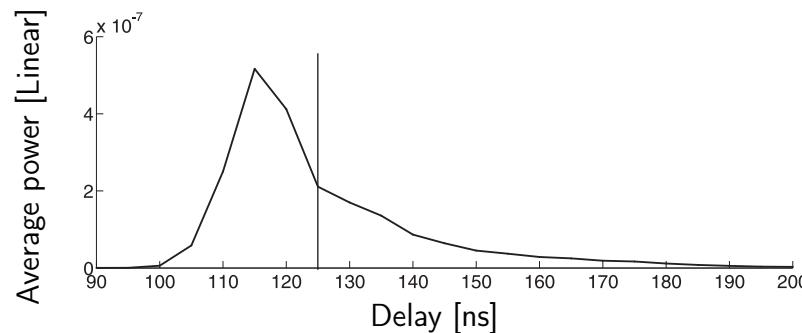


- Bartlett(Σ) : Power spectrum estimate calculated using the Bartlett beamformer
- $\hat{\Sigma}$: Sample covariance matrix
- $\Sigma(\hat{\theta})$: Covariance matrix computed based on $\hat{P}(\Omega, \tau)$
- $\hat{P}(\phi, \theta, \tau)$: Azimuth–elevation–delay power spectrum estimate

$$\hat{P}(\phi, \theta, \tau) = \sin(\theta) \cdot \hat{P}(\Omega, \tau) |_{\Omega=e(\phi, \theta)}$$

Estimates of Azimuth–Elevation–Delay Spectrum

Delay $\tau = 125$ ns:

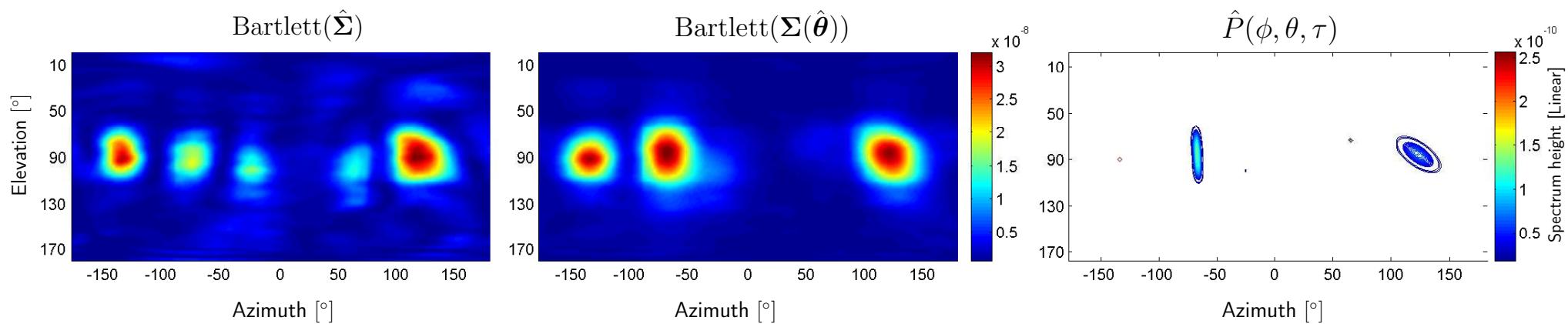
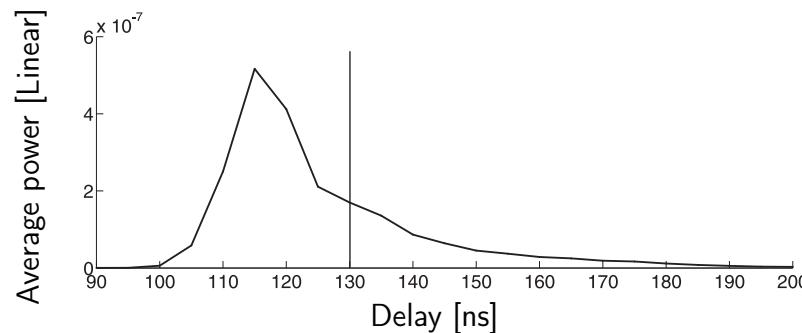


- Bartlett(Σ) : Power spectrum estimate calculated using the Bartlett beamformer
- $\hat{\Sigma}$: Sample covariance matrix
- $\Sigma(\hat{\theta})$: Covariance matrix computed based on $\hat{P}(\Omega, \tau)$
- $\hat{P}(\phi, \theta, \tau)$: Azimuth–elevation–delay power spectrum estimate

$$\hat{P}(\phi, \theta, \tau) = \sin(\theta) \cdot \hat{P}(\Omega, \tau)|_{\Omega=e(\phi, \theta)}$$

Estimates of Azimuth–Elevation–Delay Spectrum

Delay $\tau = 130$ ns:

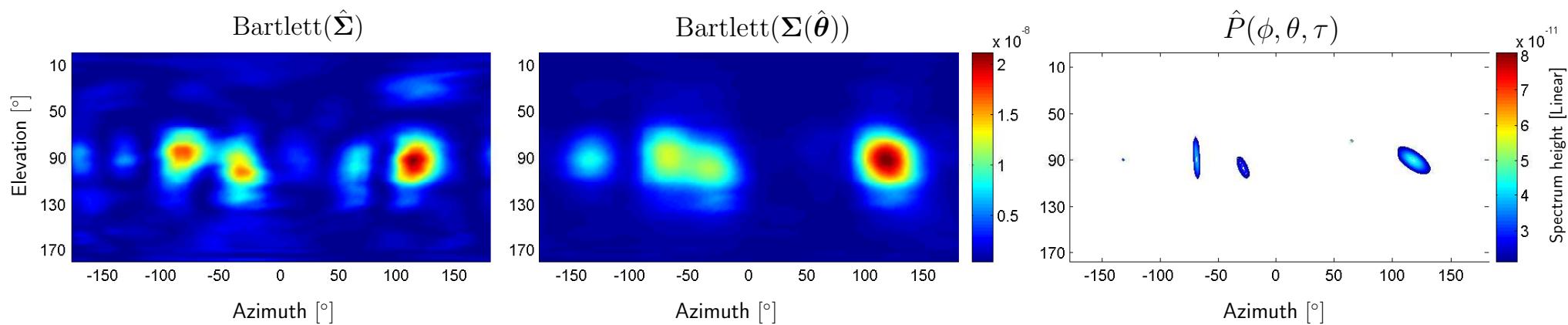
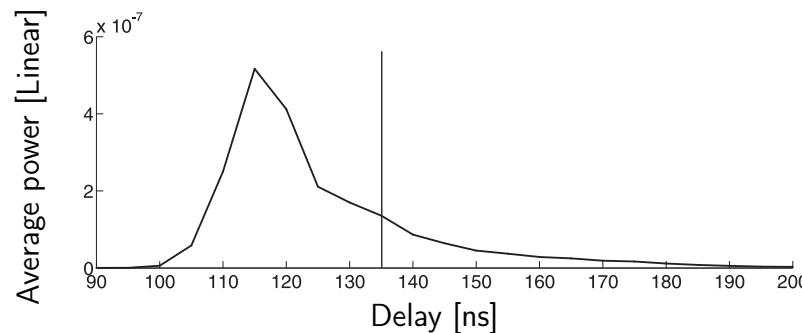


- Bartlett(Σ) : Power spectrum estimate calculated using the Bartlett beamformer
- $\hat{\Sigma}$: Sample covariance matrix
- $\Sigma(\hat{\theta})$: Covariance matrix computed based on $\hat{P}(\Omega, \tau)$
- $\hat{P}(\phi, \theta, \tau)$: Azimuth–elevation–delay power spectrum estimate

$$\hat{P}(\phi, \theta, \tau) = \sin(\theta) \cdot \hat{P}(\Omega, \tau)|_{\Omega=e(\phi, \theta)}$$

Estimates of Azimuth–Elevation–Delay Spectrum

Delay $\tau = 135$ ns:



- Bartlett(Σ) : Power spectrum estimate calculated using the Bartlett beamformer
- $\hat{\Sigma}$: Sample covariance matrix
- $\Sigma(\hat{\theta})$: Covariance matrix computed based on $\hat{P}(\Omega, \tau)$
- $\hat{P}(\phi, \theta, \tau)$: Azimuth–elevation–delay power spectrum estimate

$$\hat{P}(\phi, \theta, \tau) = \sin(\theta) \cdot \hat{P}(\Omega, \tau)|_{\Omega=e(\phi, \theta)}$$

Experimental Result III: Estimate of Biasimuth-Delay Dispersion

Experimental Investigations

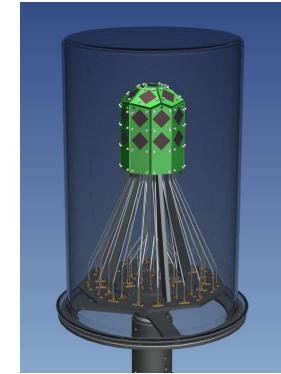
Tx environment



Rx environment



Antenna array

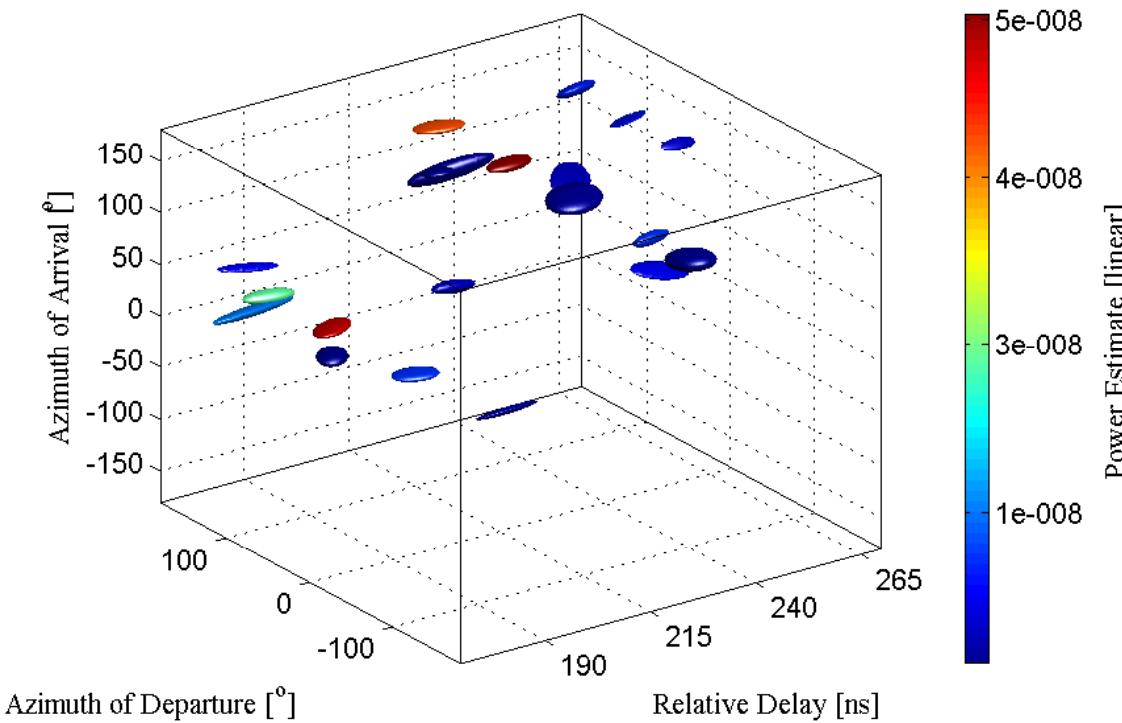


Measurement set-up:

- Carrier frequency: 5.2 GHz
- Bandwidth: 200 MHz
- 50-element Tx and Rx Patch arrays with $\pm 45^\circ$ slanted antennas
- Tx array height: 1.53 m; Rx array height: 0.82 m
- Big hall, time-variant
- 900 observations samples (60 s)

Estimate of Biasimuth-Delay Power Spectrum

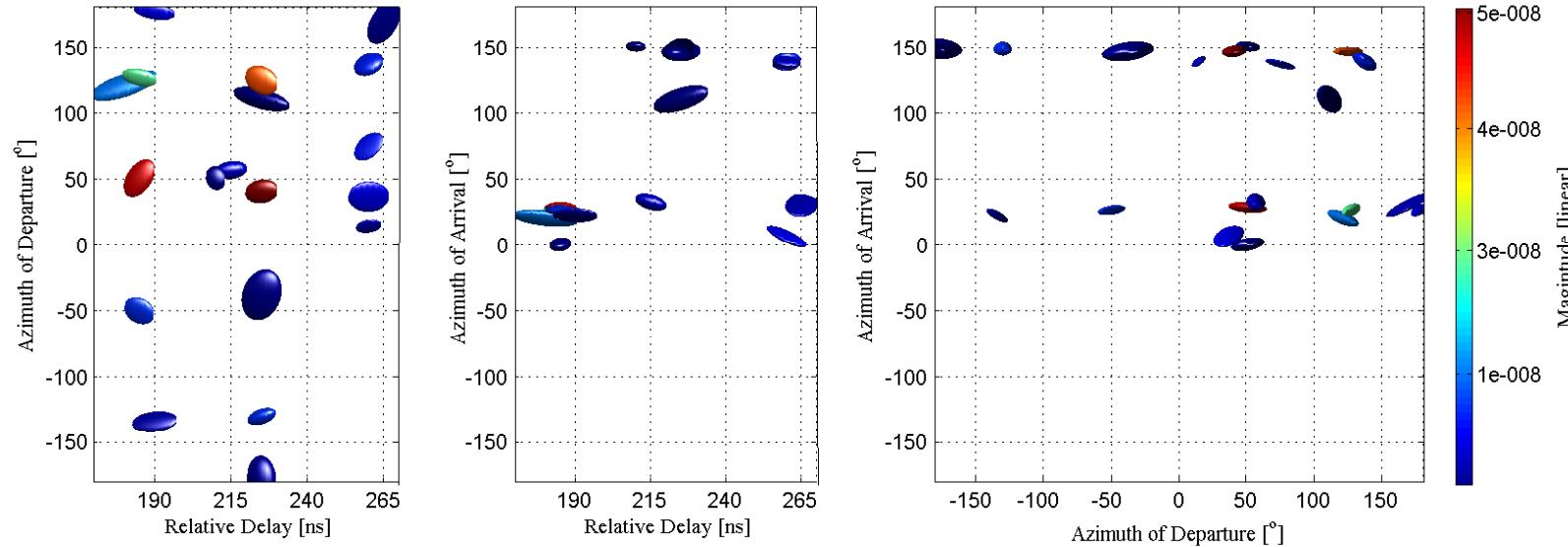
Estimated 3dB-spread surfaces of individual dispersed path components:



The color of the surfaces codes the path power estimates according to the color scale reported on the right.

Estimate of Biasimuth-Delay Power Spectrum

Estimated 3dB-spread surfaces of individual dispersed path components:



Experimental Investigations

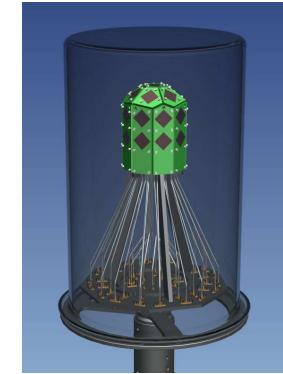
Tx environment



Rx environment



Antenna array

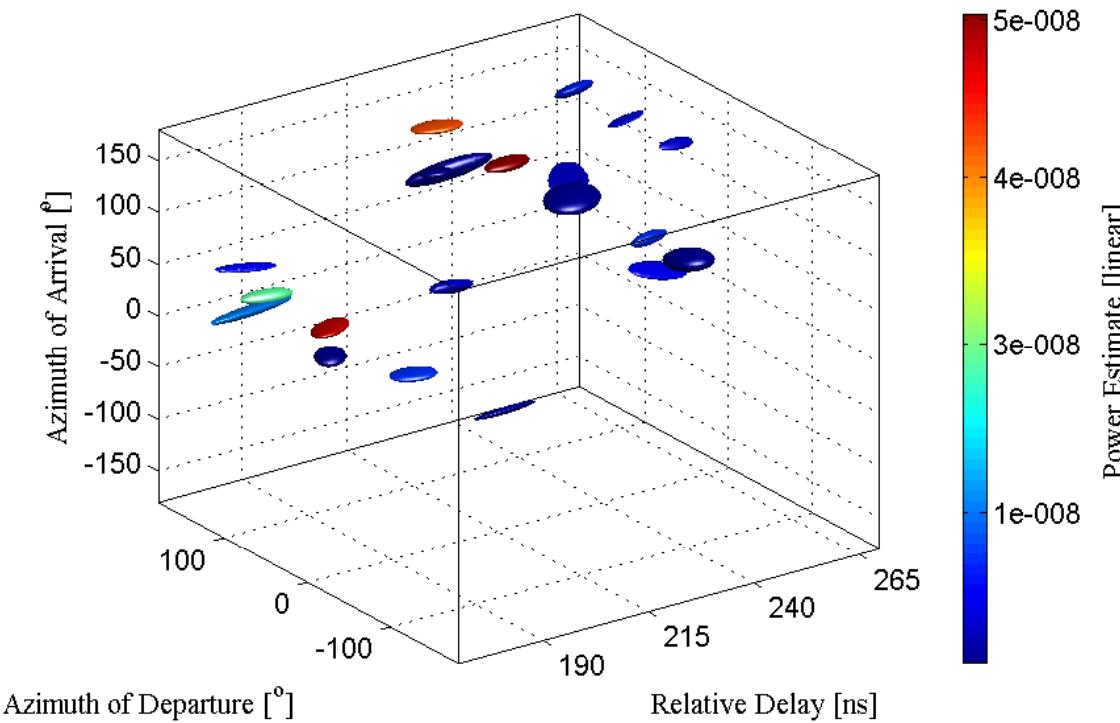


Measurement set-up:

- Carrier frequency: 5.2 GHz
- Bandwidth: 200 MHz
- 50-element Tx and Rx Patch arrays with $\pm 45^\circ$ slanted antennas
- Tx array height: 1.53 m; Rx array height: 0.82 m
- Big hall, time-variant
- 900 observations samples (60 s)

Estimate of Biasimuth-Delay Power Spectrum

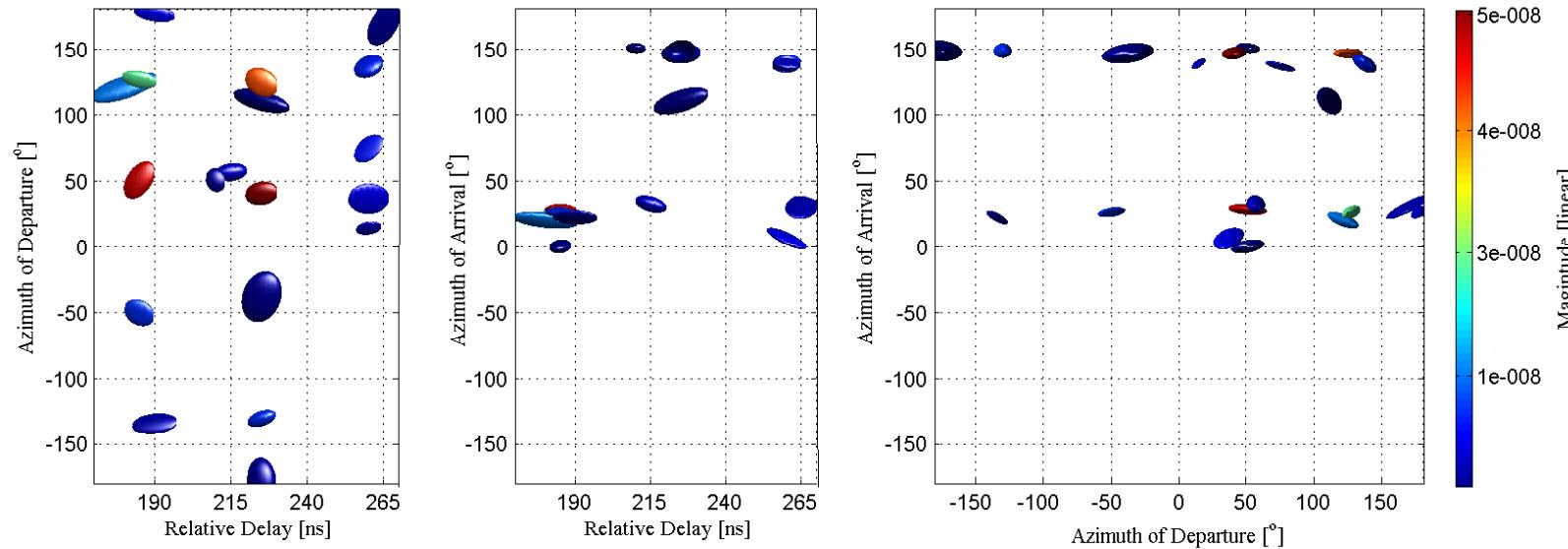
Estimated 3dB-spread surfaces of individual dispersed path components:



The color of the surfaces codes the path power estimates according to the color scale reported on the right.

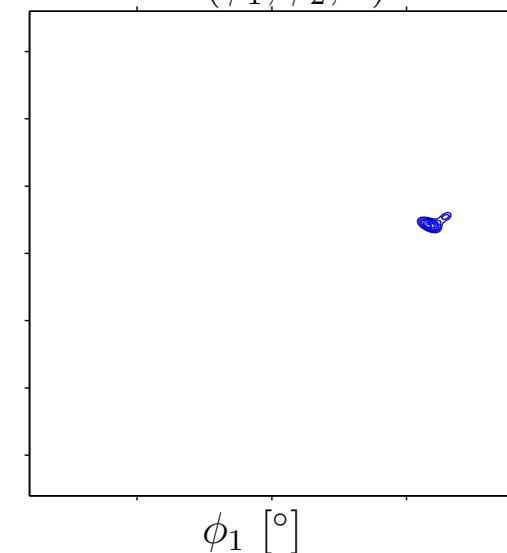
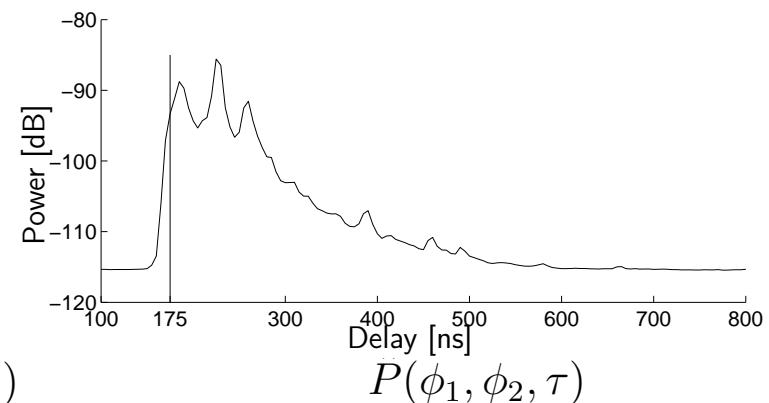
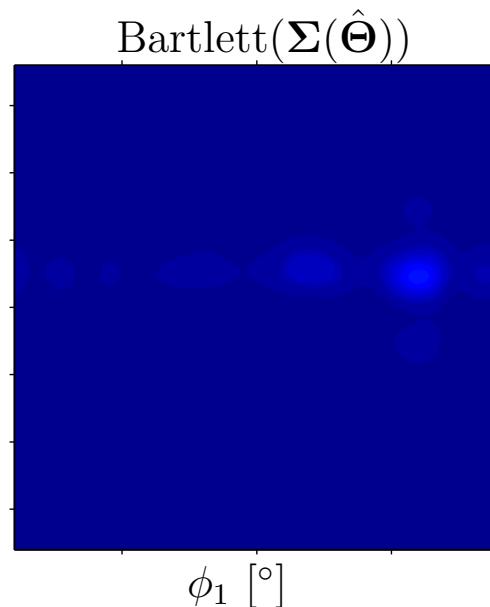
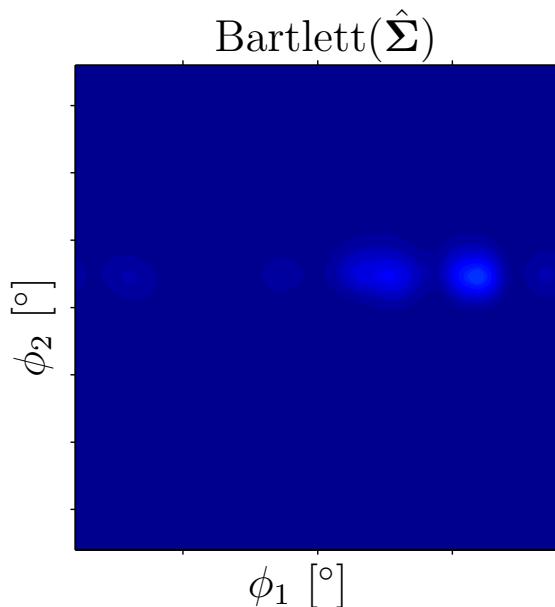
Estimate of Biasimuth-Delay Power Spectrum

Estimated 3dB-spread surfaces of individual dispersed path components:



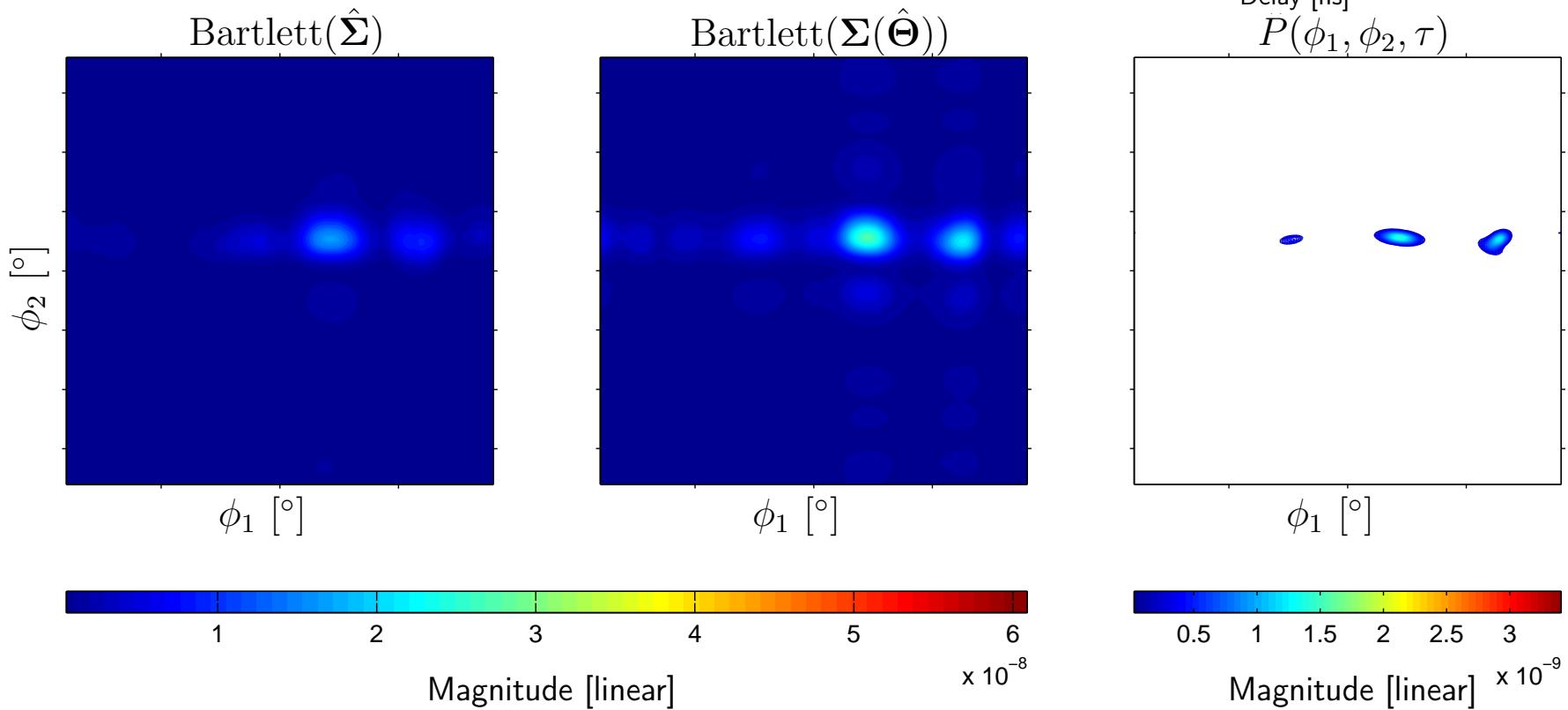
Estimate of biazimuth-delay power spectrum

$\tau = 175 \text{ ns}$



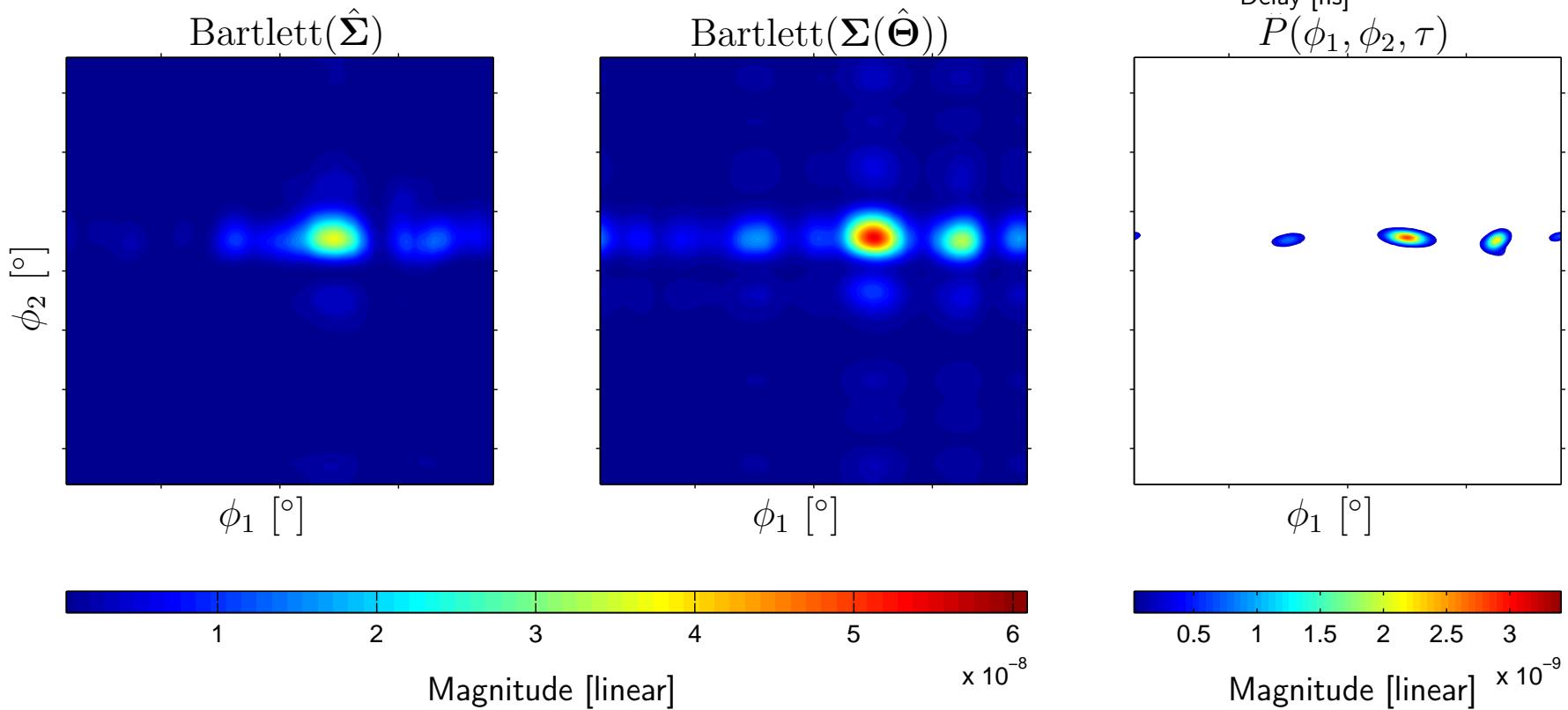
Estimate of biazimuth-delay power spectrum

$\tau = 180 \text{ ns}$



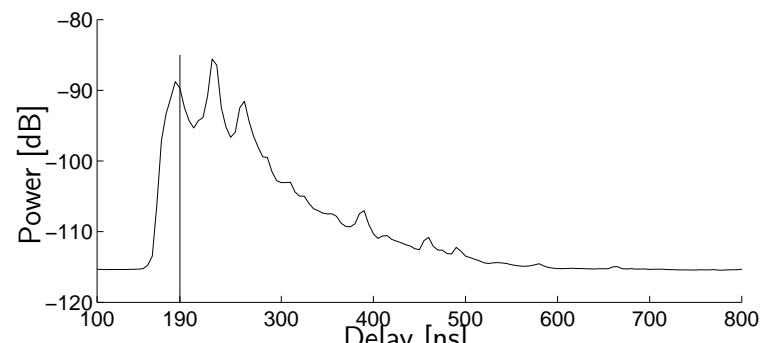
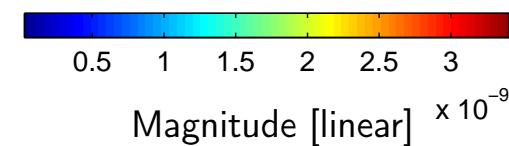
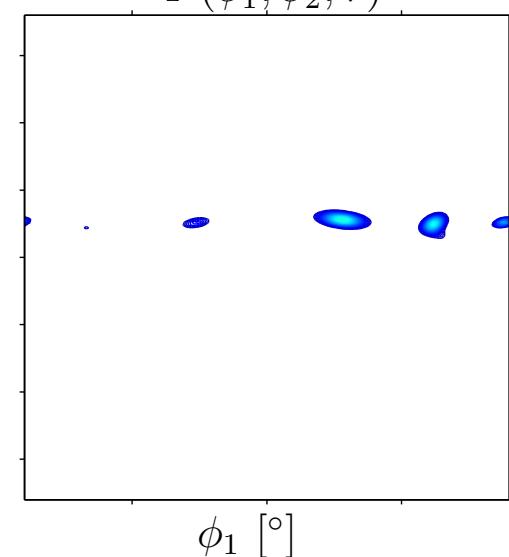
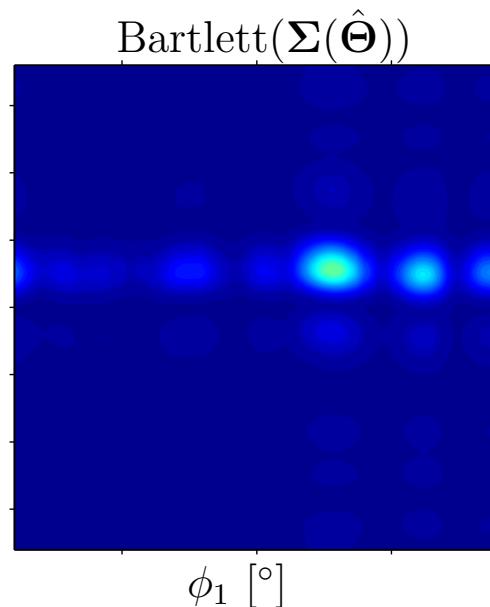
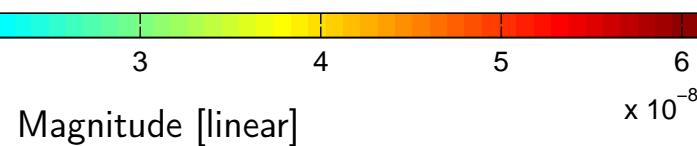
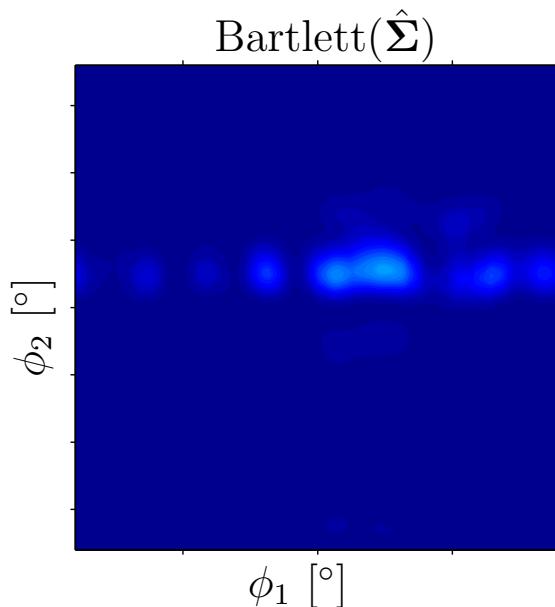
Estimate of biazimuth-delay power spectrum

$$\tau = 185 \text{ ns}$$



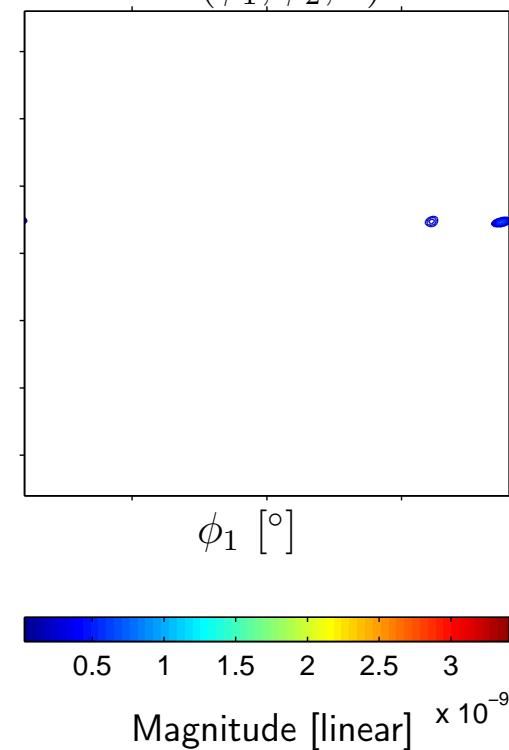
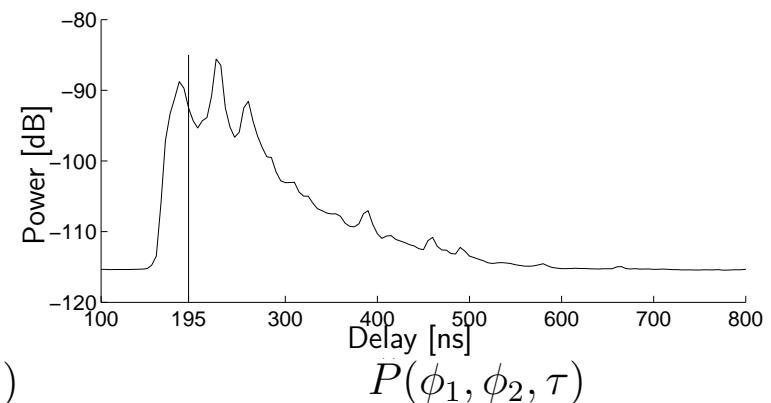
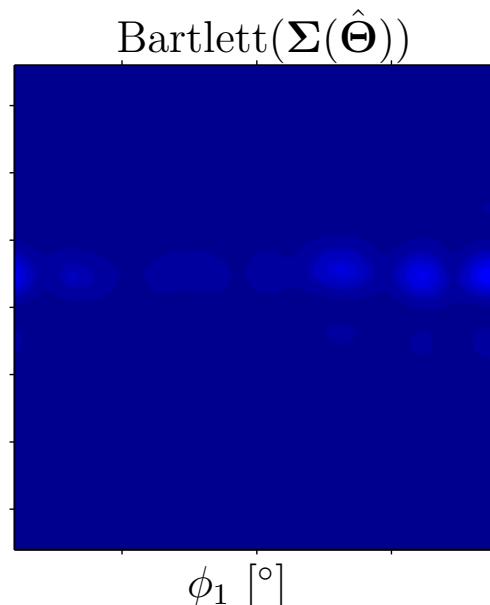
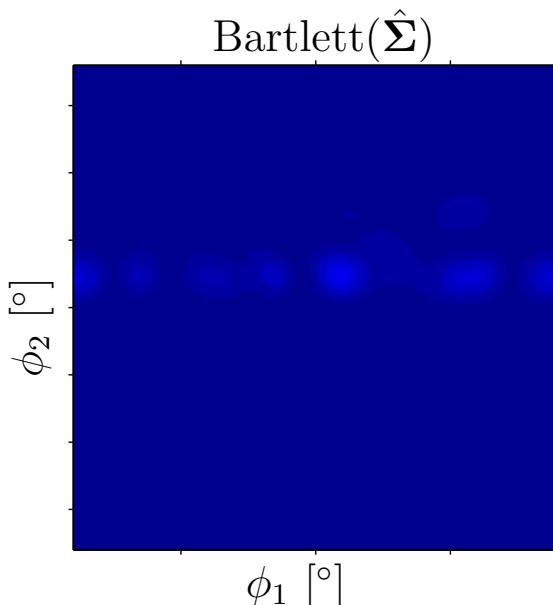
Estimate of biazimuth-delay power spectrum

$\tau = 190 \text{ ns}$



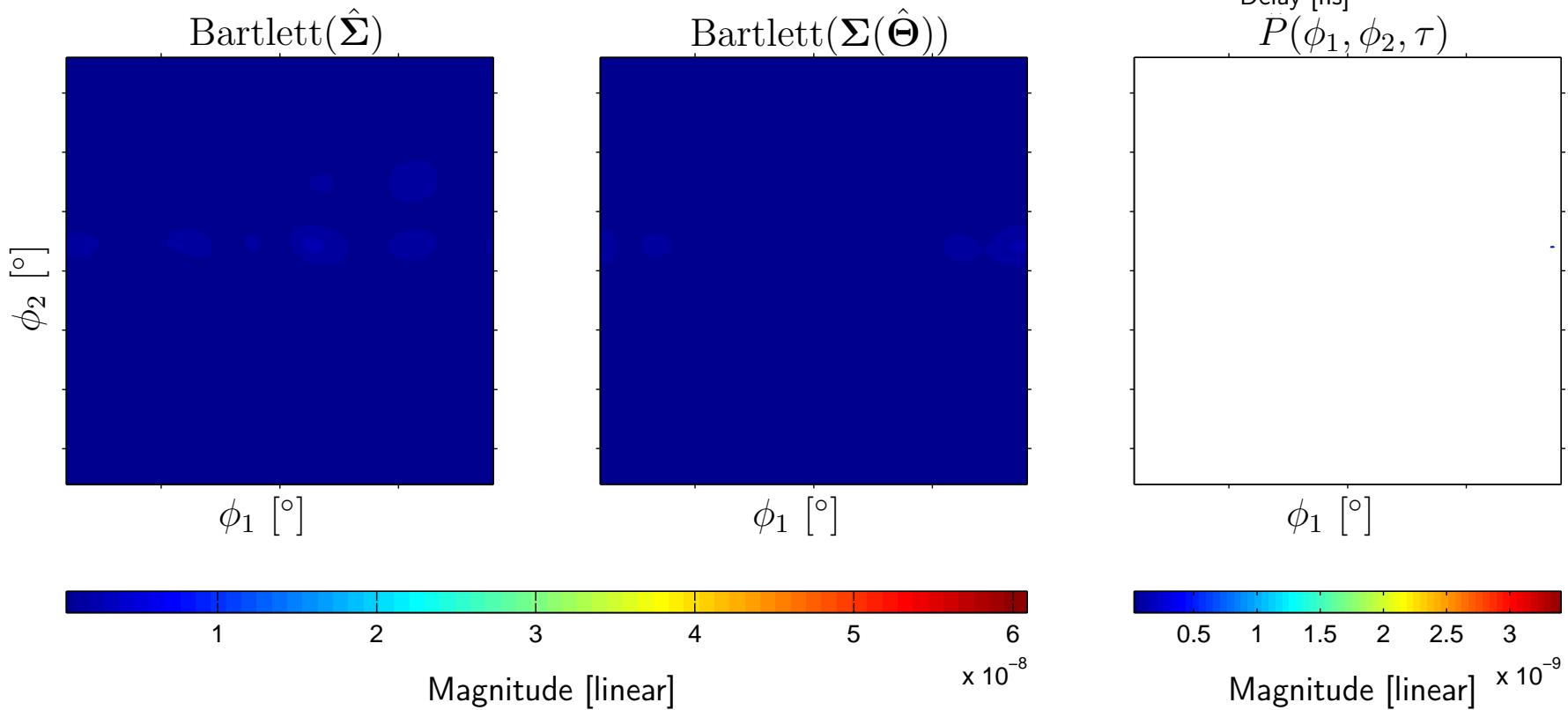
Estimate of biazimuth-delay power spectrum

$\tau = 195 \text{ ns}$



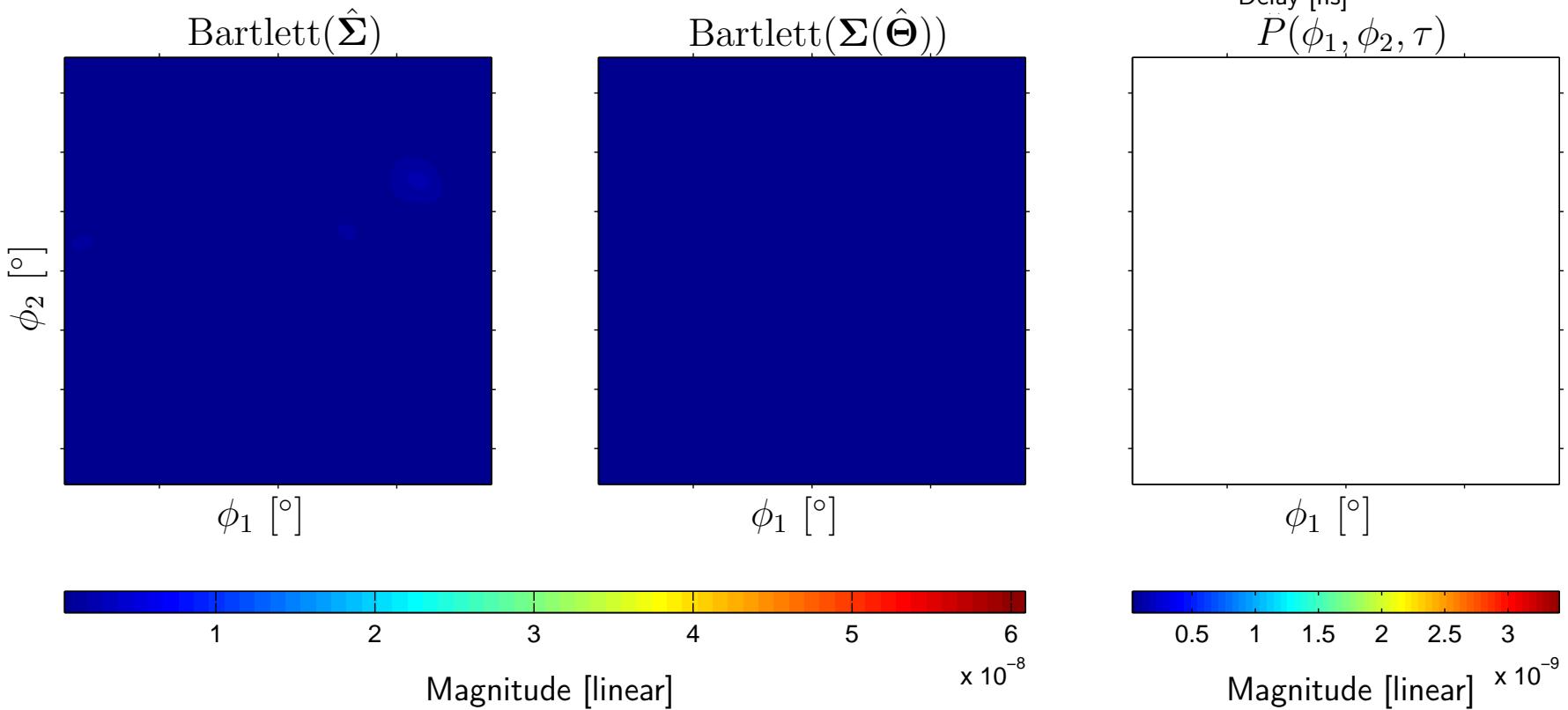
Estimate of biazimuth-delay power spectrum

$\tau = 200 \text{ ns}$



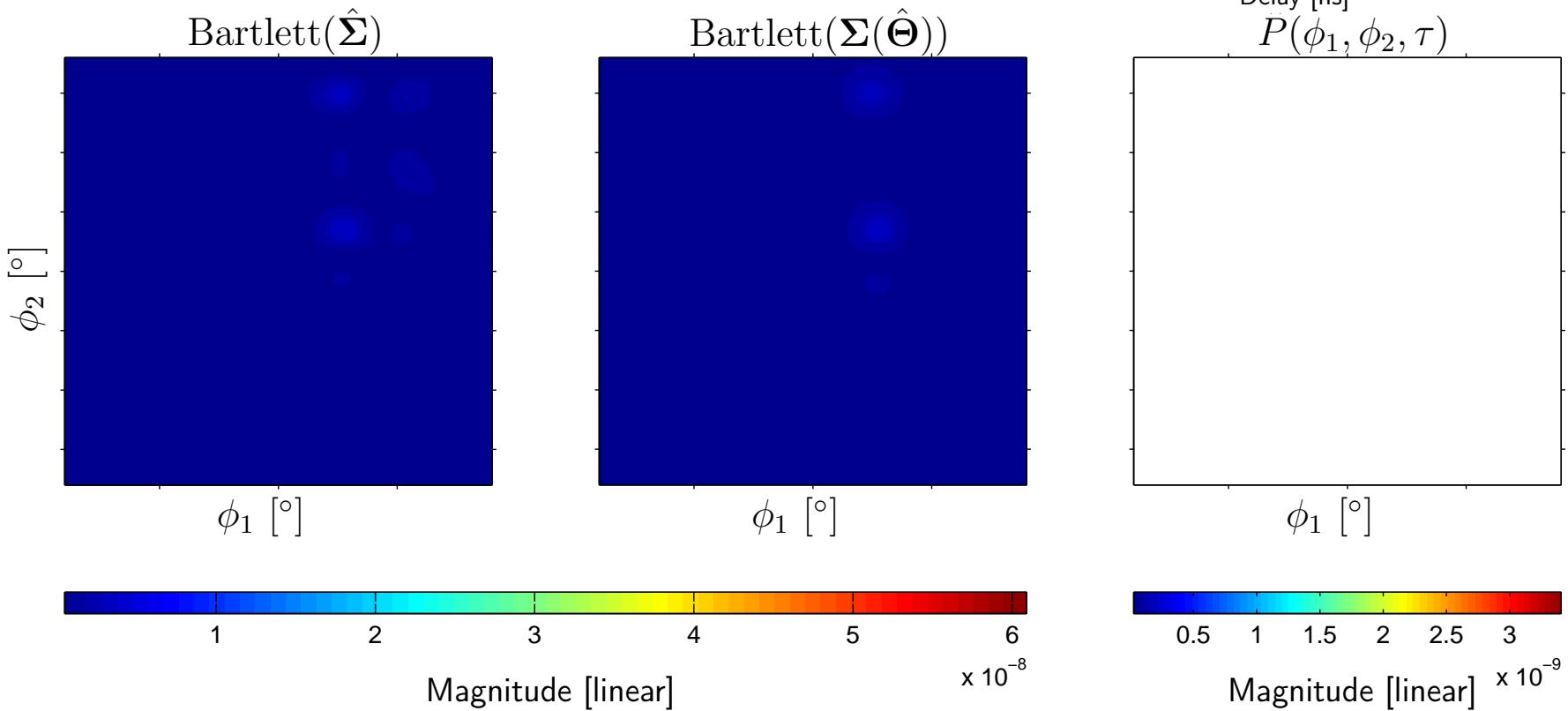
Estimate of biazimuth-delay power spectrum

$$\tau = 205 \text{ ns}$$



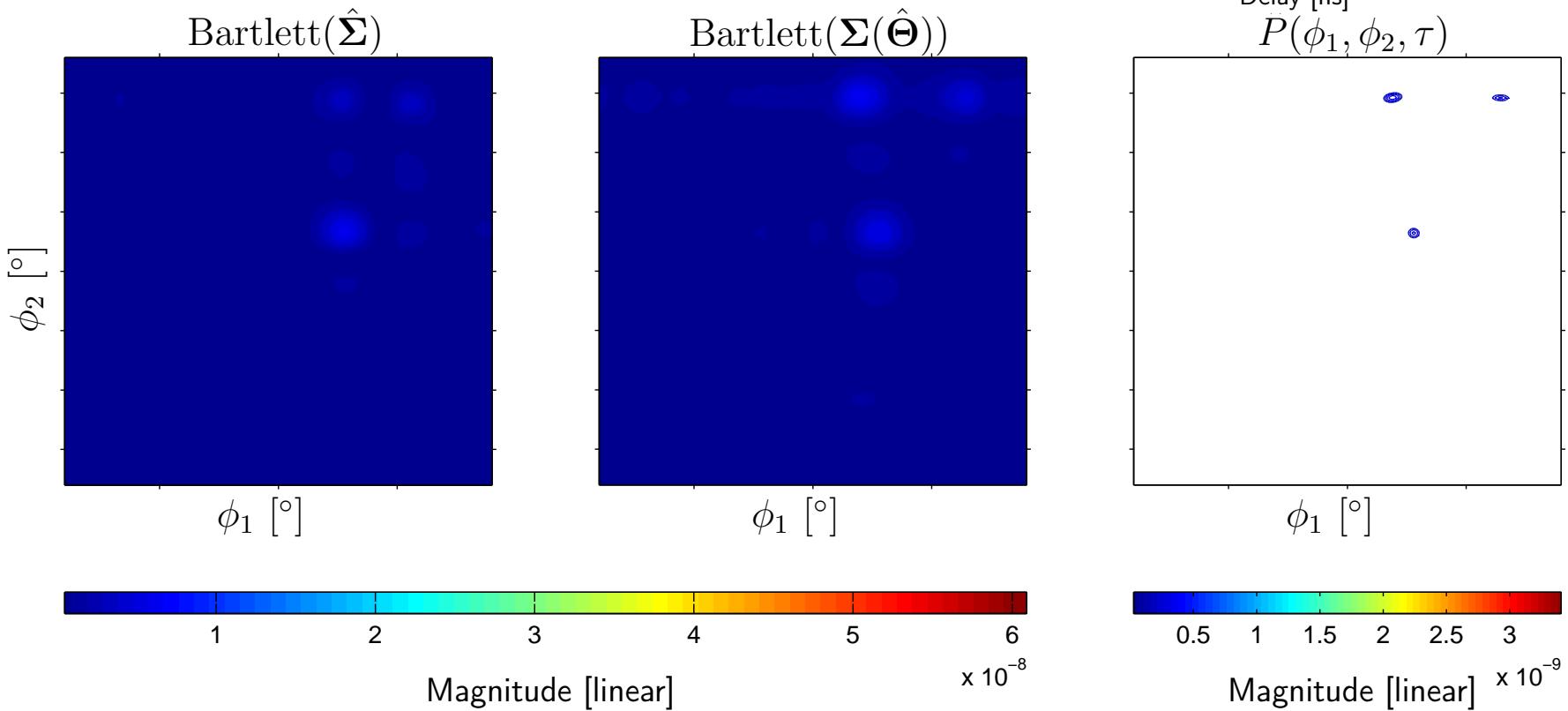
Estimate of biaximuth-delay power spectrum

$$\tau = 210 \text{ ns}$$



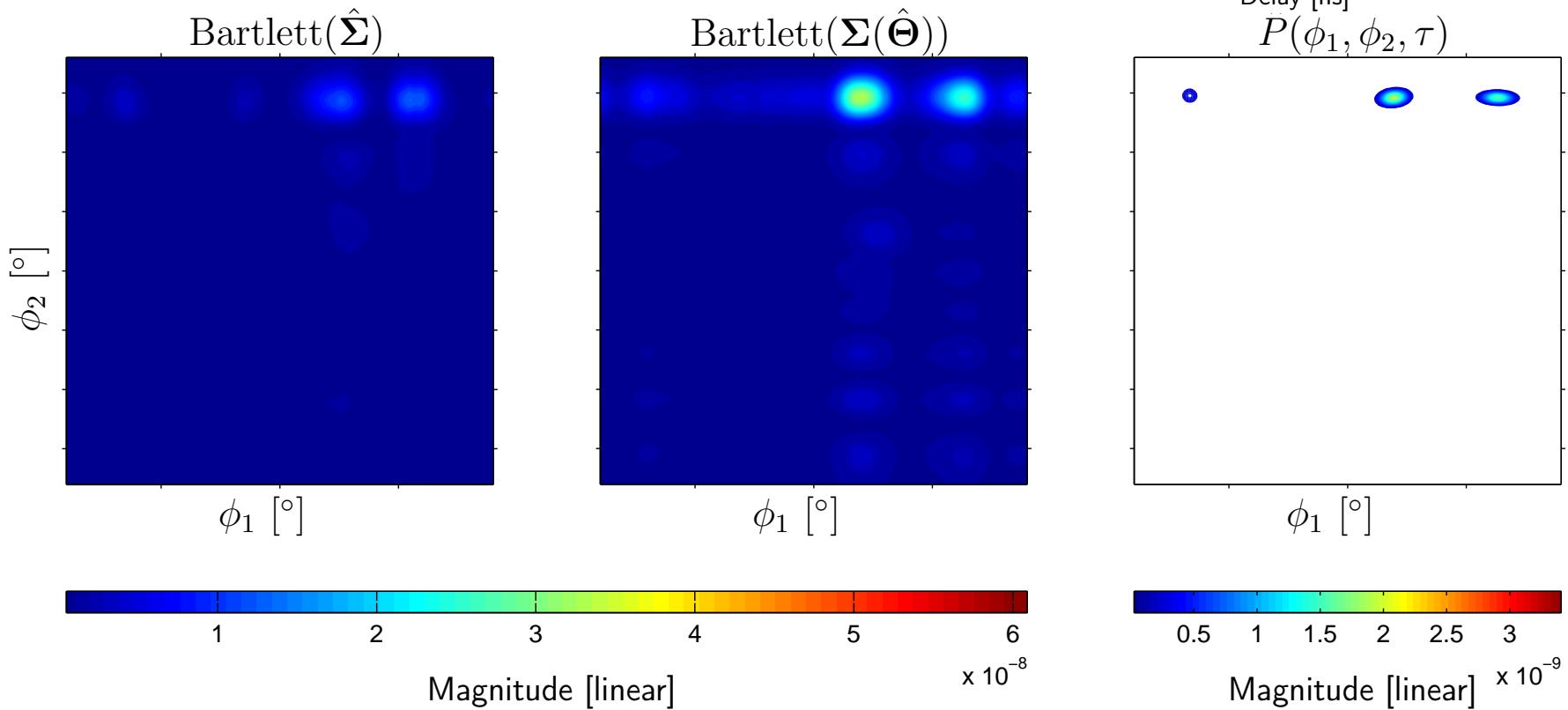
Estimate of biaximuth-delay power spectrum

$$\tau = 215 \text{ ns}$$



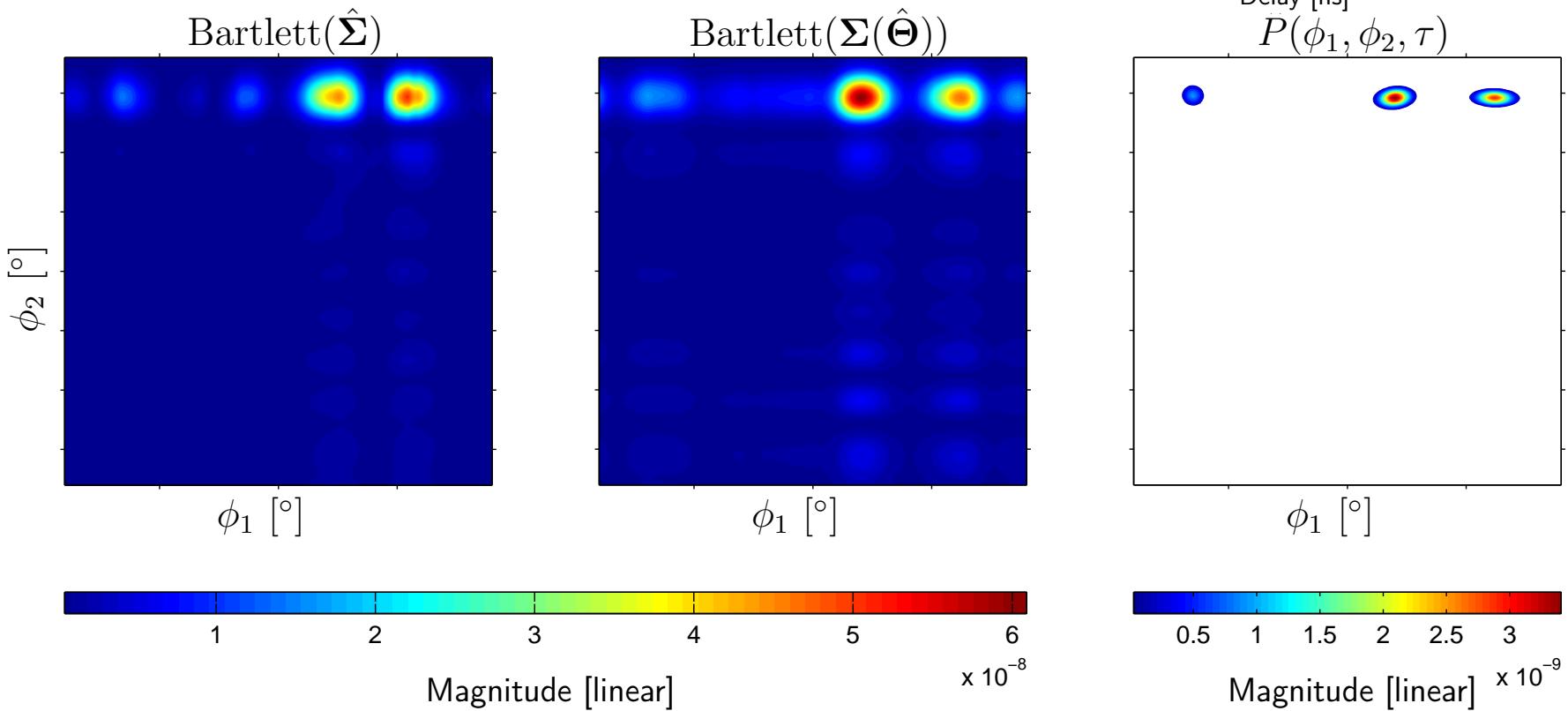
Estimate of biaximuth-delay power spectrum

$\tau = 220 \text{ ns}$



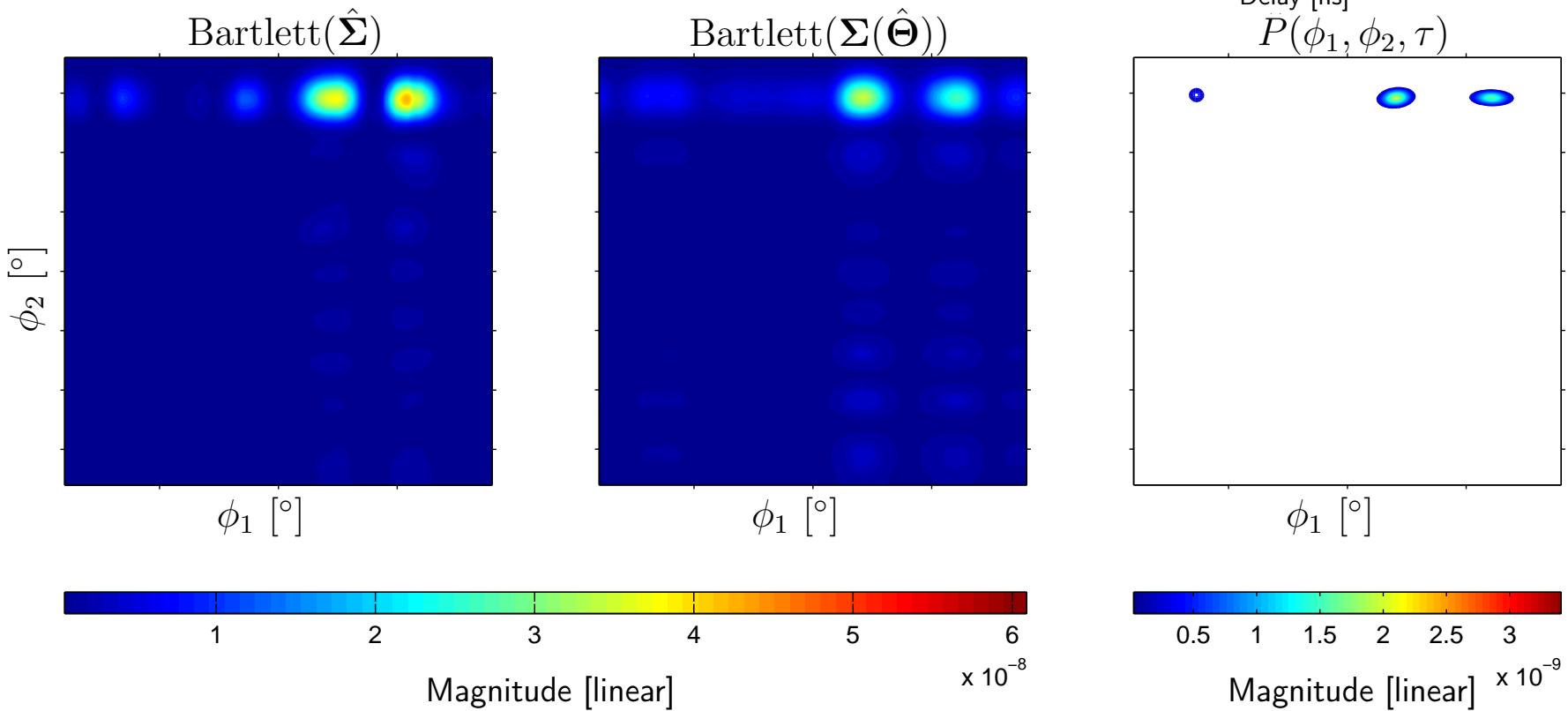
Estimate of biazimuth-delay power spectrum

$$\tau = 225 \text{ ns}$$



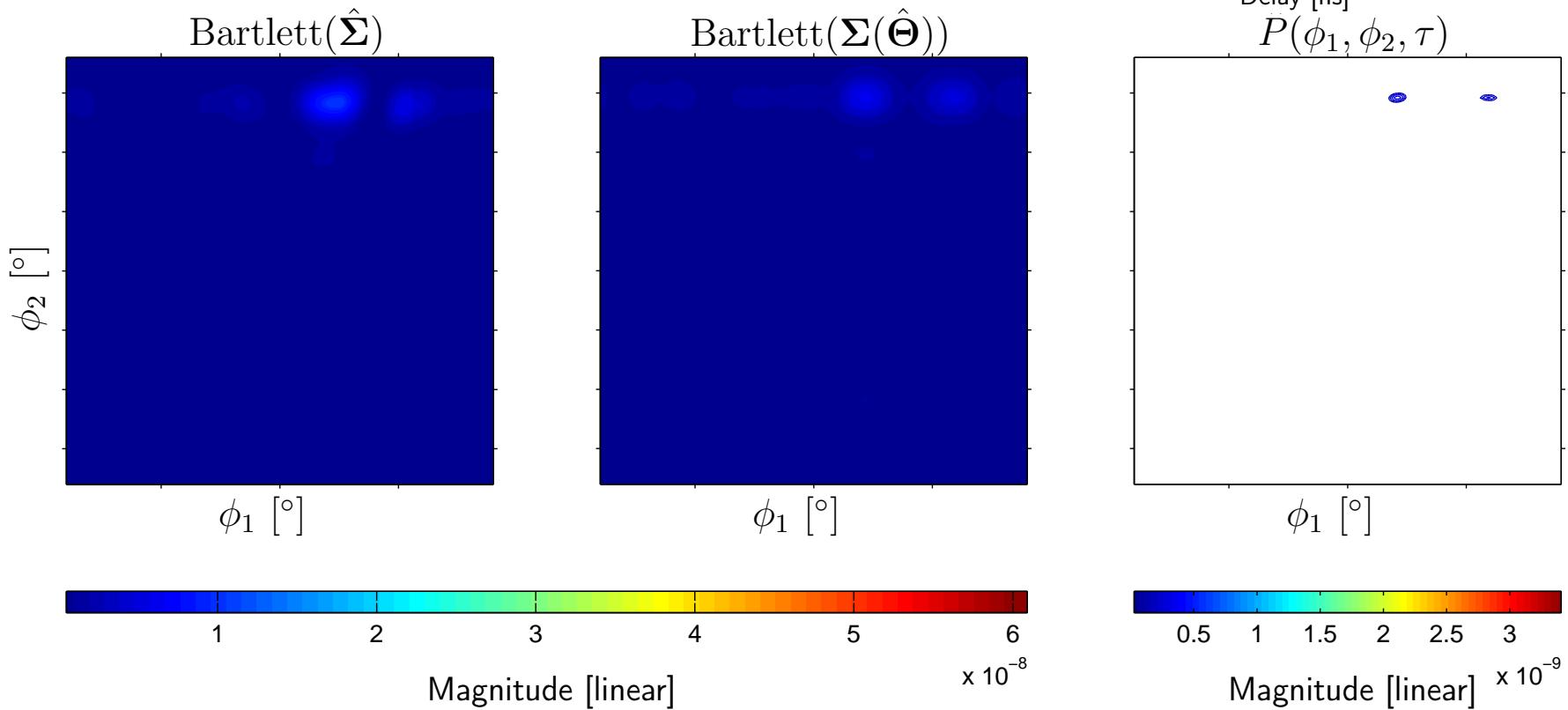
Estimate of biaximuth-delay power spectrum

$$\tau = 230 \text{ ns}$$



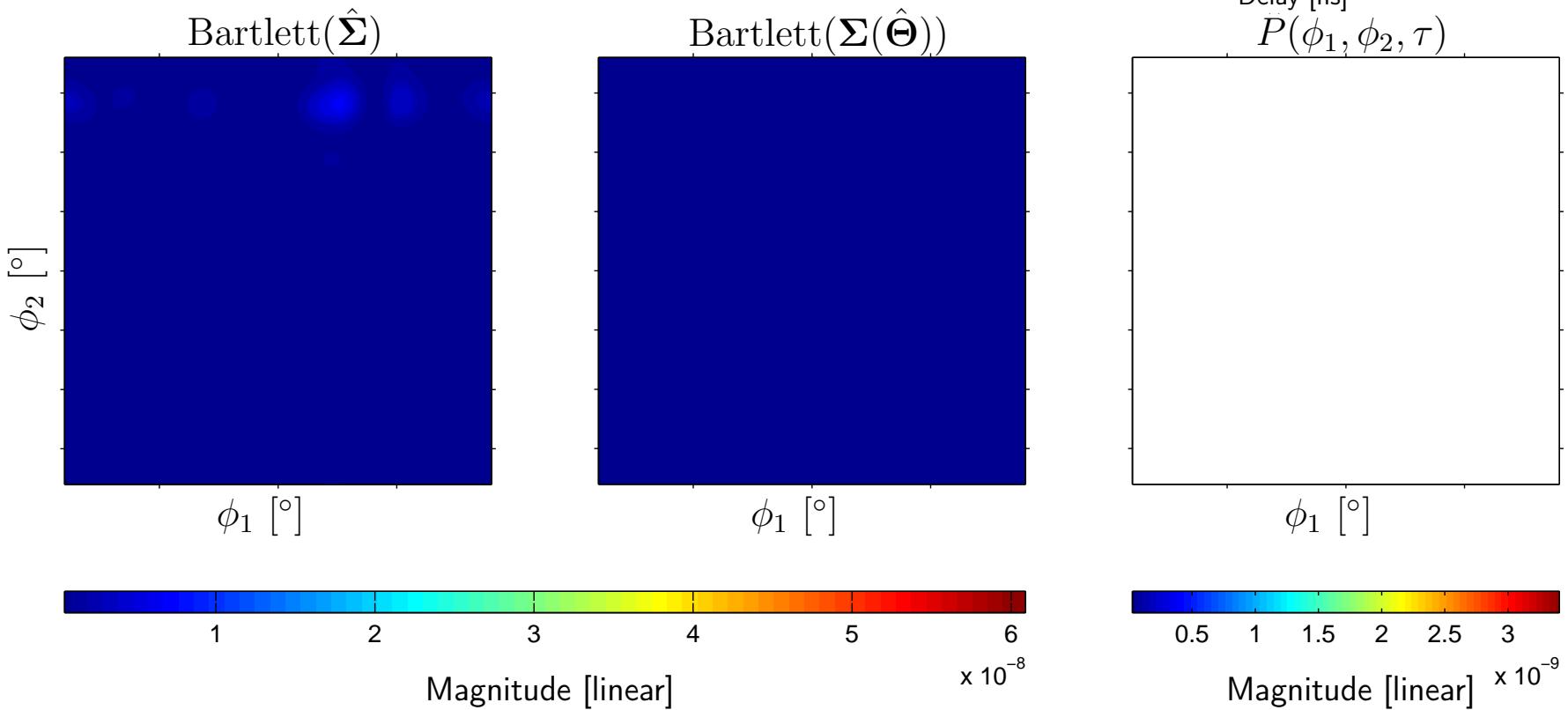
Estimate of biaximuth-delay power spectrum

$$\tau = 235 \text{ ns}$$



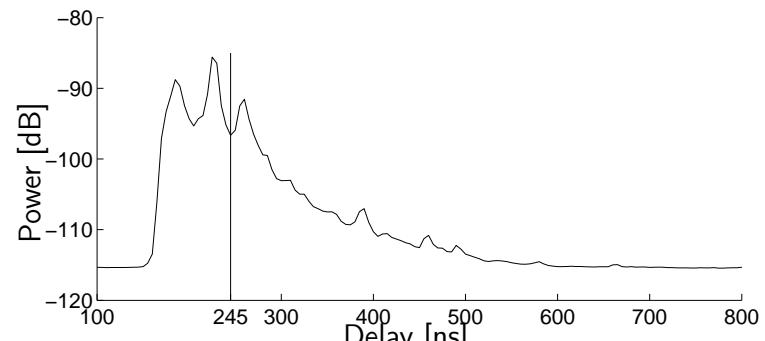
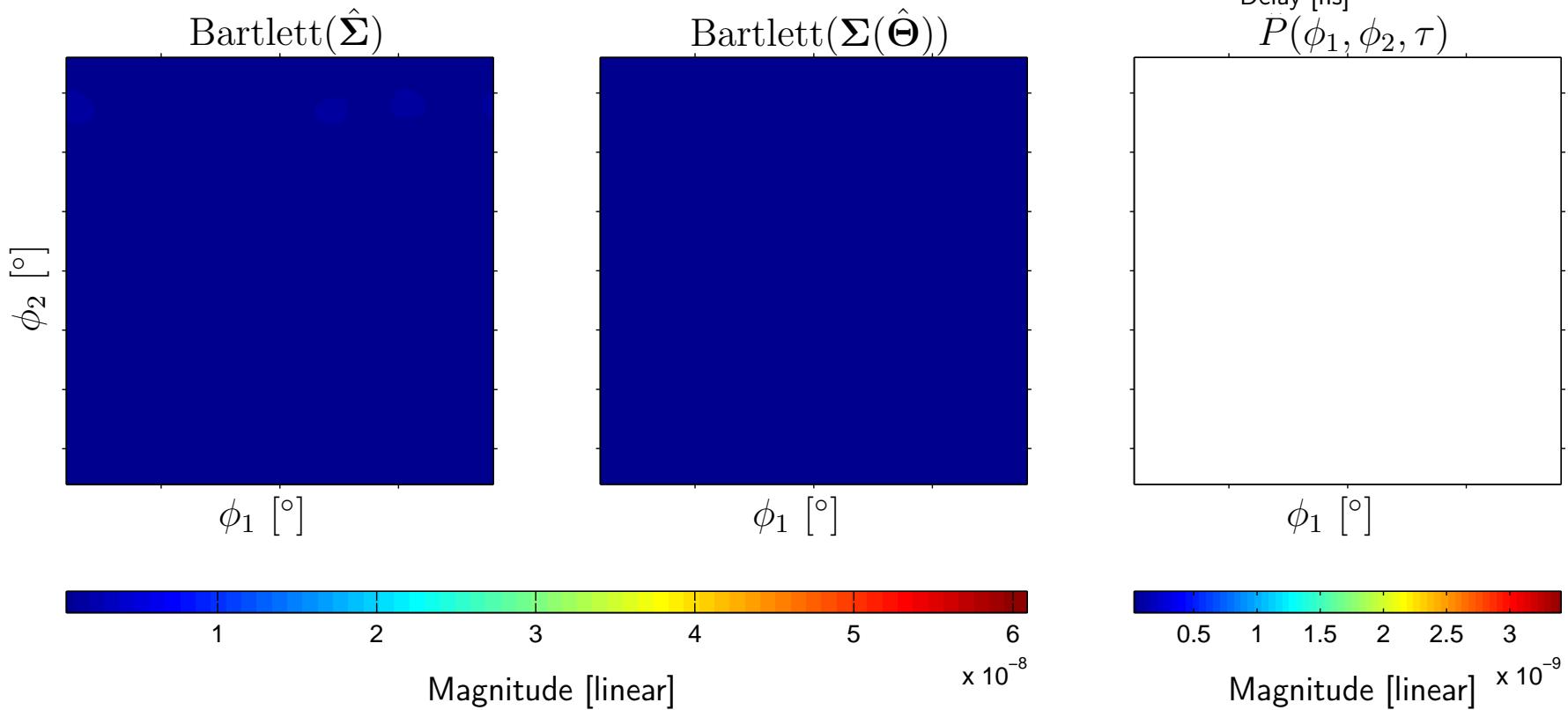
Estimate of biazimuth-delay power spectrum

$$\tau = 240 \text{ ns}$$



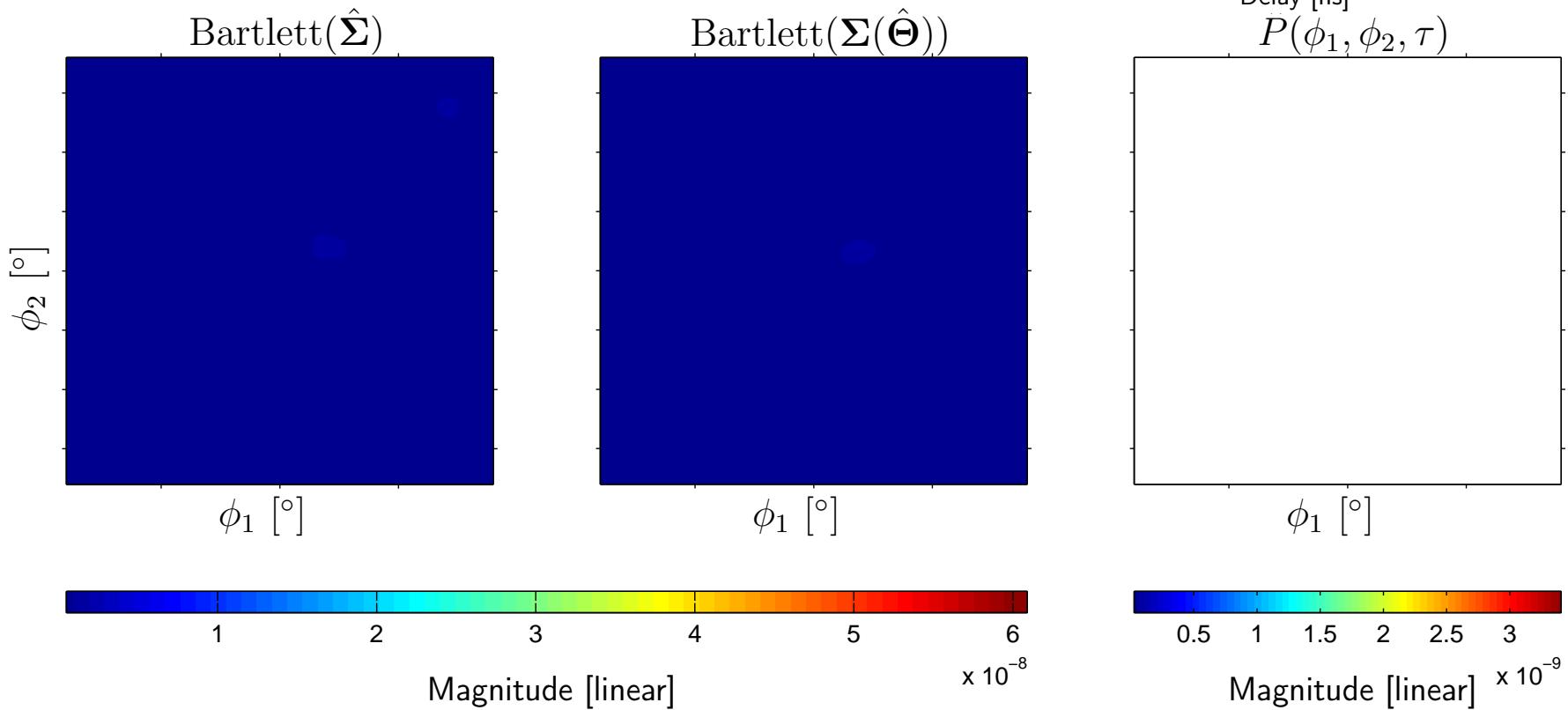
Estimate of biazimuth-delay power spectrum

$$\tau = 245 \text{ ns}$$



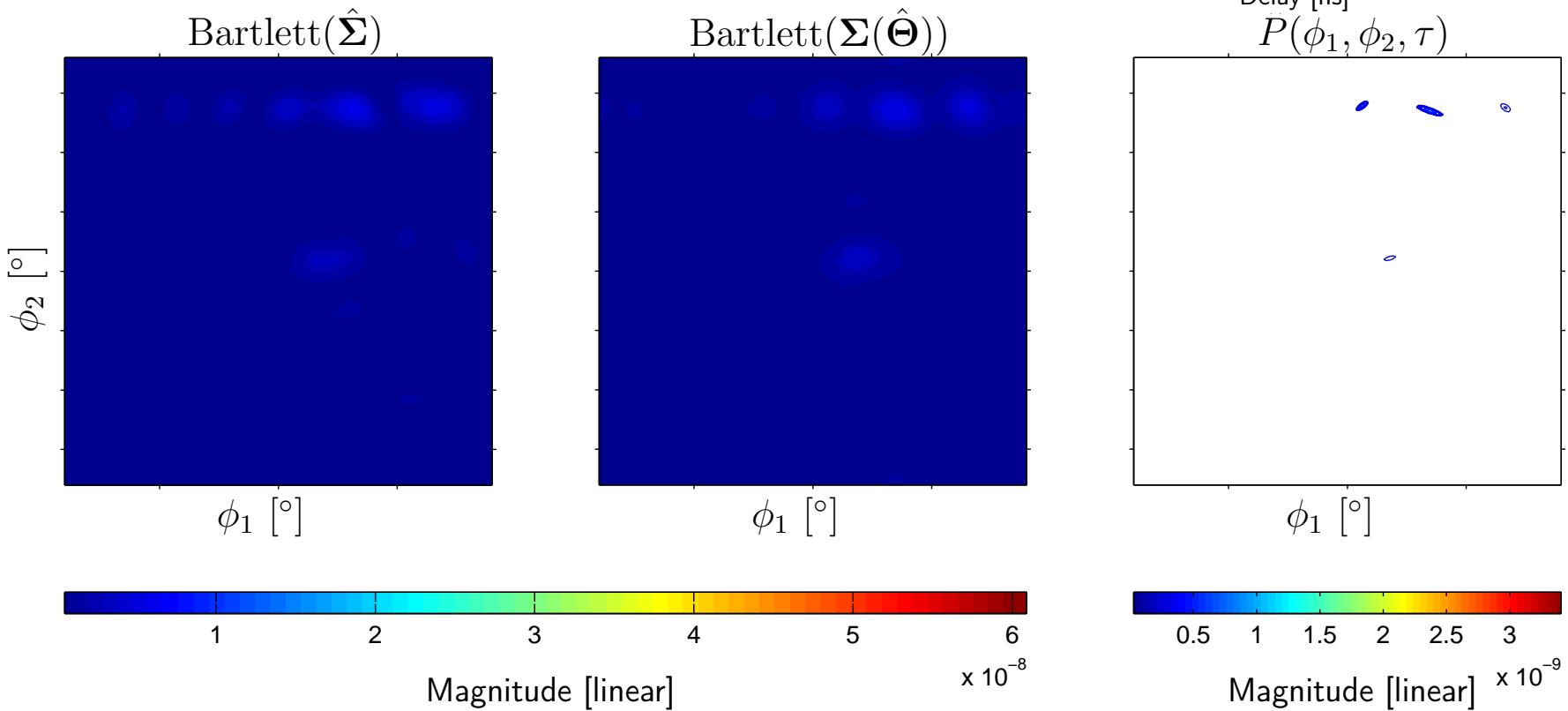
Estimate of biazimuth-delay power spectrum

$\tau = 250 \text{ ns}$



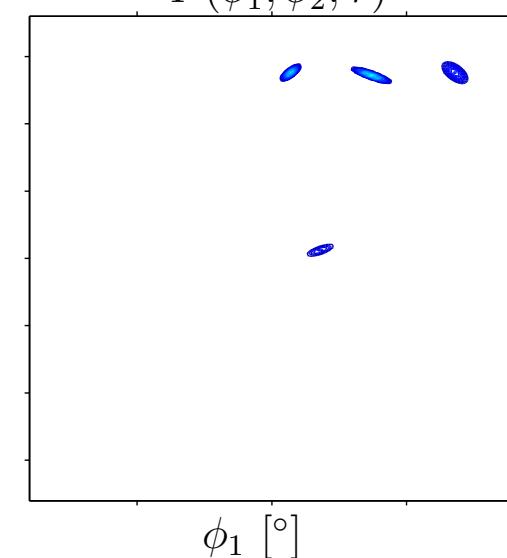
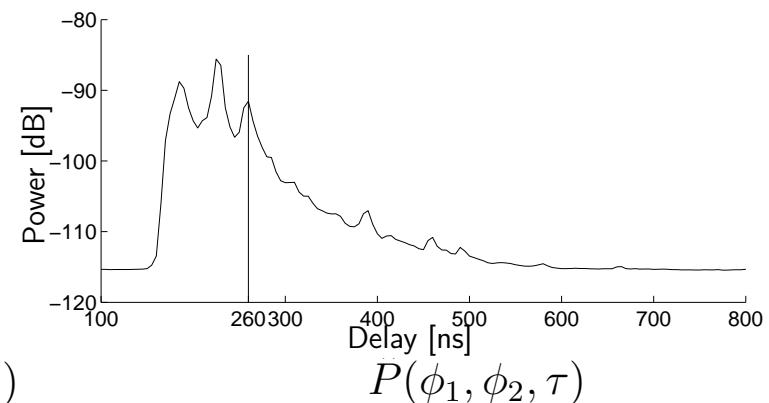
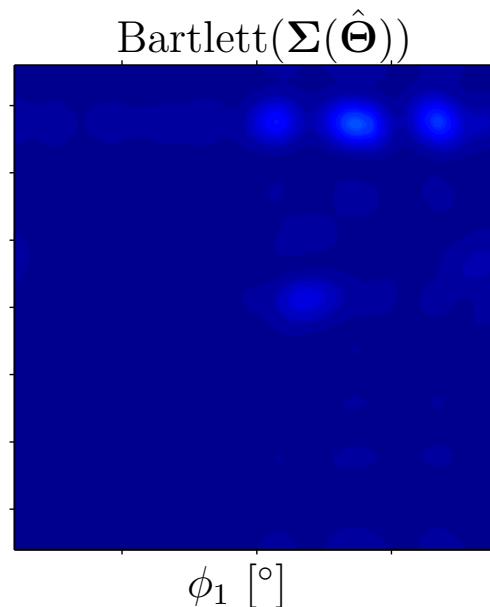
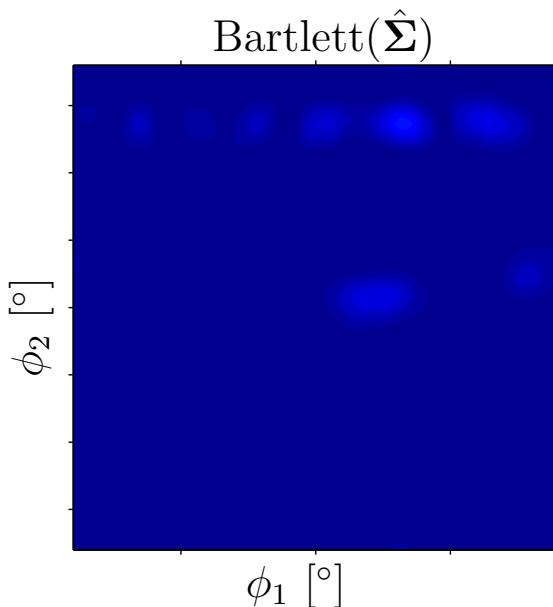
Estimate of biaximuth-delay power spectrum

$$\tau = 255 \text{ ns}$$



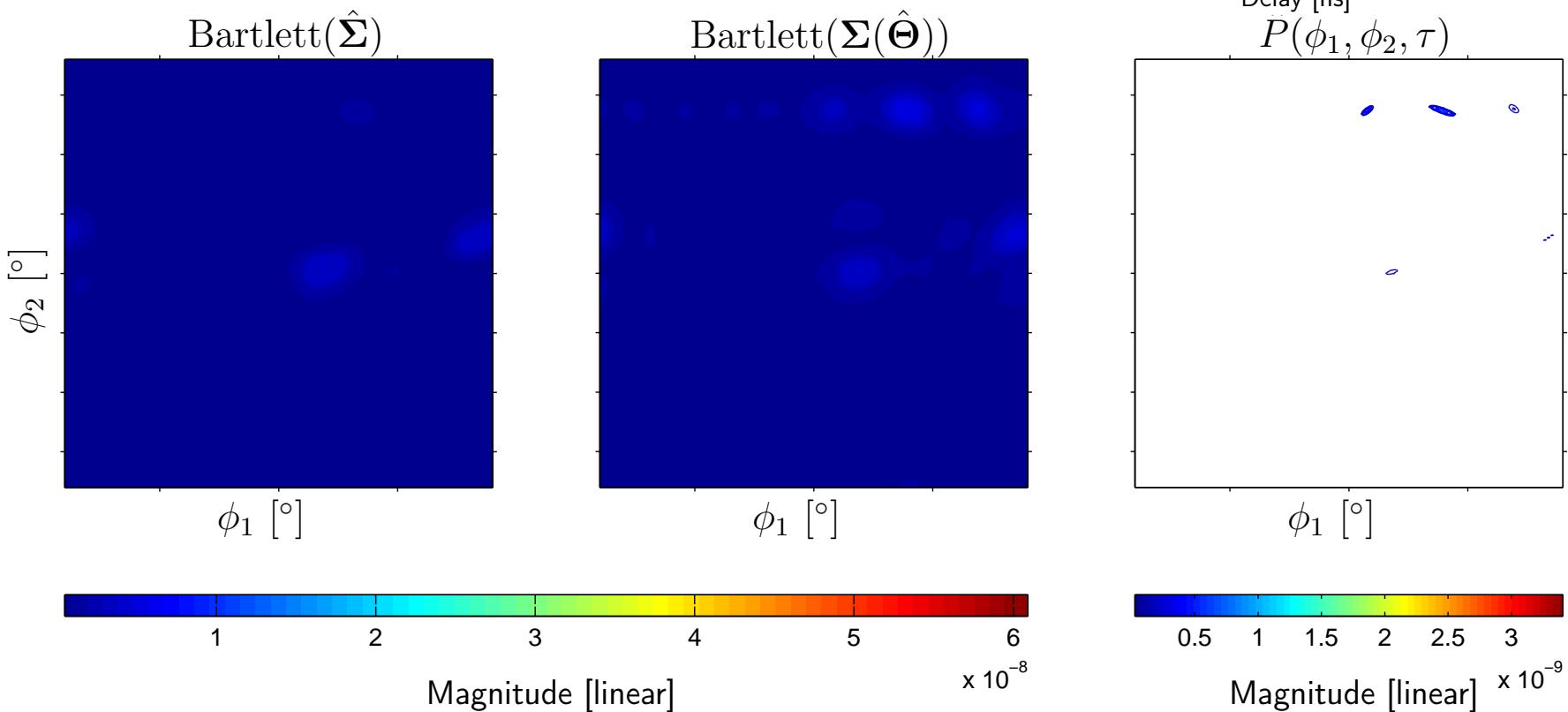
Estimate of biaximuth-delay power spectrum

$\tau = 260 \text{ ns}$



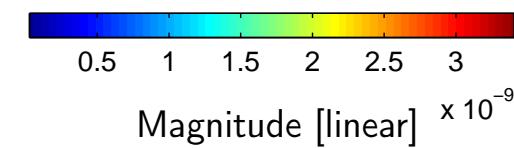
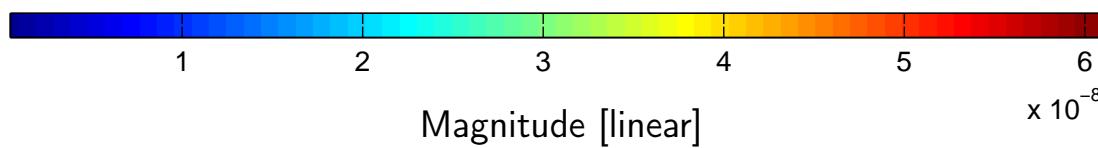
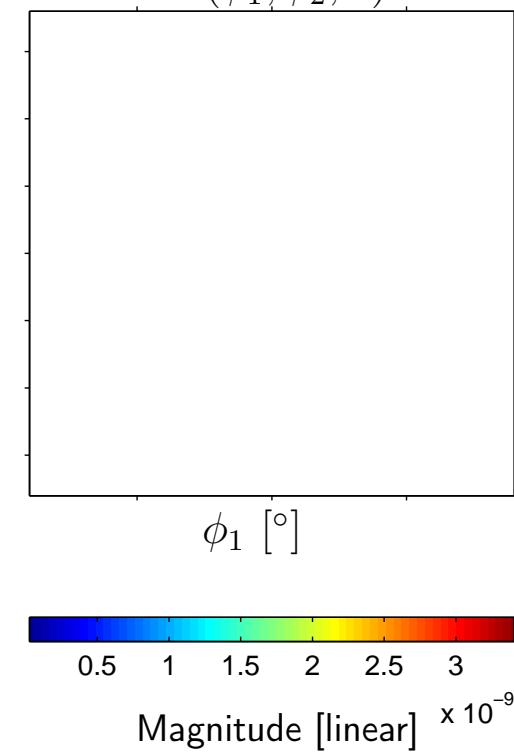
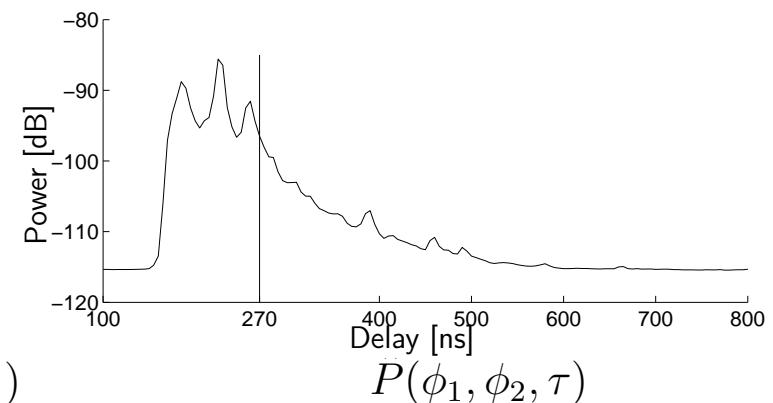
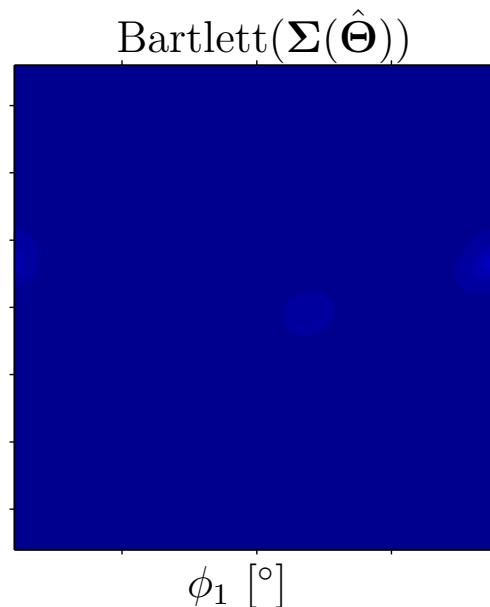
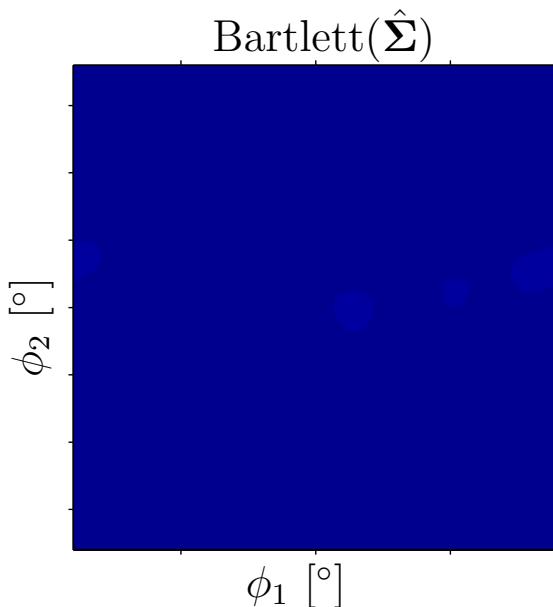
Estimate of biazimuth-delay power spectrum

$$\tau = 265 \text{ ns}$$



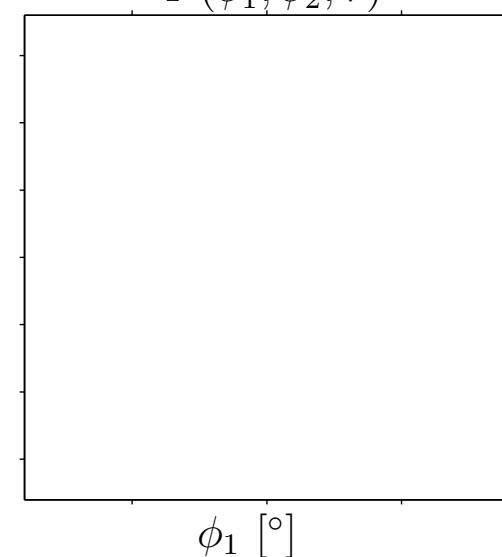
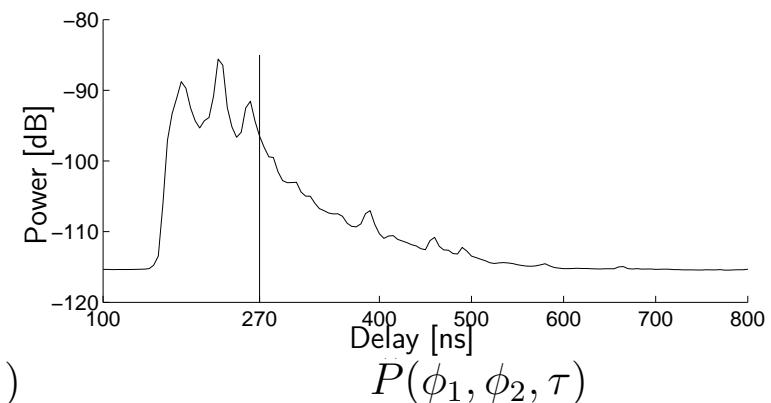
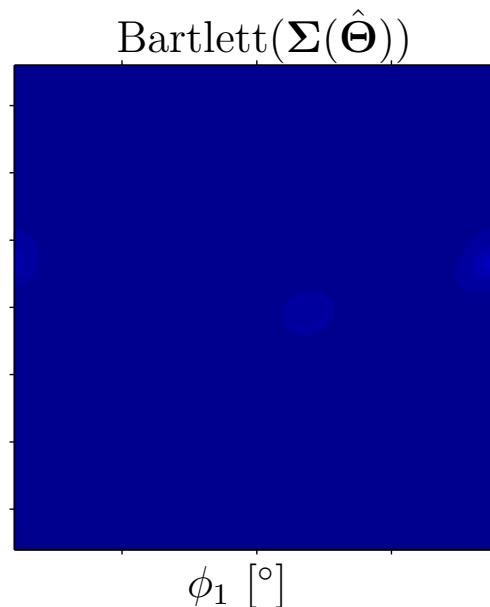
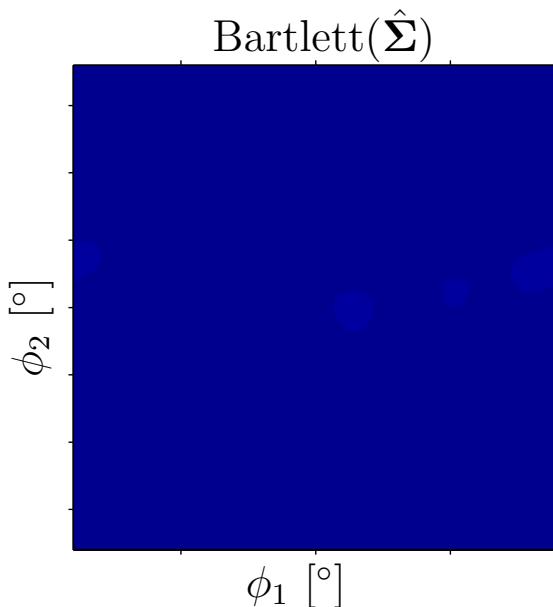
Estimate of biazimuth-delay power spectrum

$\tau = 270 \text{ ns}$



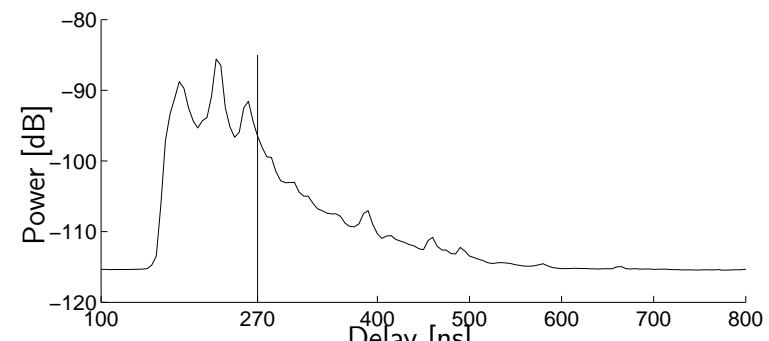
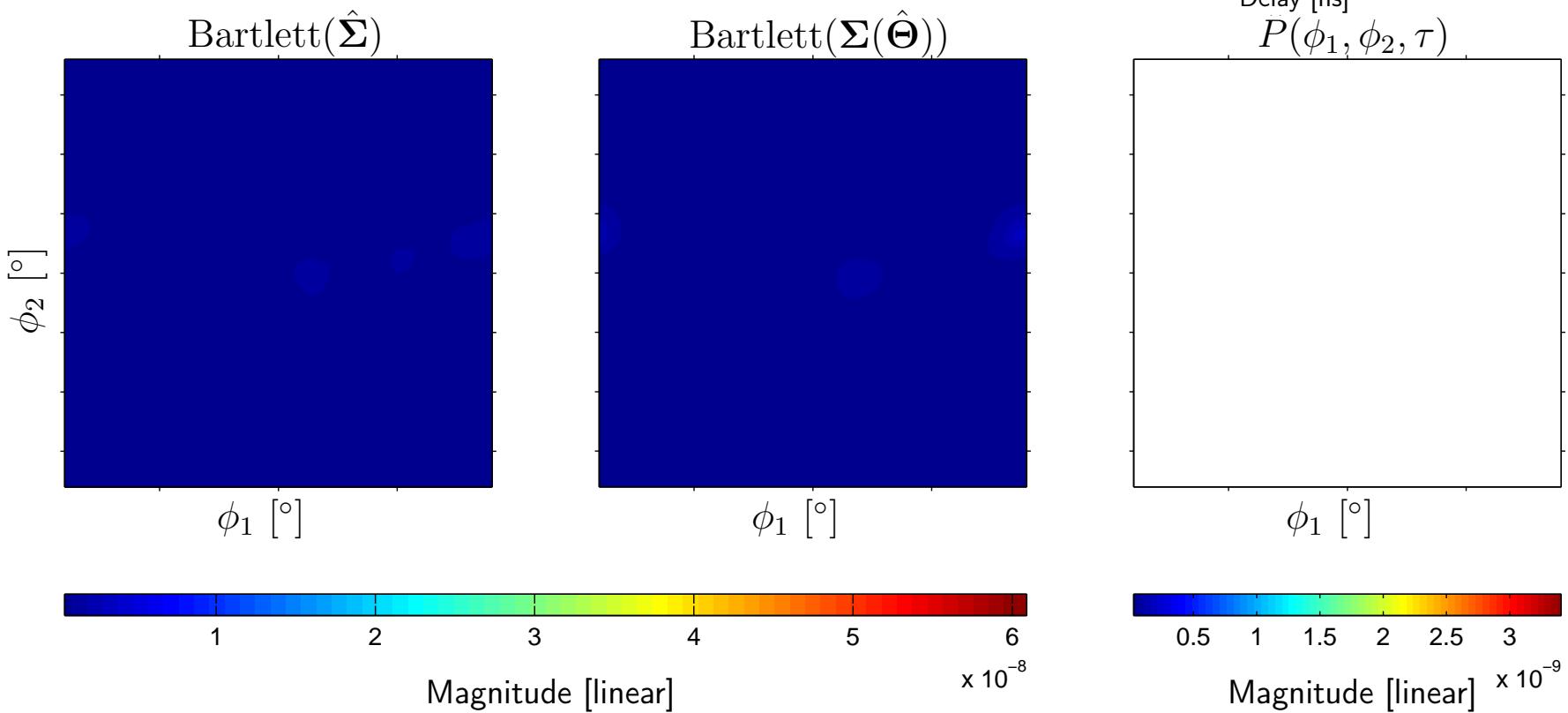
Estimate of biazimuth-delay power spectrum

$\tau = 270 \text{ ns}$



Estimate of biazimuth-delay power spectrum

$\tau = 270 \text{ ns}$



Conclusions

- Pdfs are derived and used to characterize the normalized multi-dimensional power spectral density of individual components in propagation channels.
- The proposed pdfs maximize the entropy under the constraint that its first and second moments are specified.
- Experimental investigations show that the SAGE algorithm together with the characterization method provide an efficient tool for extracting multi-dimensional dispersion of the propagation channel.
- The estimated components are noticeably more concentrated compared to their corresponding footprints in the Bartlett spectrum.
- The results demonstrate the dependence of the spread across the dispersion dimensions for individual components.