

# 3 Kalman filters

## 3.1 Scalar Kalman filter

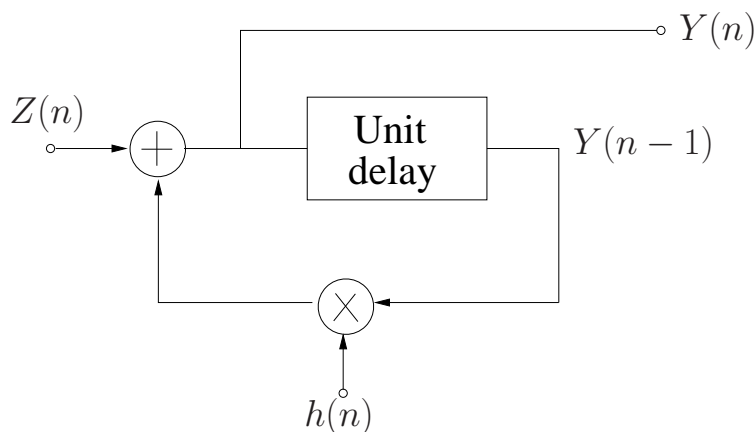
### 3.1.1 Signal model

- **System model**

$\{Y(n)\}$  is an unobservable sequence which is described by the following *state or system equation*:

$$Y(n) = h(n)Y(n - 1) + Z(n), n = 1, 2, \dots \quad (3.1)$$

*Block Diagram Representation of (3.1)*



**Initialization:**

$Y(0)$  is a random variable whose expectation  $\mu_{Y(0)} \equiv E[Y(0)]$  and variance  $\sigma_{Y(0)}^2 \equiv E[(Y(0) - \mu_{Y(0)})^2]$  are known.

**Property of the driving process/noise  $\{Z(n)\}$ :**

$\{Z(n)\}$  is a white noise with a possibly time-varying variance (non-stationary white noise) :

- $E[Z(n)] = 0$
- $E[Z(n)Z(n+k)] = \sigma_{ZZ}^2(n)\delta(k)$

**Property of the feedback coefficients  $\{h(n)\}$ :**

$\{h(n)\}$  is a known deterministic sequence.

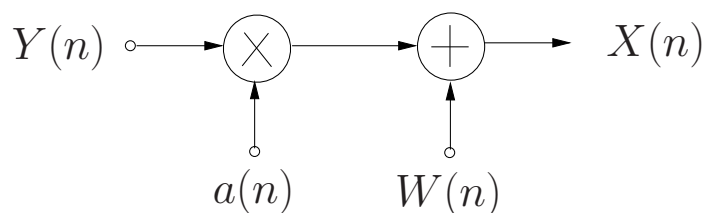
**Remark:** Provided  $\{h(n)\}$  and  $\{\sigma_{ZZ}^2(n)\}$  are constant and  $\{Z(n)\}$  is a Gaussian random process, then  $\{Y(n)\}$  is an AR(1) process.

• **Observation (or channel) model**

The observable sequence  $X(n)$  is given by

$$X(n) = a(n)Y(n) + W(n) \quad , n = 1, 2, \dots \quad (3.2)$$

**Block diagram representation of (3.2)**



**Property of the weighting sequence  $\{a(n)\}$ :**

$\{a(n)\}$  is a known deterministic sequence.

**Property of the noise  $\{W(n)\}$ :**

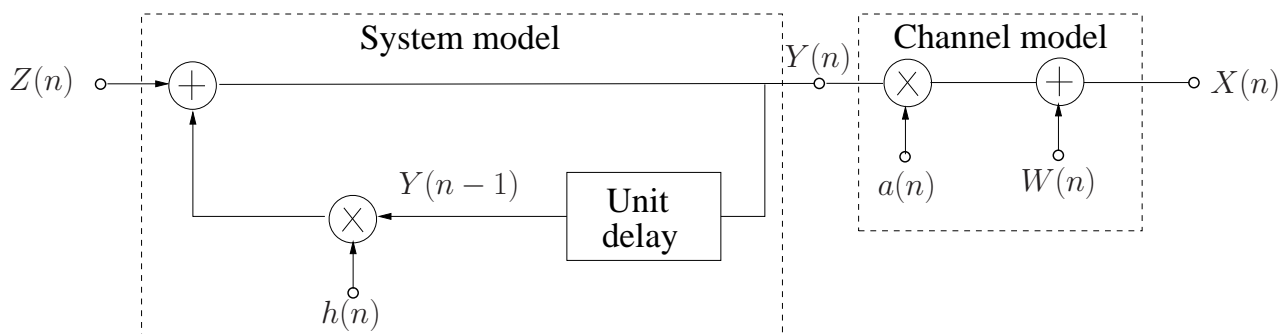
$\{W(n)\}$  is a non-stationary white noise:

- $E[W(n)] = 0$
- $E[W(n)W(n+k)] = \sigma_{WW}^2(n)\delta(k)$

• **Additional "weak independence" assumptions**

$Y(0)$ ,  $\{Z(n)\}$ , and  $\{W(n)\}$  are uncorrelated.

• **Block diagram of the complete signal model**



### 3.1.2 Recursive implementation of the LMMSEE

- **Objective:**

To find a *recursive implementation*<sup>1</sup> of the LMMSEE of  $Y(n)$  based on the observation of  $X(1), \dots, X(n)$ .

- **Recursive implementation:**

We need the following definitions:

–  $\hat{Y}(n | n) \equiv$  LMMSEE of  $Y(n)$  based on the observation of  $X(1), \dots, X(n)$

*Estimation of  $Y(n)$ .*

–  $\hat{Y}(n + 1 | n) \equiv$  LMMSEE of  $Y(n + 1)$  based on the observation of  $X(1), \dots, X(n)$

*One-step prediction of  $Y(n + 1)$  at time  $n$ .*

–  $\hat{X}(n + 1 | n) \equiv$  LMMSEE of  $X(n + 1)$  based on the observation of  $X(1), \dots, X(n)$

*One-step prediction of  $X(n + 1)$ .*

Recursive implementation of the LMMSEE of  $Y(n)$ :

$$\underbrace{\hat{Y}(n + 1 | n + 1)}_{\text{Estimation at time } n + 1} \equiv \mathcal{LF}(\underbrace{\hat{Y}(n | n)}_{\text{Estimation at time } n}, \underbrace{X(n + 1)}_{\text{Observation at time } n + 1})$$

where  $\mathcal{LF}$  denotes a linear function to be found.

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<sup>1</sup>See Section 3.3 for an example of a recursive estimator.

• **We shall know:**

1. Such a recursive implementation of the LMMSEE exists. It is called the ***Kalman Filter***.

2. The recursion is split into two steps:

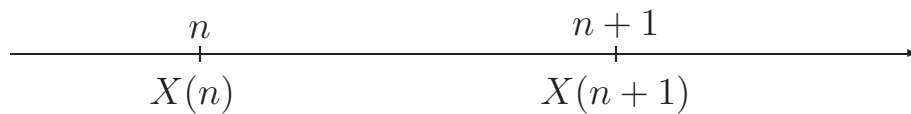
– Step 1: ***One-step prediction***:

$$P : \hat{Y}(n | n) \xrightarrow{P} \hat{Y}(n + 1 | n)$$

– Step 2: ***Updating***:

$$U : \hat{Y}(n + 1 | n) \xrightarrow{X(n+1)} \hat{Y}(n + 1 | n + 1)$$

Temporal evolution of the recursive estimation procedure in the Kalman filter:



$$\hat{Y}(n | n - 1) \xrightarrow{X(n)} \hat{Y}(n | n) \xrightarrow{P} \hat{Y}(n + 1 | n) \xrightarrow{X(n+1)} \hat{Y}(n + 1 | n + 1) \xrightarrow{P} \dots$$

3. The mean-squared estimation error  $E[(Y(n) - \hat{Y}(n | n))^2]$  can also be computed recursively.

We shall need the following definitions:

- $R(n | n) \equiv E[(Y(n) - \hat{Y}(n | n))^2]$   
 $\equiv$  mean-squared estimation error at time  $n$
- $R(n + 1 | n) \equiv E[(Y(n + 1) - \hat{Y}(n + 1 | n))^2]$   
 $\equiv$  mean-squared one-step prediction error at time  $n$

### 3.1.3 Derivation of the equations of the Kalman filter

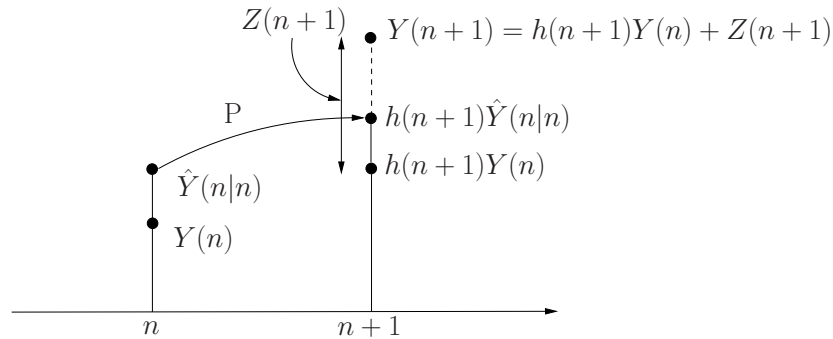
- **Prediction step:**

$$\hat{Y}(n+1 | n) = h(n+1)\hat{Y}(n | n) \quad (3.3)$$

$$R(n+1 | n) = h^2(n+1)R(n | n) + \sigma_{ZZ}^2(n+1) \quad (3.4)$$

**Proof of (3.3):**

(3.3) follows from the linearity property of the expectation.



Let us show that (3.3) satisfies the orthogonality principle (OP) and therefore is the LMMSEE:

Let  $m = 1, \dots, n$  :

$$\begin{aligned} & E[(Y(n+1) - \hat{Y}(n+1 | n))X(m)] \\ &= E[\overbrace{(h(n+1)Y(n) + Z(n+1))} - \overbrace{h(n+1)\hat{Y}(n | n)}] X(m)] \\ &= h(n+1) \underbrace{E[(Y(n) - \hat{Y}(n | n))X(m)]}_{=0} + \underbrace{E[Z(n+1)X(m)]}_{=0} \\ & \hspace{10em} \text{OP for } \hat{Y}(n | n) \hspace{10em} Z(n+1) \text{ and } X(n) \\ & \hspace{10em} \text{are uncorrelated} \\ &= 0 \quad \checkmark \end{aligned}$$

**Proof of (3.4):**

$$\begin{aligned}
 R(n+1 | n) &= E[(Y(n+1) - \hat{Y}(n+1 | n))^2] \\
 &= E[(\underbrace{h(n+1)Y(n) + Z(n+1)} - \underbrace{h(n+1)\hat{Y}(n | n)} )^2] \\
 &= E[(\underbrace{h(n+1)[Y(n) - \hat{Y}(n | n)]} + \underbrace{Z(n+1)} )^2]
 \end{aligned}$$

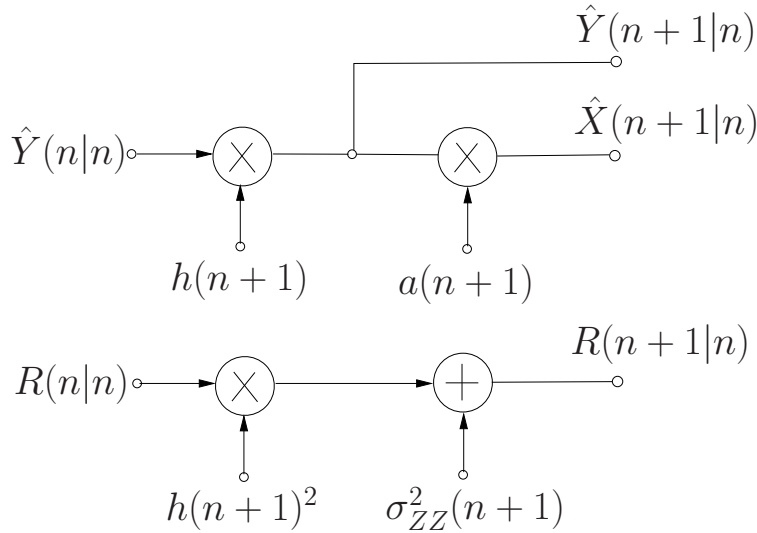
These two random variables are uncorrelated

$$\begin{aligned}
 &= h(n+1)^2 E[(Y(n) - \hat{Y}(n | n))^2] + E[Z(n+1)^2] \\
 &= h(n+1)^2 R(n | n) + \delta_{ZZ}^2(n+1)
 \end{aligned}$$

With the same argument as that used for the proof of (3.3) we show that

$$\hat{X}(n+1 | n) = a(n+1)\hat{Y}(n+1 | n)$$

**Block diagram of the prediction step:**





• **Updating step:**

$$\hat{Y}(n+1 | n+1) = \hat{Y}(n+1 | n) + b(n+1)[X(n+1) - \hat{X}(n+1 | n)] \quad (3.5)$$

$$R(n+1 | n+1) = [1 - b(n+1)a(n+1)]R(n+1 | n) \quad (3.6)$$

with

$$b(n+1) \equiv \frac{a(n+1)R(n+1 | n)}{a(n+1)^2R(n+1 | n) + \sigma_{WW}^2(n+1)}$$

*Interpretation of (3.5):*

$$\hat{Y}(n+1 | n+1) = \underbrace{\hat{Y}(n+1 | n)}_{\substack{\text{One-step} \\ \text{prediction} \\ \text{of } Y(n+1)}} + b(n+1) \underbrace{\left[ \underbrace{X(n+1)}_{\substack{\text{New} \\ \text{observation}}} - \underbrace{\hat{X}(n+1 | n)}_{\substack{\text{One-step} \\ \text{prediction} \\ \text{of } X(n+1)}} \right]}_{\substack{\text{Residual error} \\ \text{of } \hat{X}(n+1 | n)}} \underbrace{\hspace{10em}}_{\text{Correction factor}}$$

**Kalman gain:**

The coefficient  $b(n)$  is called the **Kalman gain** of the filter.

**Proof of (3.5) :**

We seek an updating equation given by (3.5) and determine  $b(n + 1)$  so that (3.5) satisfies the orthogonality principle.

1<sup>st</sup> case:  $m = 1, \dots, n$

$$\begin{aligned}
 & E[(Y(n + 1) - \hat{Y}(n + 1 | n + 1))X(m)] \\
 &= \underbrace{E[(Y(n + 1) - \hat{Y}(n + 1 | n))X(m)]}_{= 0} - b(n + 1) \underbrace{E[(X(n + 1) - \hat{X}(n + 1 | n))X(m)]}_{= 0} \\
 &\qquad\qquad\qquad \text{OP for } \hat{Y}(n+1 | n) \qquad\qquad\qquad \text{OP for } \hat{X}(n + 1 | n) \\
 &= 0 \quad \checkmark
 \end{aligned}$$

2<sup>nd</sup> case:  $m = n + 1$

$$\begin{aligned}
 & E[(Y(n + 1) - \hat{Y}(n + 1 | n + 1))X(n + 1)] \\
 &= E[(Y(n + 1) - \hat{Y}(n + 1 | n))X(n + 1)] \\
 &\quad - b(n + 1)E[(X(n + 1) - \hat{X}(n + 1 | n))X(n + 1)]
 \end{aligned}$$

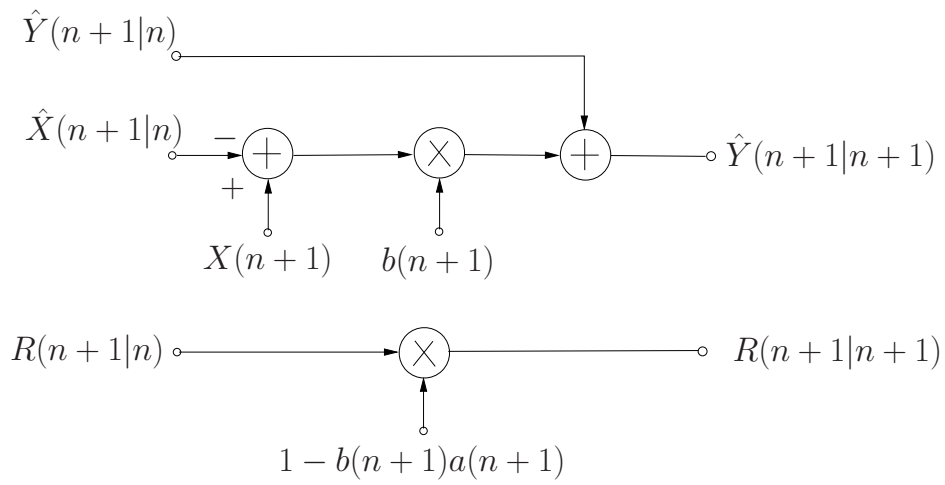
We determine  $b(n + 1)$  such that the above expression vanishes:

$$b(n + 1) = \frac{E[(Y(n + 1) - \hat{Y}(n + 1 | n))X(n + 1)]}{E[(X(n + 1) - \hat{X}(n + 1 | n))X(n + 1)]} = \frac{I}{II}$$





**Block diagram of the updating step:**

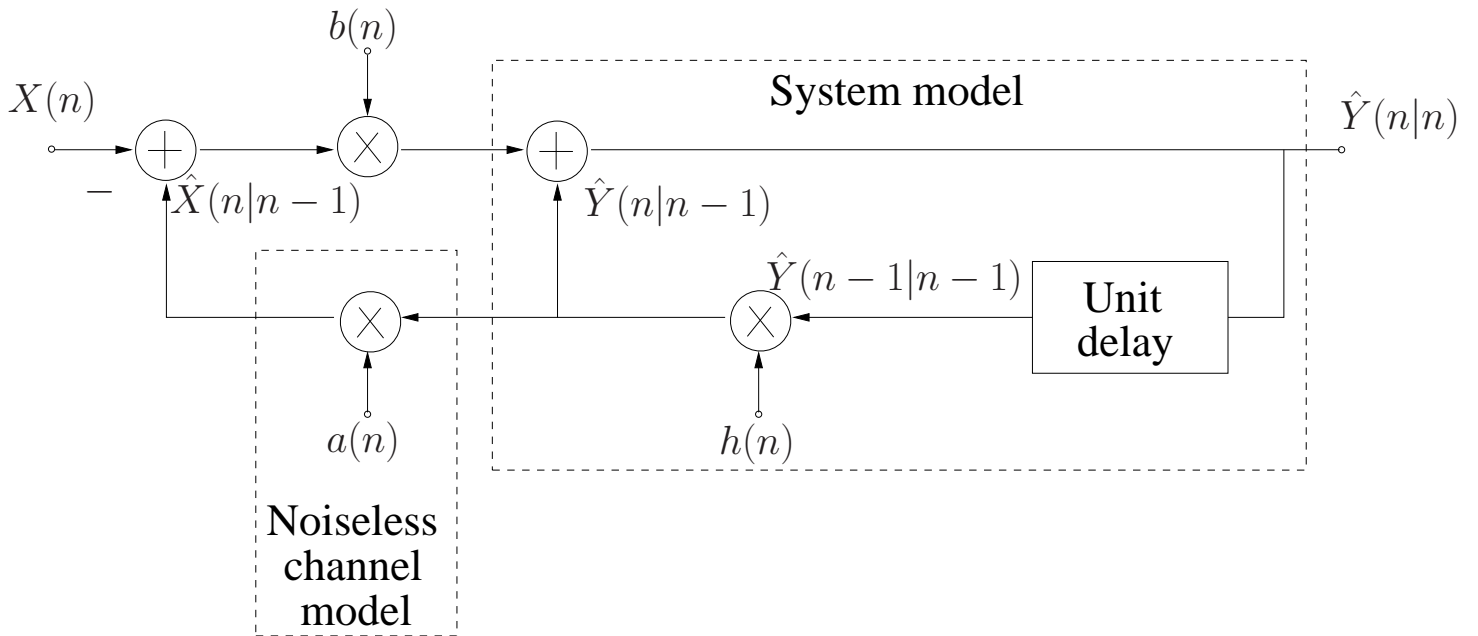


• **Initialization:**

$$\hat{Y}(0 | 0) = \mu_{Y(0)}$$

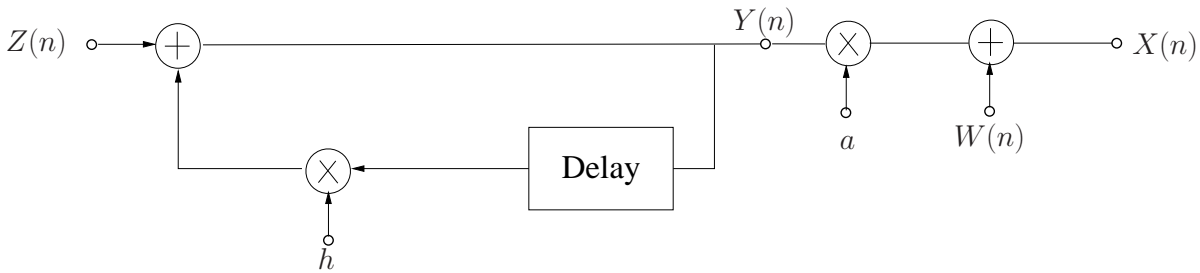
$$R(0 | 0) = \sigma_{Y(0)}^2$$

*Block diagram of the Scalar Kalman filter:*



### 3.1.4 Steady-state Kalman filter when the system and channel models are time-invariant

We consider the time-invariant system and channel models as depicted below:



The system-driving process  $Z(n)$  and the channel noise  $W(n)$  are uncorrelated white wide-sense stationary process:

- $E[Z(n)Z(n+k)] = \sigma_{ZZ}^2(n)\delta(k)$
- $E[W(n)W(n+k)] = \sigma_{WW}^2\delta(k)$

Equations of the Kalman filter estimating  $Y(n)$ :

$$R(n+1 | n) = h^2 R(n | n) + \sigma_{ZZ}^2$$

$$b(n+1) = \frac{aR(n+1 | n)}{a^2 R(n+1 | n) + \sigma_{WW}^2}$$

$$R(n+1 | n+1) = [1 - ab(n+1)]R(n+1 | n)$$

For  $n \rightarrow \infty$  the three sequences  $\{R(n+1 | n)\}$ ,  $\{b(n)\}$ , and  $\{R(n+1 | n+1)\}$  converge, i.e.

$$\begin{aligned} R(n+1 | n) &\rightarrow R_{p\infty} \\ b(n) &\rightarrow b_\infty & n \rightarrow \infty \\ R(n+1 | n+1) &\rightarrow R_\infty \end{aligned}$$

The Kalman filter converges to its *steady state*.

The above limits can be calculated by inserting them into the equations of the Kalman filter:

$$R_{p\infty} = h^2 R_\infty + \sigma_{ZZ}^2 \quad (3.7)$$

$$b_\infty = \frac{a R_{p\infty}}{a^2 R_{p\infty} + \sigma_{WW}^2} \quad (3.8)$$

$$R_\infty = [1 - a b_\infty] R_{p\infty} \quad (3.9)$$

Inserting (3.8) into (3.9), we obtain

$$\begin{aligned} R_\infty &= \left[ 1 - \frac{a^2 R_{p\infty}}{a^2 R_{p\infty} + \sigma_{WW}^2} \right] R_{p\infty} \\ &= \frac{\sigma_{WW}^2 R_{p\infty}}{a^2 R_{p\infty} + \sigma_{WW}^2} \end{aligned}$$



Substituting (3.7) into the last expression yields the so-called steady-state Ricatti equation

$$R_{\infty} = \frac{\sigma_{WW}^2 [h^2 R_{\infty} + \sigma_{ZZ}^2]}{a^2 [h^2 R_{\infty} + \sigma_{ZZ}^2] + \sigma_{WW}^2}$$

The Ricatti equation is a quadratic equation that can be solved numerically. e.g. by using Newton's method.

Then,  $R_{p\infty}$  and  $b(\infty)$  follow by inserting the numerical solution for  $R_{\infty}$  into (3.7) and (3.8), respectively.

***Example:***

The steady-state solutions for the model with parameter setting

- $h = 0.9$
- $a = 0.1$
- $\sigma_{ZZ} = \sigma_{WW} = 1$

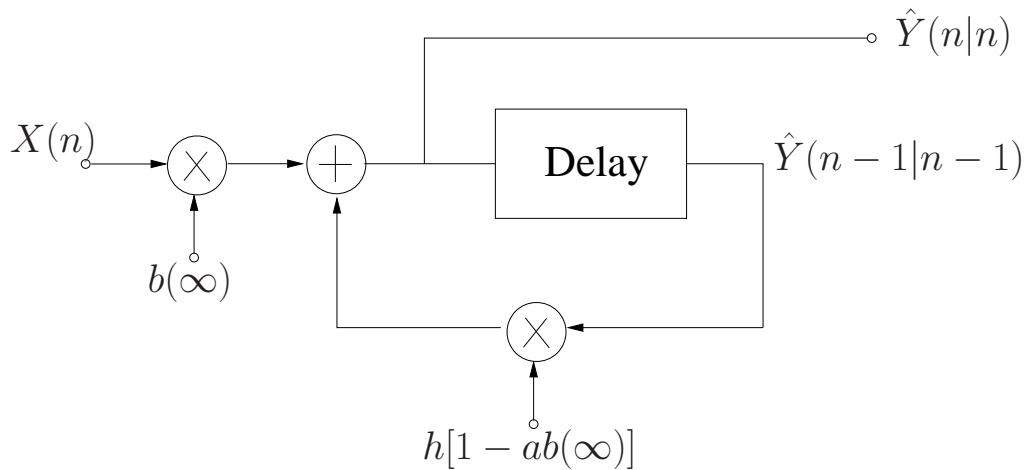
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- $R_{\infty} = 0.5974$
- $R_{p\infty} = 1.4839$
- $b_{\infty} = 0.5974$

Input-output relationship of the steady-state Kalman filter:

$$\begin{aligned}\hat{Y}(n | n) &= h\hat{Y}(n - 1 | n - 1) + b_\infty[X(n) - ah\hat{Y}(n - 1 | n - 1)] \\ &= b_\infty X(n) + h[1 - ab_\infty]\hat{Y}(n - 1 | n - 1)\end{aligned}$$

Block-diagram of the steady-state Kalman filter



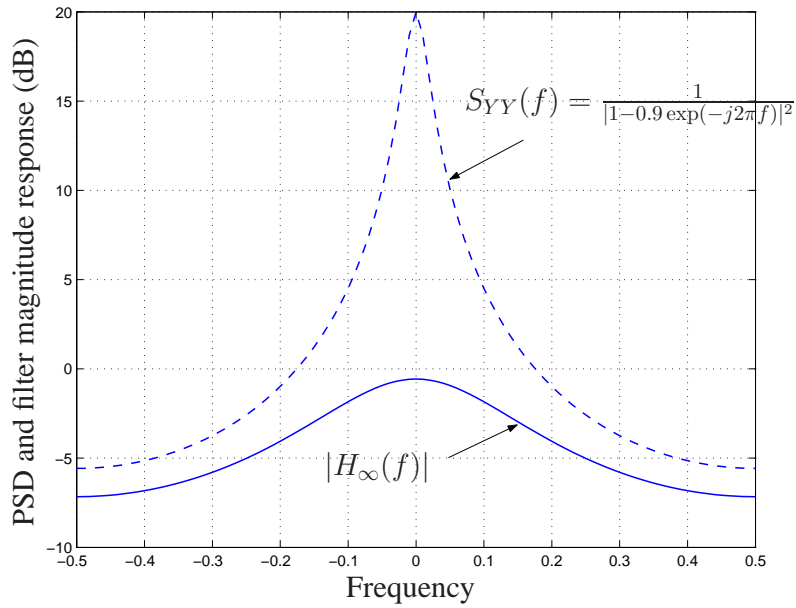
The steady-state Kalman filter is an infinite impulse response (IIR) filter with transfer function

$$H_\infty(f) = \frac{b_\infty}{1 - h[1 - ab_\infty] \exp(-j2\pi f)}$$

$$H_\infty(z) = \frac{b_\infty}{1 - h[1 - ab_\infty]z^{-1}}$$

**Example (cont'd):**

$$H_\infty(f) = \frac{0.5974}{1 - 0.3623 \cdot \exp(-j2\pi f)}$$



**Comment:**

The steady-state Kalman filter calculates the LMMSEE of  $Y(n)$  based on the observation of the sequence  $\{X(n)\}$  in the time window  $[n, n - 1, n - 2, \dots]$

Hence, the steady-state Kalman filter implements the **Causal Wiener filter**.

## 3.2 Vector Kalman Filter

### 3.2.1 Signal Model

- System model

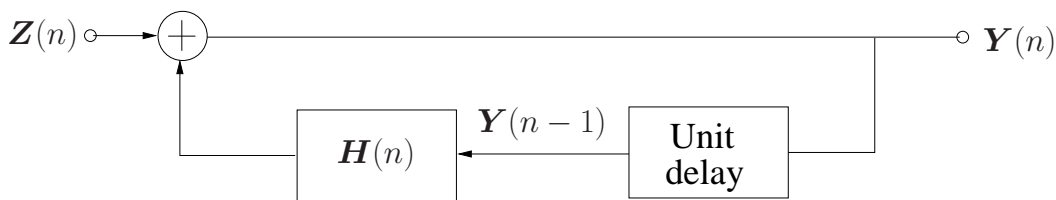
$$\mathbf{Y}(n) = \mathbf{H}(n)\mathbf{Y}(n-1) + \mathbf{Z}(n), \quad n = 1, 2, \dots \quad (3.10)$$

where:

- $\mathbf{Y}(n) = [Y_1(n), \dots, Y_r(n)]^T$ :  $r$ -dimensional ( $r$ -D) random vector.
- $\{\mathbf{Z}(n)\}$ :  $r$ -D non-stationary white noise vector:
  - $E[\mathbf{Z}(n)] = \mathbf{0}$
  - $\sum \mathbf{z}(n)\mathbf{z}(n+k) = \mathbf{Q}_z(n)\delta(k)$
- $\{\mathbf{H}(n)\}$ : sequence of known  $r \times r$  matrices.

See the example discussed in Section 3.4.

**Block diagram:**



**Initialization**

$\mathbf{Y}(0)$  is a random vector specified by its expectation  $\mu_{\mathbf{Y}(0)}$  and covariance matrix  $\Sigma_{\mathbf{Y}(0)\mathbf{Y}(0)}$ .

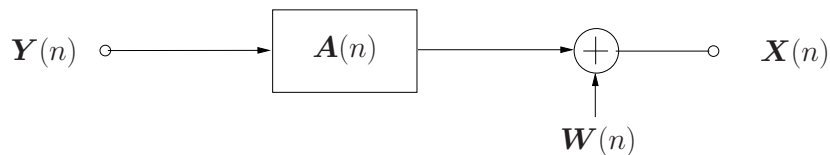
- **Observation Model**

$$\mathbf{X}(n) = \mathbf{A}(n)\mathbf{Y}(n) + \mathbf{W}(n), n = 1, 2, \dots \quad (3.11)$$

where:

- $\mathbf{X}(n) = [X_1(n), \dots, X_s(n)]^T$ :  $s$ -D random vector.
- $\{\mathbf{W}(n)\}$ :  $s$ -D non-stationary white noise vector with auto-covariance
 
$$\Sigma_{\mathbf{W}(n)\mathbf{W}(n+k)} = \mathbf{Q}_{\mathbf{W}(n)}\delta(k)$$
- $\{\mathbf{A}(n)\}$ : sequence of known  $s \times r$  matrices.

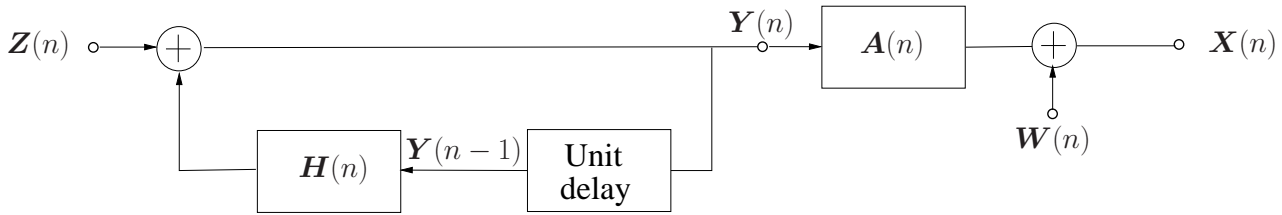
**Block Diagram:**



- **Additional independence assumption**

$\mathbf{Y}(0)$ ,  $\{\mathbf{Z}(n)\}$ , and  $\{\mathbf{W}(n)\}$  are uncorrelated.

- **Complete Signal Model**



### 3.2.2 Equation of the vector Kalman filter

Let us define

- $\hat{\mathbf{Y}}(n | n) \equiv$  LMMSEE of  $\mathbf{Y}(n)$  based on  $\mathbf{X}(1), \dots, \mathbf{X}(n)$
- $\hat{\mathbf{Y}}(n + 1 | n) \equiv$  LMMSEE of  $\mathbf{Y}(n + 1)$  based on  $\mathbf{X}(1), \dots, \mathbf{X}(n)$
- $\hat{\mathbf{X}}(n + 1 | n) \equiv$  LMMSEE of  $\mathbf{X}(n + 1)$  based on  $\mathbf{X}(1), \dots, \mathbf{X}(n)$
- $\mathbf{R}(n | n) \equiv E[(\mathbf{Y}(n) - \hat{\mathbf{Y}}(n | n))(\mathbf{Y}(n) - \hat{\mathbf{Y}}(n | n))^T]$
- $\mathbf{R}(n + 1 | n) \equiv E[(\mathbf{Y}(n + 1) - \hat{\mathbf{Y}}(n + 1 | n))(\mathbf{Y}(n + 1) - \hat{\mathbf{Y}}(n + 1 | n))^T]$

We can apply the same reasoning as used for the scalar Kalman filter to show that the recursive equations of the vector Kalman filter are given as follows.

- **Recursive equations of the Kalman filter**

*Prediction Step :*

$$\begin{aligned}\hat{\mathbf{Y}}(n+1 | n) &= \mathbf{H}(n+1)\hat{\mathbf{Y}}(n | n) \\ \hat{\mathbf{X}}(n+1 | n) &= \mathbf{A}(n+1)\hat{\mathbf{Y}}(n+1 | n) \\ \mathbf{R}(n+1 | n) &= \mathbf{H}(n+1)\mathbf{R}(n | n)\mathbf{H}(n+1)^T + \mathbf{Q}_z(n+1)\end{aligned}$$

*Updating Step :*

$$\begin{aligned}\hat{\mathbf{Y}}(n+1 | n+1) &= \hat{\mathbf{Y}}(n+1 | n) + \mathbf{B}(n+1)[\mathbf{X}(n+1) - \hat{\mathbf{X}}(n+1 | n)] \\ \mathbf{R}(n+1 | n+1) &= [\mathbf{I} - \mathbf{B}(n+1)\mathbf{A}(n+1)]\mathbf{R}(n+1 | n)\end{aligned}$$

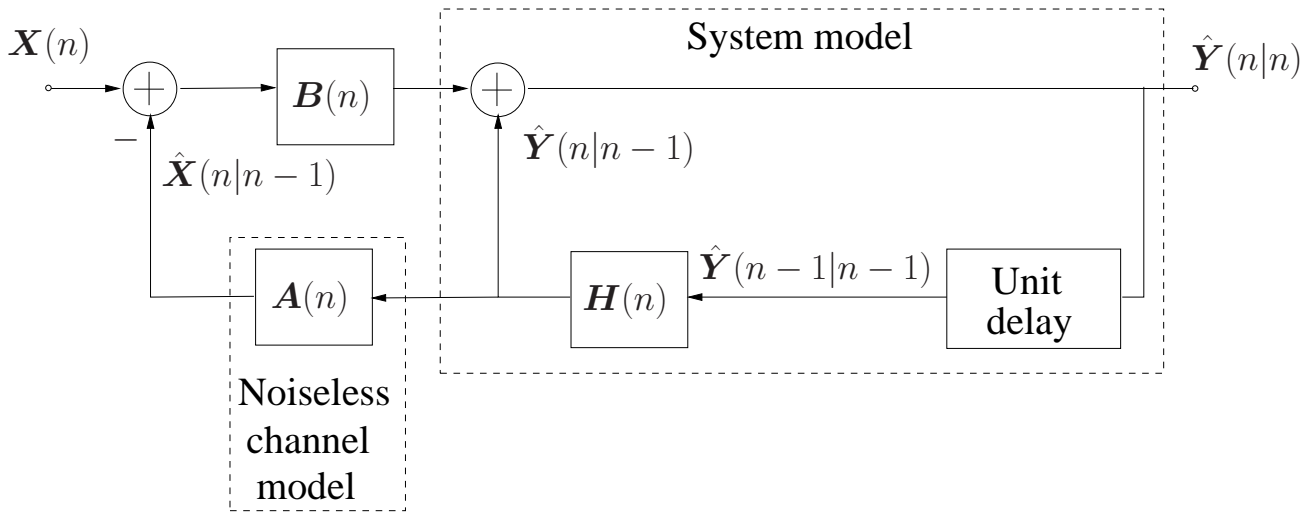
with the Kalman matrix

$$\mathbf{B}(n+1) \equiv \mathbf{R}(n+1 | n)\mathbf{A}(n+1)^T[\mathbf{A}(n+1)\mathbf{R}(n+1 | n)\mathbf{A}(n+1)^T + \mathbf{Q}_w(n+1)]^{-1}$$

*Initialization :*

$$\begin{aligned}\hat{\mathbf{Y}}(0 | 0) &= \mu_{\mathbf{Y}}(0) \\ \mathbf{R}(0 | 0) &= \sum_{\mathbf{Y}(0)} \mathbf{Y}(0)\end{aligned}$$

- **Block diagram of the vector Kalman filter**





### 3.3 Example of a recursive estimator

- **Signal model**

$$X(n) = Y + W(n) \quad n = 1, 2, 3, \dots$$

Where:

- $Y$  is an unknown constant to be estimated based on the observation of  $\{X(n)\}$ .
- $\{W(n)\}$  is a white noise sequence.

- **Arithmetic mean**

An appealing linear estimator for  $Y$  is the *arithmetic mean*

$$\hat{Y}(n) = \frac{1}{n} \sum_{m=1}^n X(m)$$

**Drawback:** To compute  $\hat{Y}(n)$  based on the above formula,  $X(1), \dots, X(n)$  need to be stored. The required memory grows linearly with  $n$ .

- **Recursive implementation**

$$\hat{Y}(n+1) = \frac{1}{n+1} \sum_{m=1}^n X(m) + \frac{1}{n+1} X(n+1)$$

$$\hat{Y}(n+1) = \frac{n}{n+1} \hat{Y}(n) + \frac{1}{n+1} X(n+1)$$

This estimator requires storage of one value, i.e.  $\hat{Y}(n)$ , only.

### 3.4 Example of a signal model: Target tracking

- **Equations of the movement of a target:**

**Position:**  $U(t) = \int_0^t V(t') dt' + U(0), \quad U(0): \text{initial position}$

**Velocity:**  $V(t) = \int_0^t G(t') dt' + V(0), \quad V(0): \text{initial velocity}$

**Acceleration:**  $G(t)$  is assumed to be white noise.

- **Discrete-time model:**

$$\frac{du}{dt}(t) = V(t) \quad \frac{du}{dt}(nT_s) \approx [U((n+1)T_s) - U(nT_s)]/T_s$$

$$\frac{dv}{dt}(t) = G(t) \quad \frac{dv}{dt}(nT_s) \approx [V((n+1)T_s) - V(nT_s)]/T_s$$

$$U((n+1)T_s) - U(nT_s) = V(nT_s) \cdot T_s \quad T_s: \text{Sampling interval}$$

$$V((n+1)T_s) - V(nT_s) = \tilde{G}(nT_s) \cdot T_s \quad \tilde{G} = \tilde{G}(t) \text{ low-pass filtered with bandwidth } \frac{1}{2T_s}.$$

**State model**

$$\underbrace{\begin{bmatrix} U(nT_s) \\ V(nT_s) \end{bmatrix}}_{\mathbf{Y}(n)} = \underbrace{\begin{bmatrix} 1 & T_s \\ 0 & 1 \end{bmatrix}}_{\mathbf{H}(n)} \underbrace{\begin{bmatrix} U((n-1)T_s) \\ V((n-1)T_s) \end{bmatrix}}_{\mathbf{Y}(n-1)} + \underbrace{\begin{bmatrix} 0 \\ T_s \tilde{G}((n-1)T_s) \end{bmatrix}}_{\mathbf{Z}(n)}$$

with

$$\mathbf{Y}(0) = [U(0), V(0)]^T,$$

$$\mathbf{Q}_Z(n) = \begin{bmatrix} 0 & 0 \\ 0 & T_s^2 E[\tilde{G}(nT_s)^2] \end{bmatrix}.$$

***Observation model***

$$X(n) = U(nT_s) + \underbrace{W(n)}_{\text{Measurement error}}$$

$$X(n) = \underbrace{[1 \quad 0]}_{\mathbf{A}(n)} \mathbf{Y}(n) + W(n)$$

where  $W(n)$  is white noise with variance  $\sigma_{WW}^2$ .